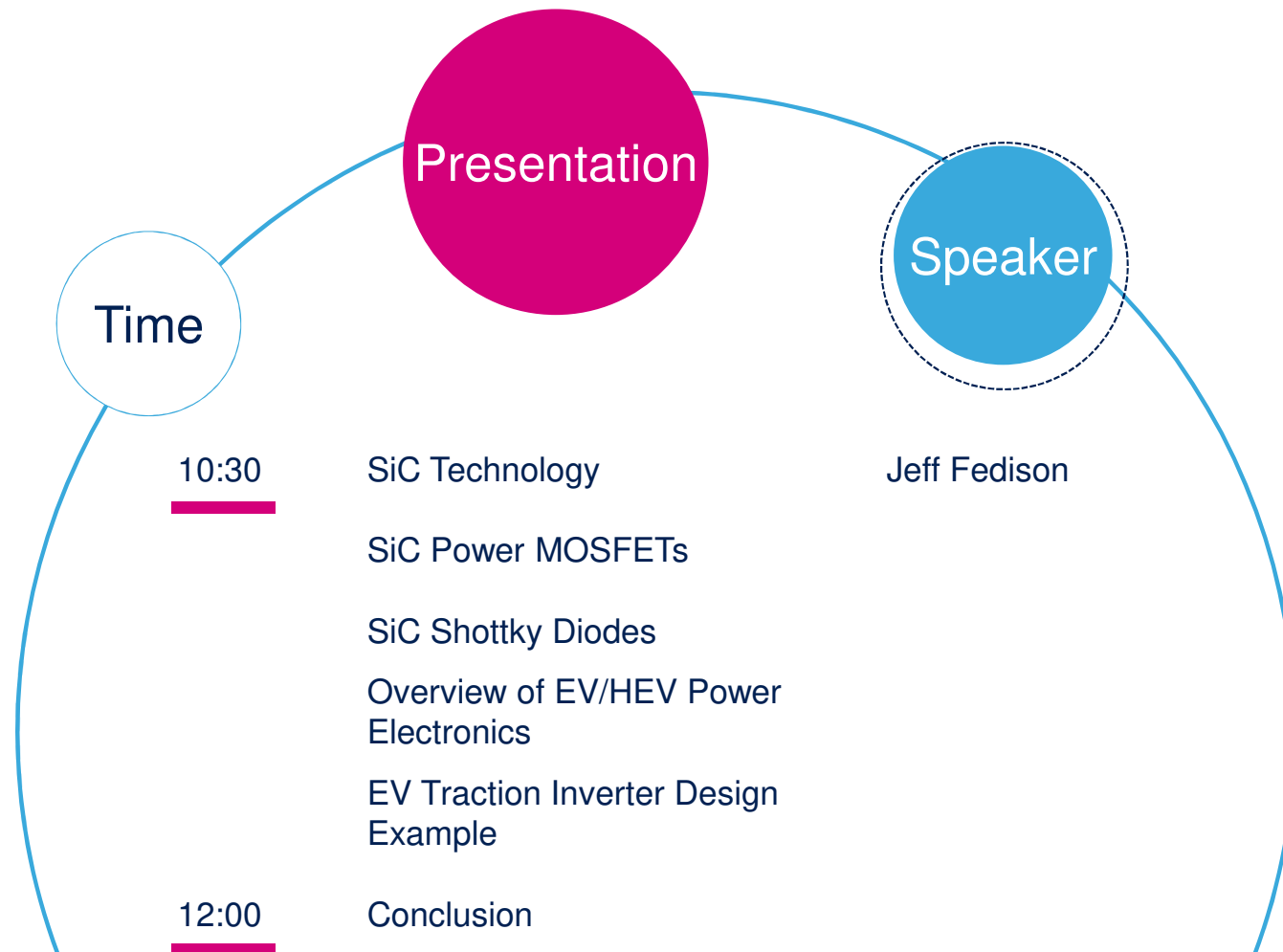
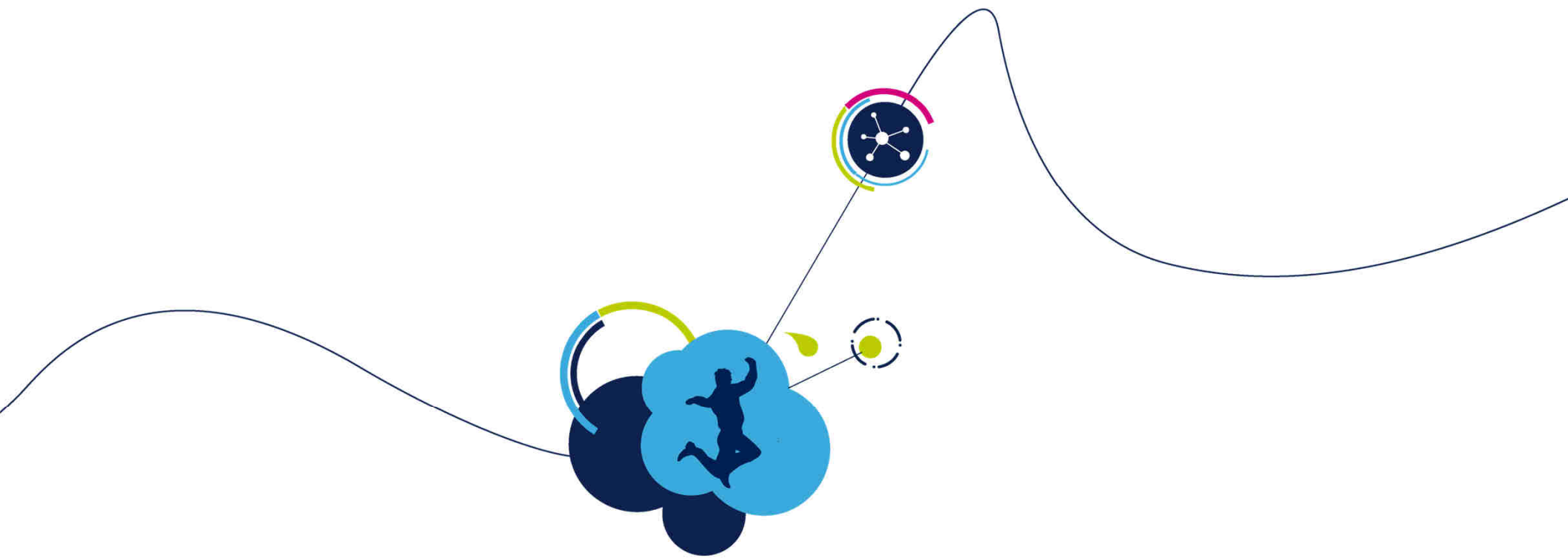




# Agenda

2





# SiC Technology

# What is Silicon Carbide?

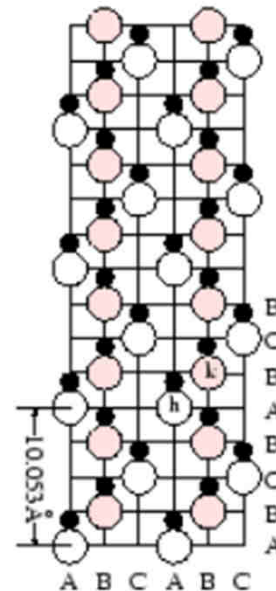
4

- 4H-SiC and 6H-SiC are both commercially available
- 4H-SiC is most important for power electronics

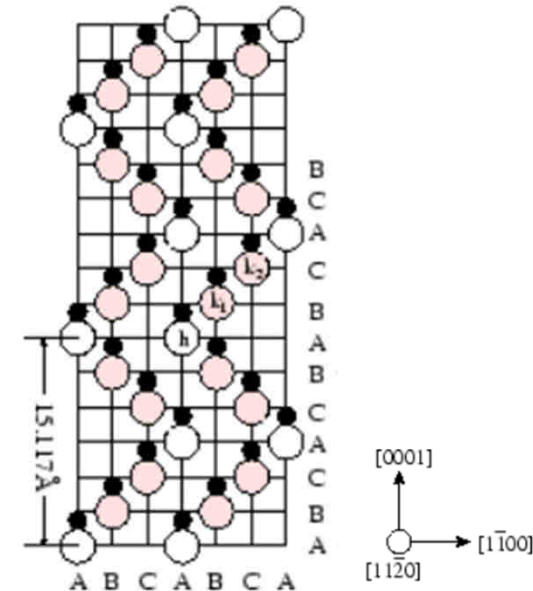
## 4H-SiC

- Is a wide bandgap semiconductor material
- Remains a solid up to 2830°C
- Is available in semiconductor grade wafers up to 6 inches in diameter

4H-SiC



6H-SiC



# What makes 4H-SiC useful for Power Electronics?

- Can be doped both p-type and n-type
- High electron mobility
- SiO<sub>2</sub> is native oxide
- 3x higher thermal conductivity vs Si
- Large band gap energy allows very high temperature operation
- **High critical electric field, 10x that of silicon!**

# Wide Bandgap Materials

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## Radical innovation for Power Electronics

	Si	GaN	4H-SiC
$E_g$ (eV) – Band gap	1.1	3.4	3.3
$V_s$ (cm/s) – Electron saturation velocity	$1 \times 10^7$	$2.2 \times 10^7$	$2 \times 10^7$
$\epsilon_r$ – dielectric constant	11.8	10	9.7
$E_c$ (V/cm) – Critical electric field	$3 \times 10^5$	$2.2 \times 10^6$	$2.5 \times 10^6$
$k$ (W/cm K) thermal conductivity	1.5	1.7	5

$E_c$  → low on resistance

$E_g$  → low leakage, high  $T_j$

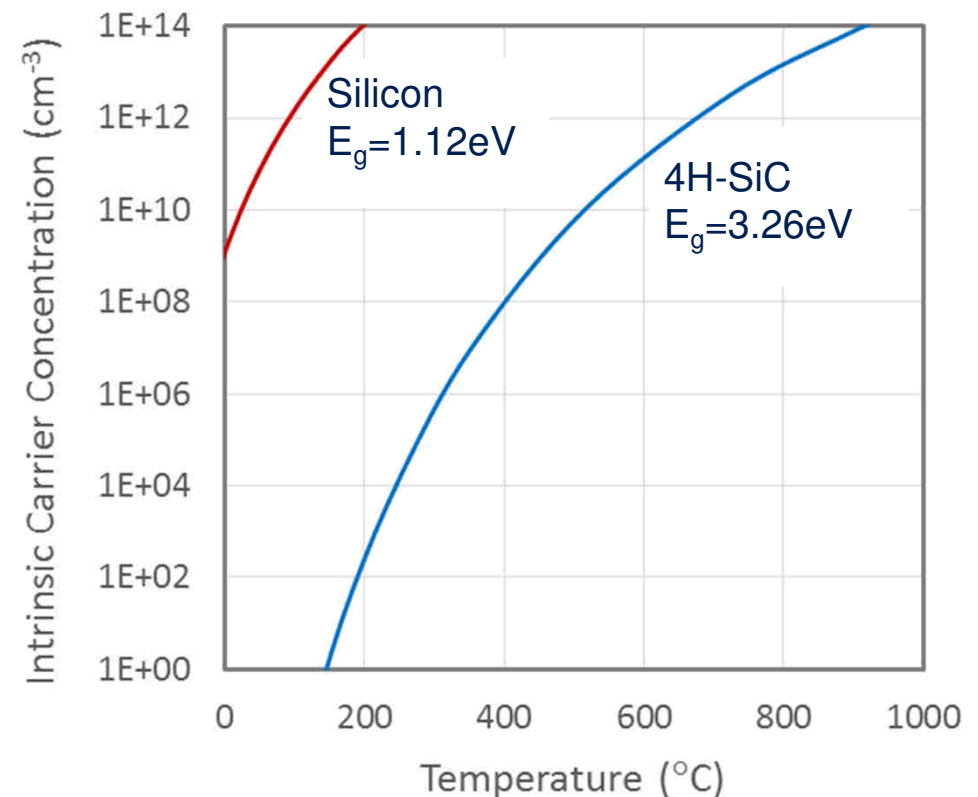
$k$  → Operation > 200 °C  
Reduced Cooling Requirements

$V_s$  → Higher switching frequency  
Lower switching losses

# Intrinsic carrier concentration, $n_i$ SiC vs. Silicon

7

- Intrinsic carriers are thermally generated and increase in number at higher temperatures
- Because of its larger band gap energy, SiC maintains low intrinsic carrier concentration up to 900°C



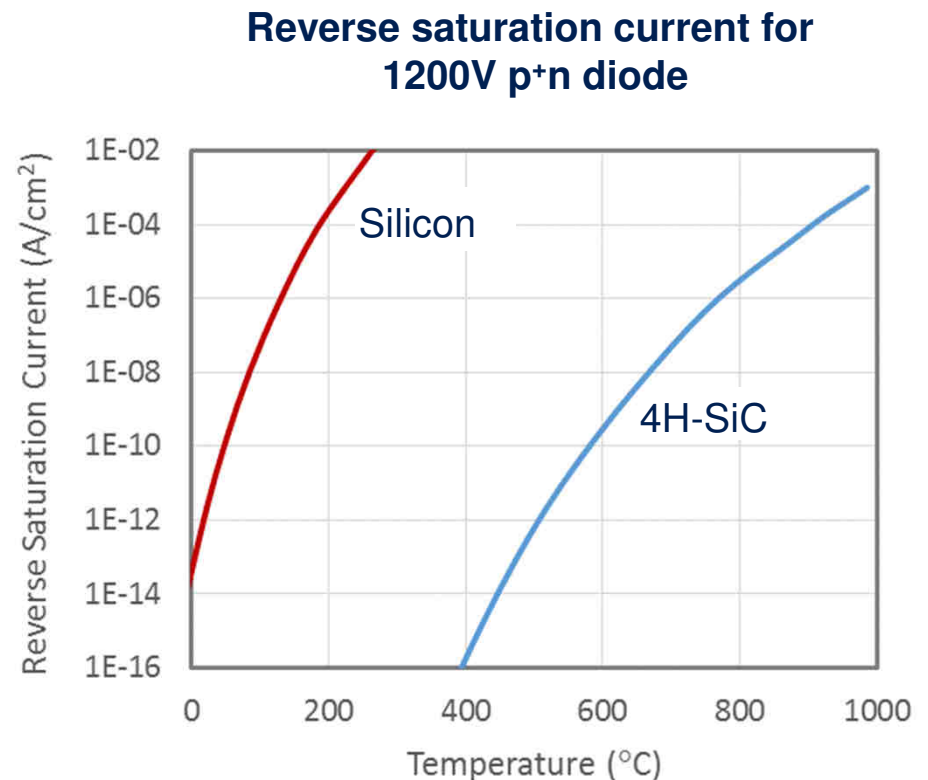
# SiC provides low reverse leakage current up to high temperature

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Reverse saturation current of a p<sup>+</sup>n diode:

$$J_S = q n_i^2 \left( \frac{1}{N_D} \sqrt{\frac{D_p}{\tau_p}} \right)$$

- Silicon becomes unusable above ~ 250°C due to high leakage current
- 4H-SiC has low reverse leakage current up to 900°C





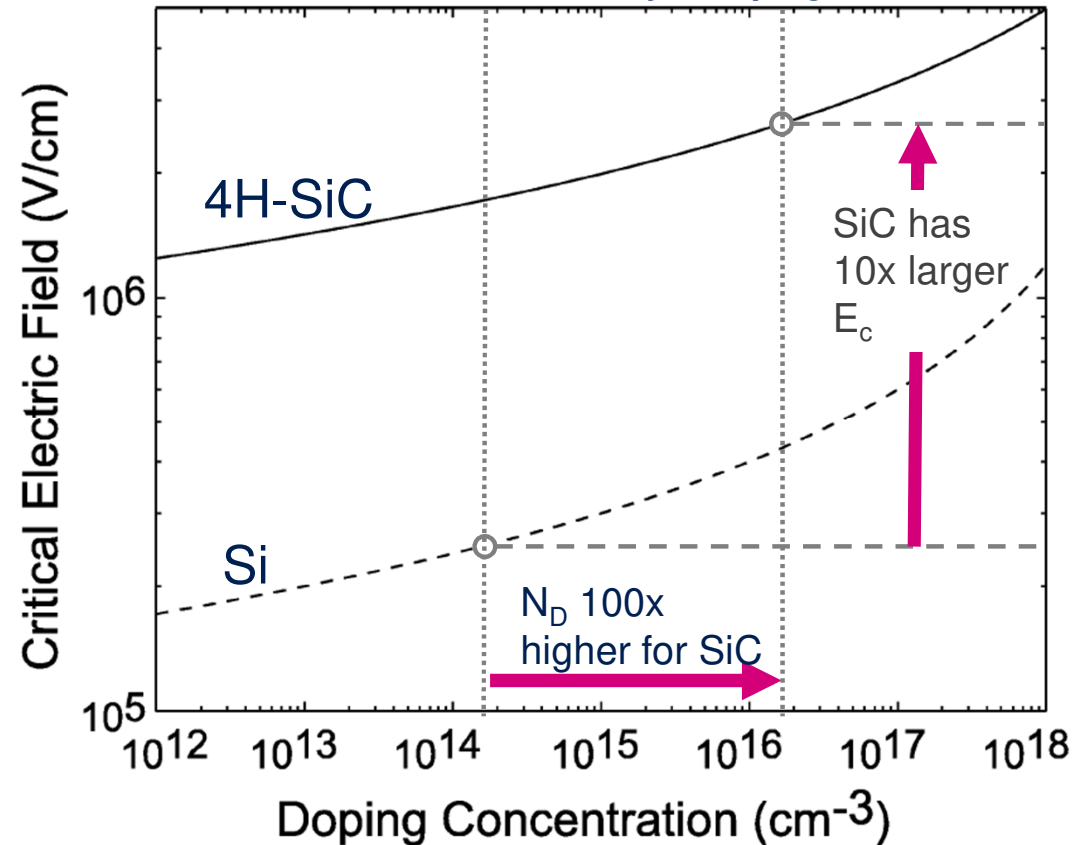
# Critical electric field

9

- For a given breakdown voltage, the larger critical electric field of SiC enables much higher drift layer doping vs Si

Material	Drift layer doping for BV=1200V
Si	$1.5 \times 10^{14} \text{ cm}^{-3}$
SiC	$1.6 \times 10^{16} \text{ cm}^{-3}$

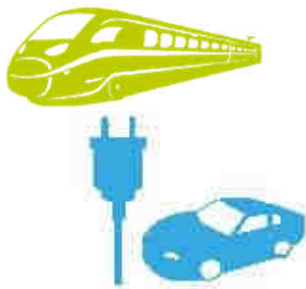
Critical electric field vs n-drift layer doping concentration



Note: circles show  $N_D$  required for 1200V NPT design

# Key Applications for SiC

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HEV/EV &  
Traction



SMPS  
& PFC



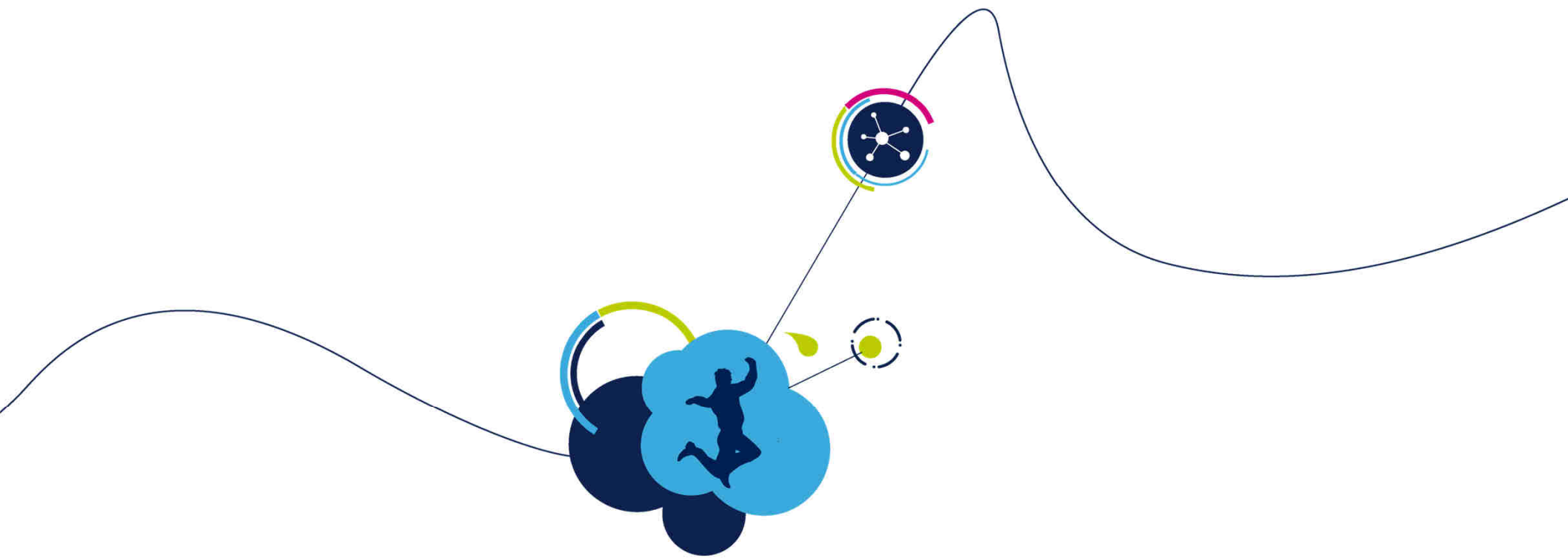
Solar  
Inverter



OBC &  
Charging  
station



Industrial  
Drives



# SiC Power MOSFETs

# Benefits of SiC MOSFETs

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## Key Benefits



### Extremely low Switching Losses and Ultra-Low $R_{DS(on)}$

Higher operating frequency for smaller and lighter systems



### Good Thermal Performance

High operating temperature (  $T_{jmax} = 200^{\circ}C$  )  
Reduced cooling requirements & heat-sink, Increased lifetime



### Easy to Drive

Fully compatible with standard Gate Drivers

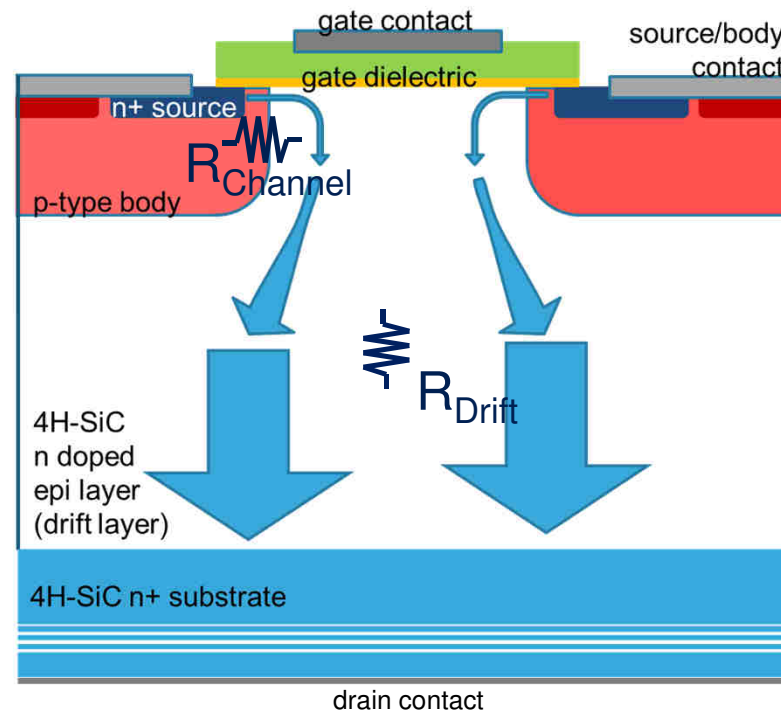


### Very fast and robust intrinsic body diode

Separate antiparallel diode not required

# High-Voltage DMOSFET Structure

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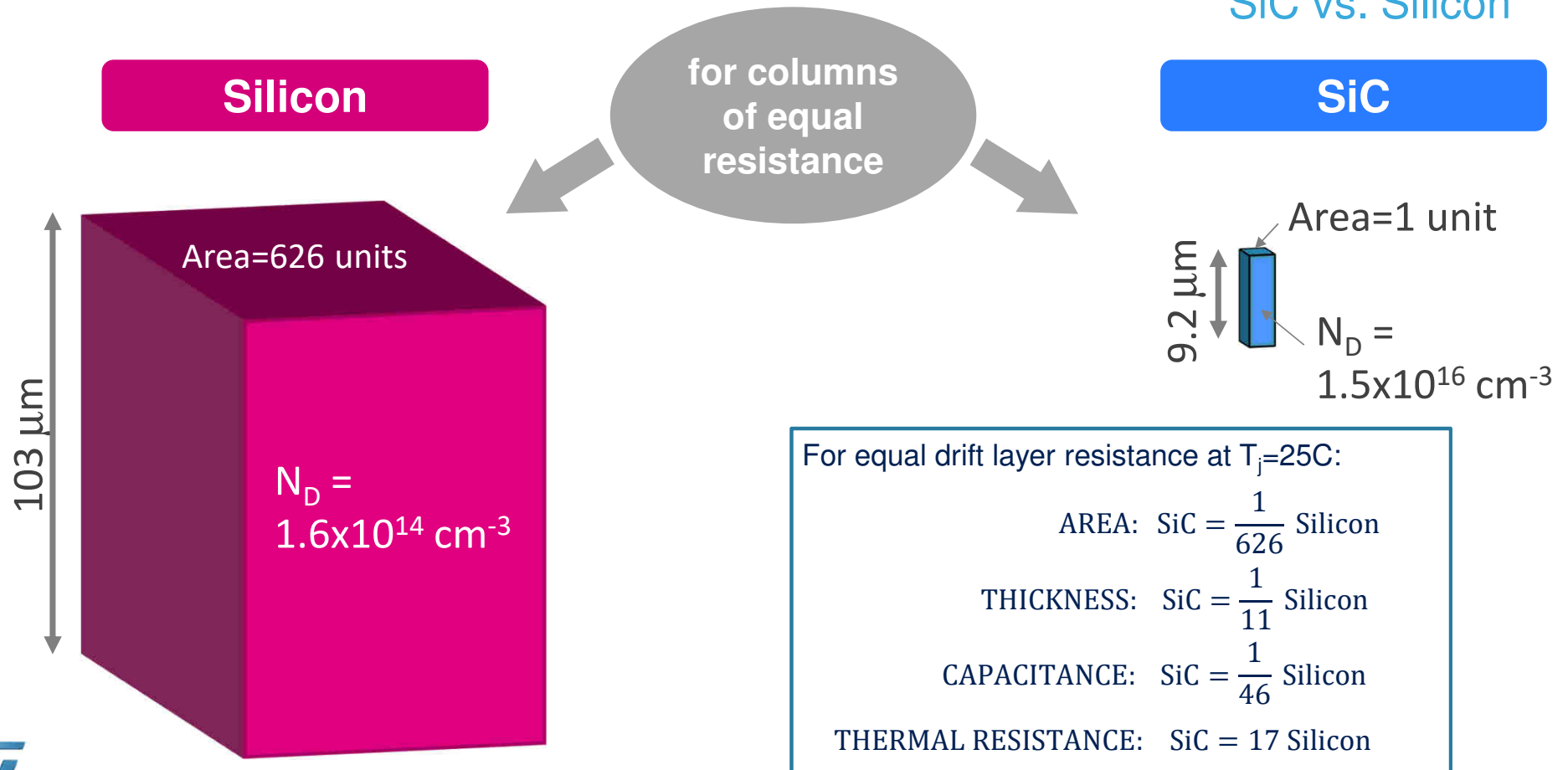


$R_{\text{DS(on)}}$  is determined mainly by  $R_{\text{Drift}}$  and to lesser extent  $R_{\text{Channel}}$

# Drift layer dimensions for 1200V DMOSFET

14

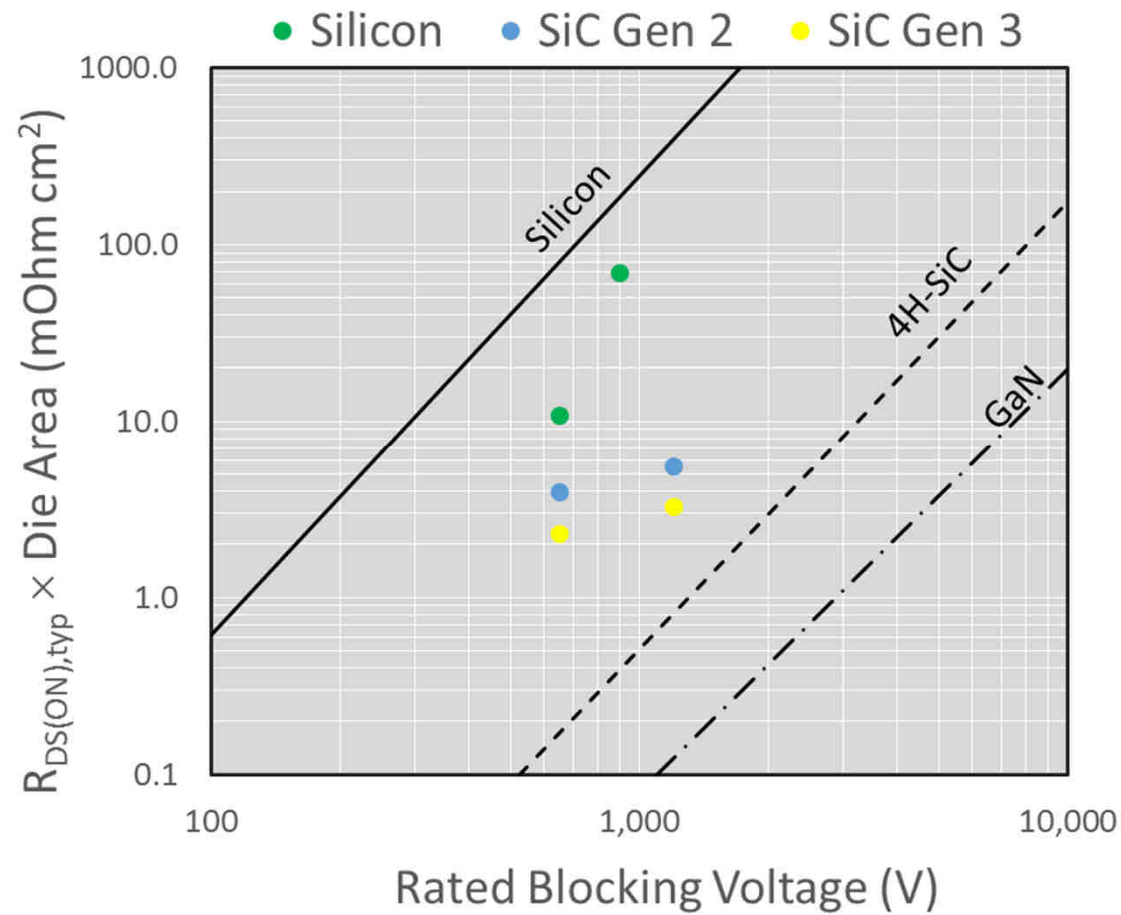
SiC vs. Silicon



SiC offers dramatic reduction in device footprint!

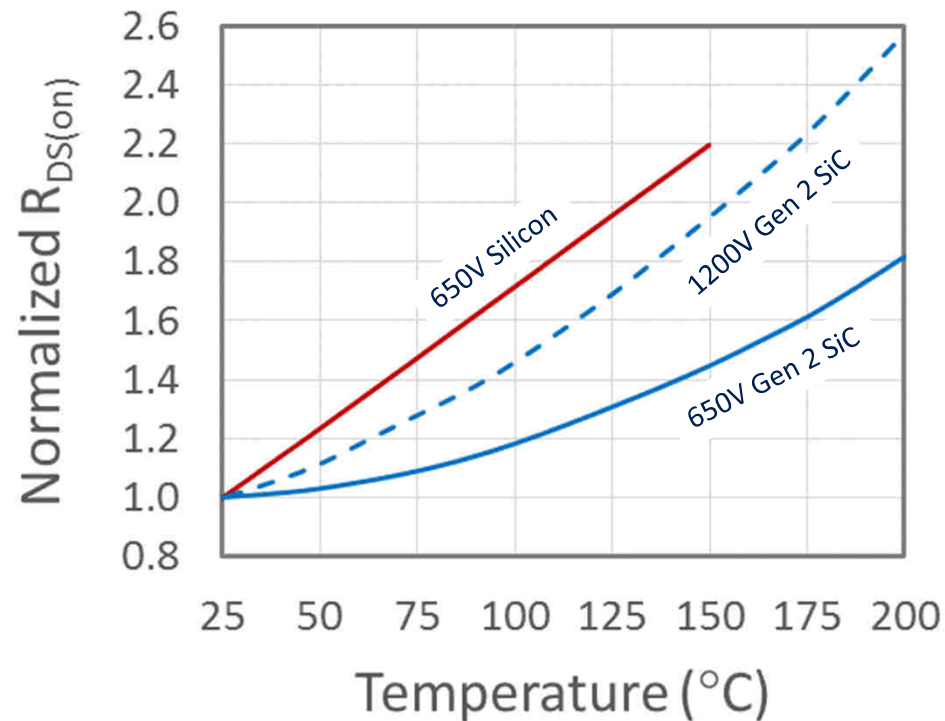
# MOSFET $R_{DS(on)}$ Figure of Merit

15



# $R_{DS(on)}$ variation with temperature

16

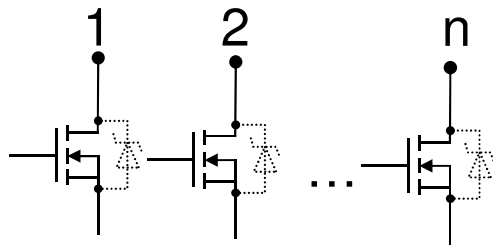


ST is the only supplier to guarantee max  $T_j$  as high as 200°C in plastic package

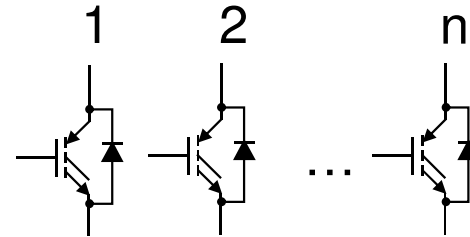


# SiC MOSFET Allows Lowest Conduction Losses

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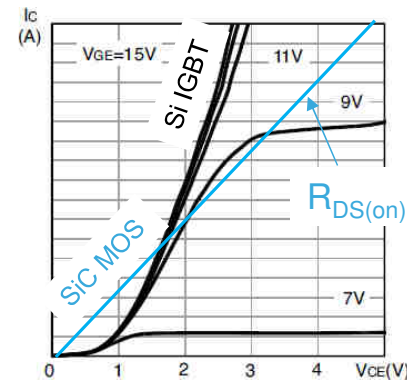


When “n” MOSFET are paralleled the total  $R_{DS(on)}$  must be divided by “n” allowing ideally zero conduction losses



When “n” IGBTs are paralleled the  $V_{ce(sat)}$  doesn't decrease linearly, the minimum achievable on-state voltage drop is about 0.8 – 1V

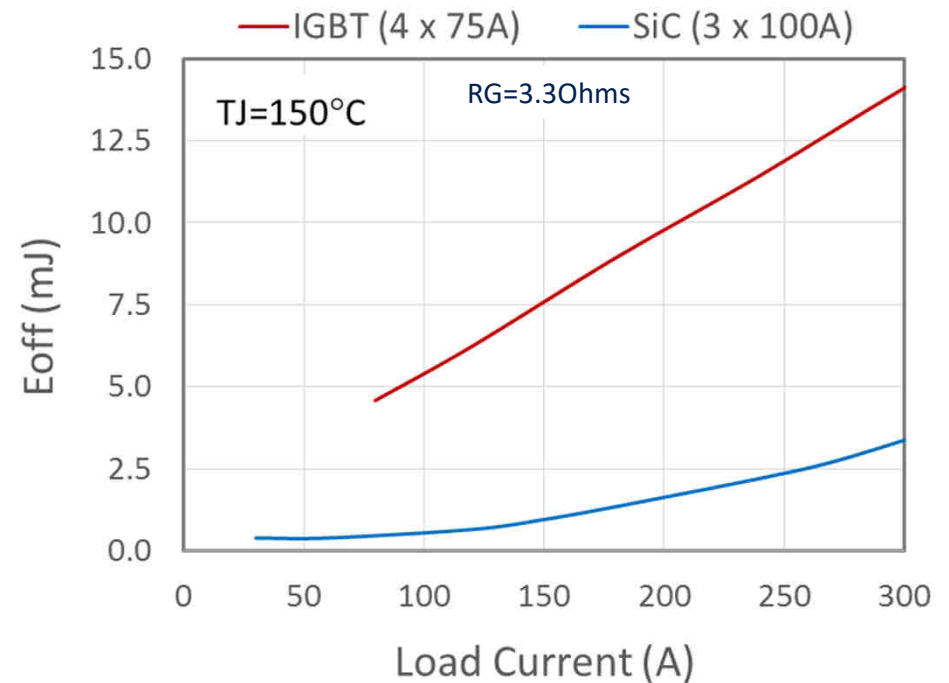
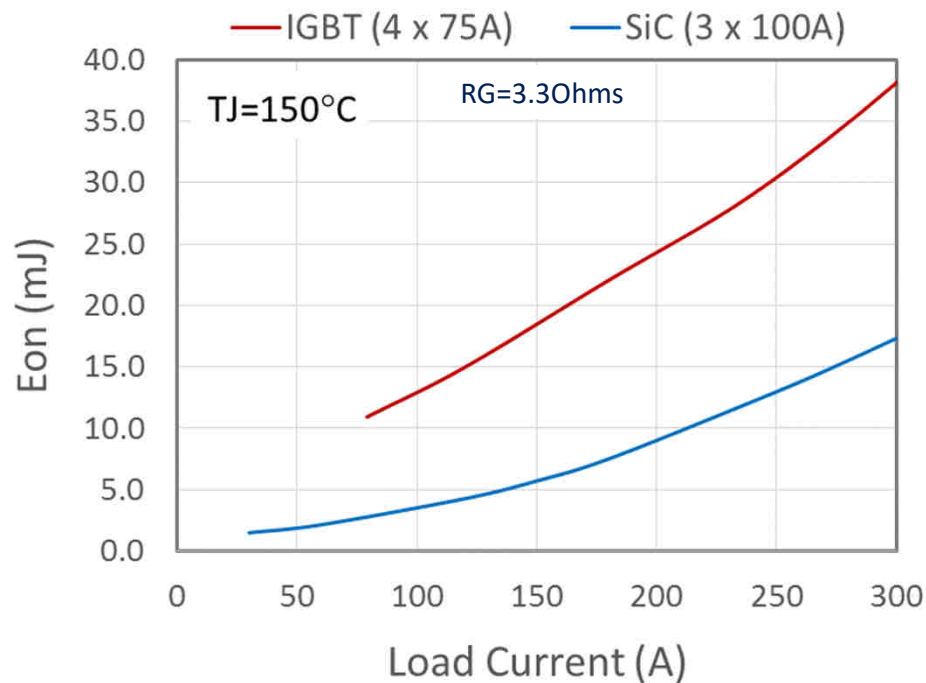
The lowest possible conduction losses can be achieved only with MOSFETs



# Switching energies for 1200V rated devices

## IGBT vs SiC MOSFET





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IGBT: 4 x FGY75T120SQDN-D (1200V, 75A, 46 mm<sup>2</sup> per die)  
SiC MOSFET: 3 x SCT110N120G3D2AG (1200V, 100A, 26 mm<sup>2</sup> per die)



# SiC MOSFETs in Production

$V_{DS}$ [V]	$R_{DS(on)}$ @ 25 °C [Ω]	$I_d$	Package	P/N
1200	0.052	65	HiP247, H2PAK-7	SCT50N120 SCTWA50N120 SCTH50N120-7
	0.08	80	HiP247, H2PAK-2	SCT30N120 SCTWA30N120 SCT30N120H
	0.169	20	HiP247, H2PAK-2	SCT20N120 SCT20N120AG SCTWA20N120 SCT20N120H 
	0.52	12	HiP247, H2PAK-2	SCT10N120 SCT10N120AG SCTWA10N120 SCT10N120H 
650	0.045	45	HiP247, H2PAK-7	SCTH35N65G2V-7 SCTH35N65G2V-7AG SCTW35N65G2V 
	0.018	90	HiP247, H2PAK-7	SCTW90N65G2V SCH90N65G2V-7
	0.020	100	HiP247	SCTE100N65G2AG 



HiP-247 (STD & LL) <sup>TM</sup>



H2PAK  
2 and 7 leads

- HiP-247 rated at 200°C Tj max
- H2PAK-7L (with kelvin source) SMD option (175°C Tj max)





Automotive Grade


# Silicon Carbide MOSFET Technology Roadmap

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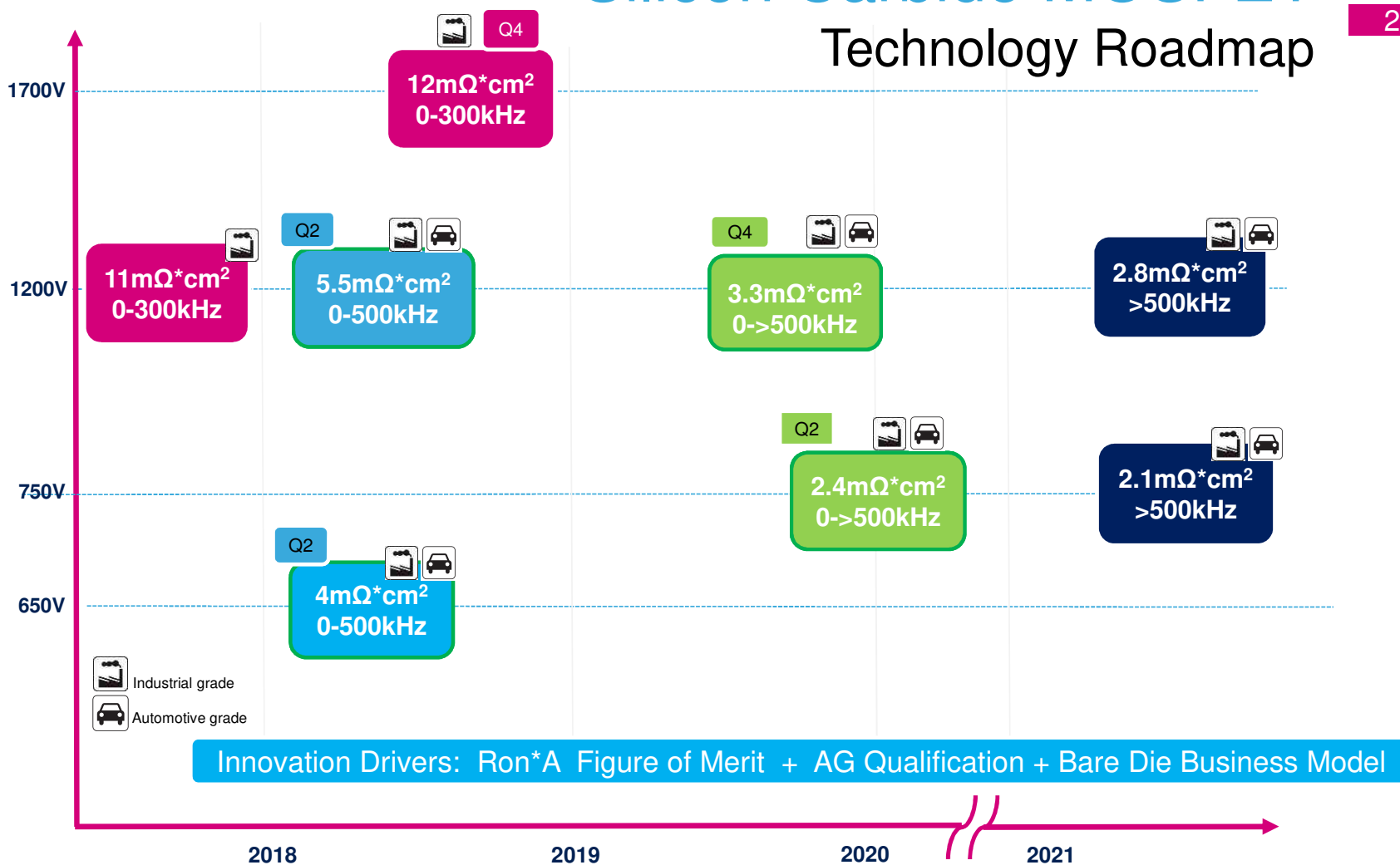
**1st Gen. Planar**

**2nd Gen. Planar** AG-REV.D 

**3rd Gen. Planar** AG-REV.D 

**Trench Gate** AG-REV.D 

**Traction Inverter**  
**OBC & Chargers**  
 UPS /Solar/Welding  
**SMPS/high end PFC**  
 DC-DC converters  
**high end PFC**  
 Auxiliary Power Supply



# SiC MOSFET driving requirements

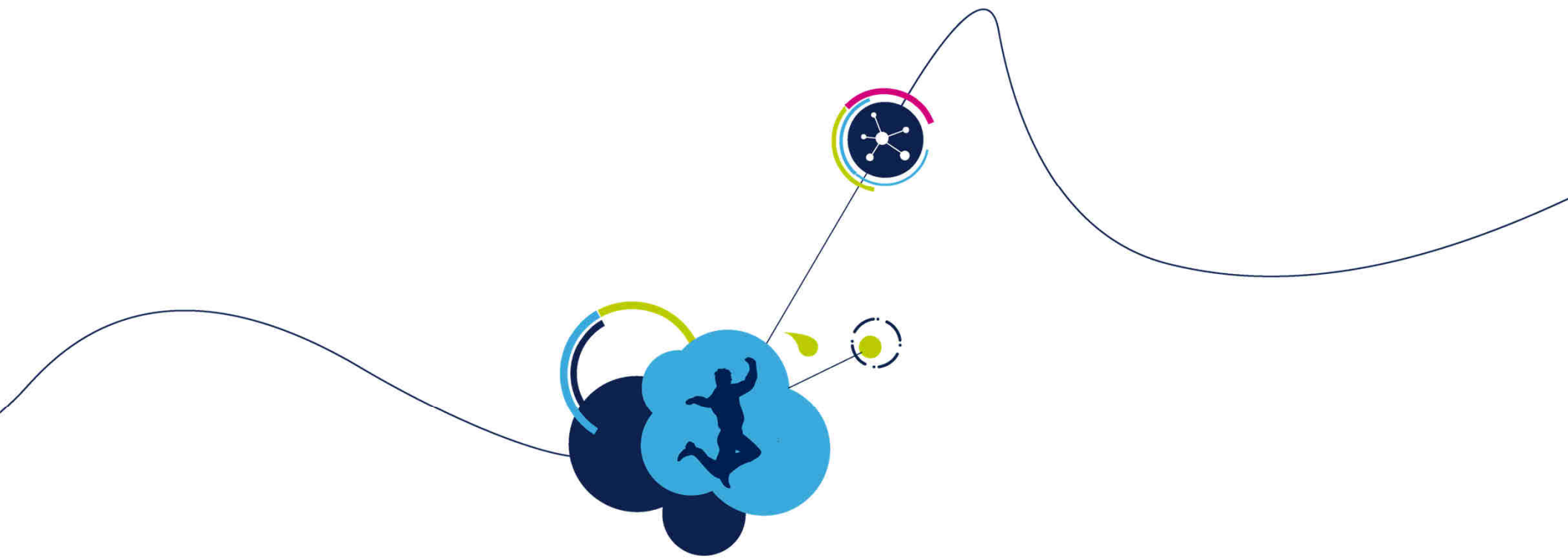
21

- Driving a SiC MOSFET is almost as easy as driving a silicon MOSFET:
  - Just need  $V_{GS} = 20V$  to get the right  $R_{DS(on)}$
  - Adequate current capability to ensure high speed (2-3 A would be the best)
- Very simple and very mature standard gate drivers can be used
  - ST TD350 + push-pull stage (to increase current capability) in production
  - The new ST isolated GAPdriver available now
- A detailed *Application Note* focused on “how to drive a SiC MOSFET” has been published on st.com: **AN4671**

# SiC MOSFET benefits

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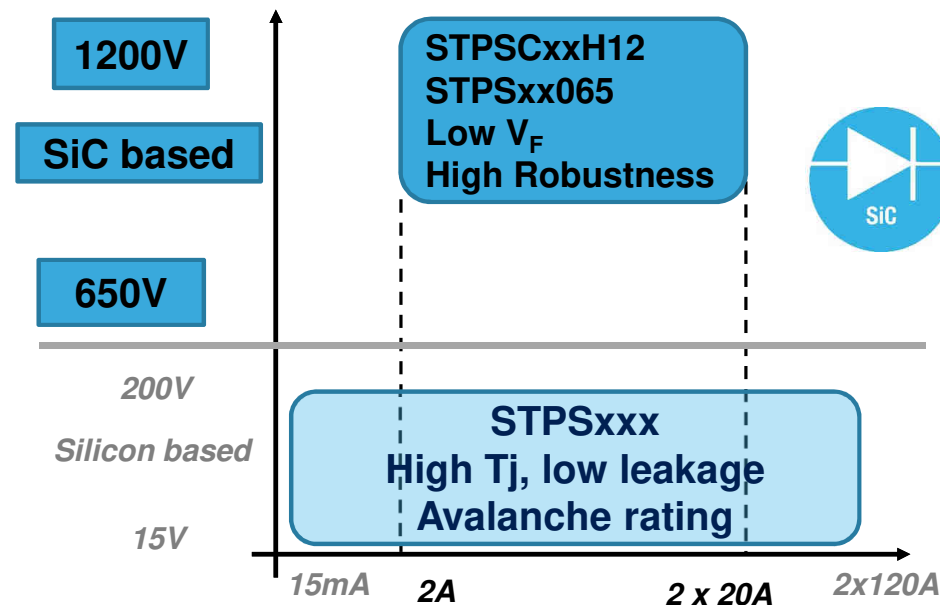
- Switching losses are dramatically reduced even in hard-switching topologies
- Unlike the IGBT, the MOSFET has no turn-on knee voltage giving low conduction losses across the entire load range
- The ONLY SiC alternative that offers intrinsic body diode with very low reverse recovery charge
- Easy to drive – use of conventional gate drivers ensures low component count
- Reliability – Very good final Result and qualified @ 200 °C!



# SiC Schottky Diodes

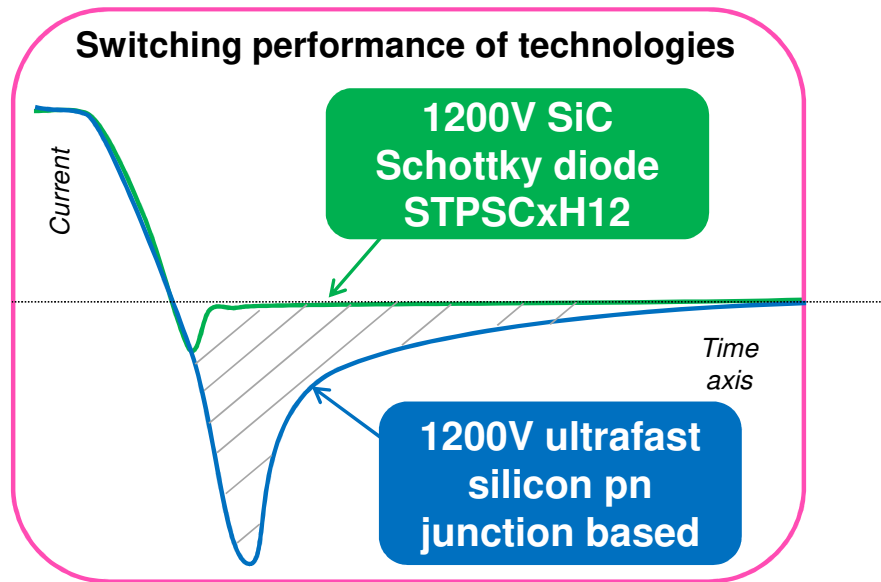
Enabling 1200V Diodes with High Efficiency

SiC allows Schottky diode structure to be used at much higher breakdown voltage vs Silicon versions





# Improved switching performance is key advantage of SiC Schottky diodes

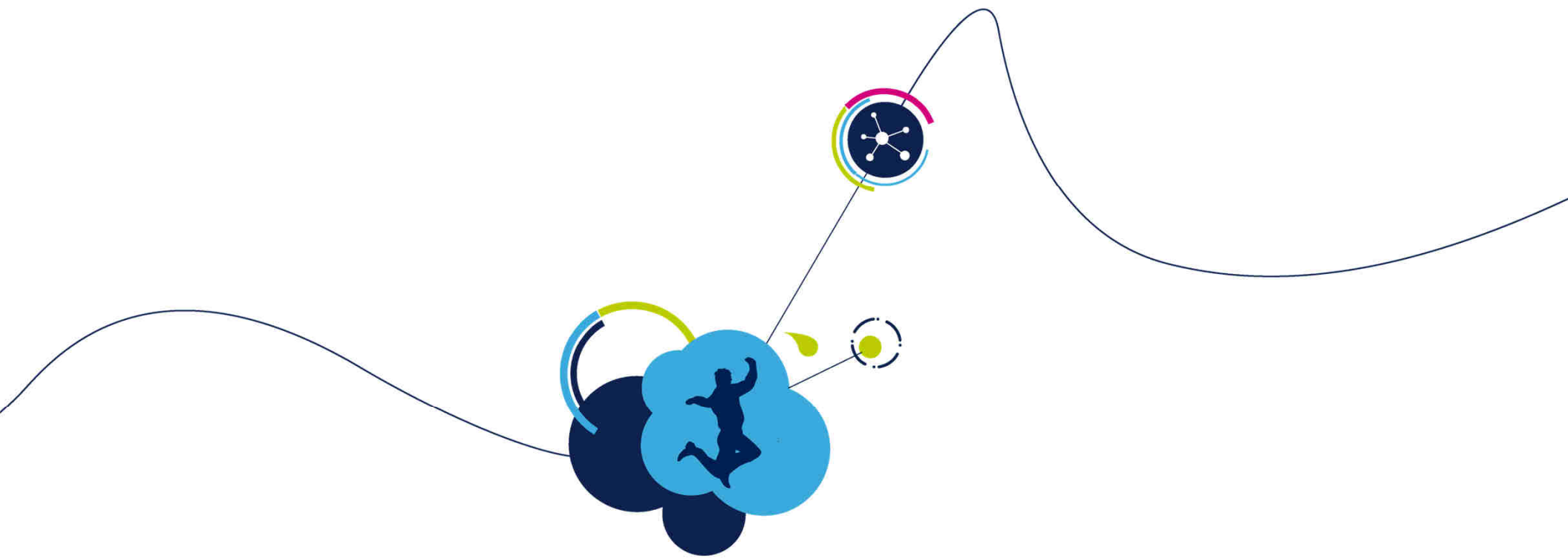


**1200V SiC diodes provide higher system efficiency, remain cooler during operation**

# Key features of latest ST SiC Schottky diode technology

## 1200V SiC Schottky Diode Technology

- Negligible reverse recovery loss
- Based on new design to improve  $V_F/I_{FSM}$  trade-off
- Wide range of available sizes: 2A to 40A.
- Low  $V_F$  (1200V SiC Schottky has best-in-class  $V_F$ )
- High forward and reverse surge robustness
- Low leakage current for higher reliability
- AEC-Q101 versions available



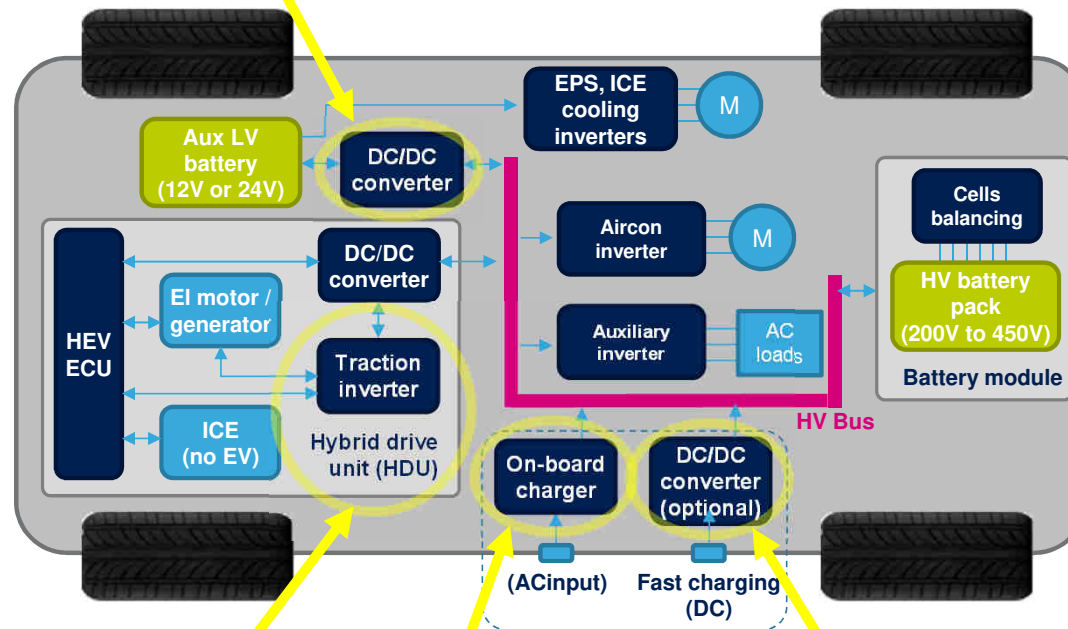
# Overview of EV/HEV Power Electronics

# e-Vehicle block diagram

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HEV/EV

Output Power: **4kW**  
IGBT → SiC MOSFET  
50kHz – 200kHz



- Silicon content
- Mechanical or electro-mechanical
- Batteries

Output Power (EV): **80kW — 250kW**  
IGBT → SiC MOSFET  
12kHz and higher

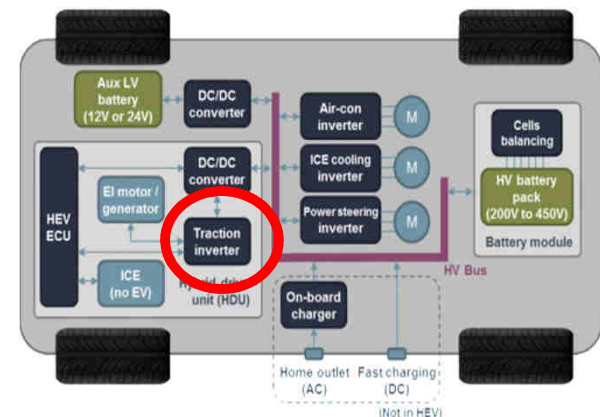
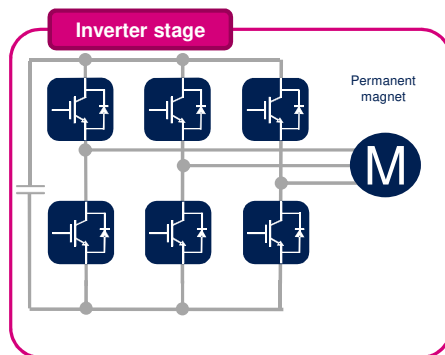
Output Power: **20kW**  
SiC MOSFET/SiC SBD  
50kHz – 200kHz

Output Power: **50kW**  
Si MOSFET → SiC MOSFET  
50kHz – 200kHz

# Main inverter for HEV/EV

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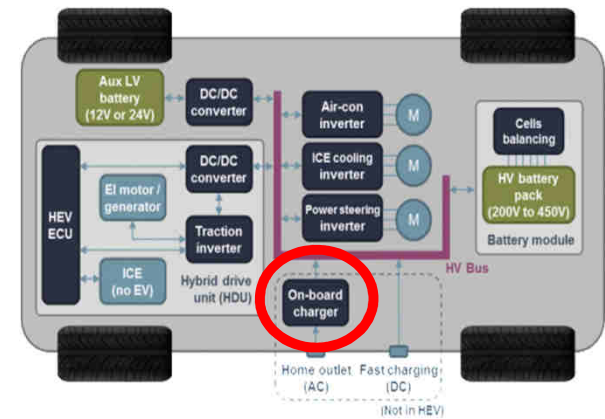
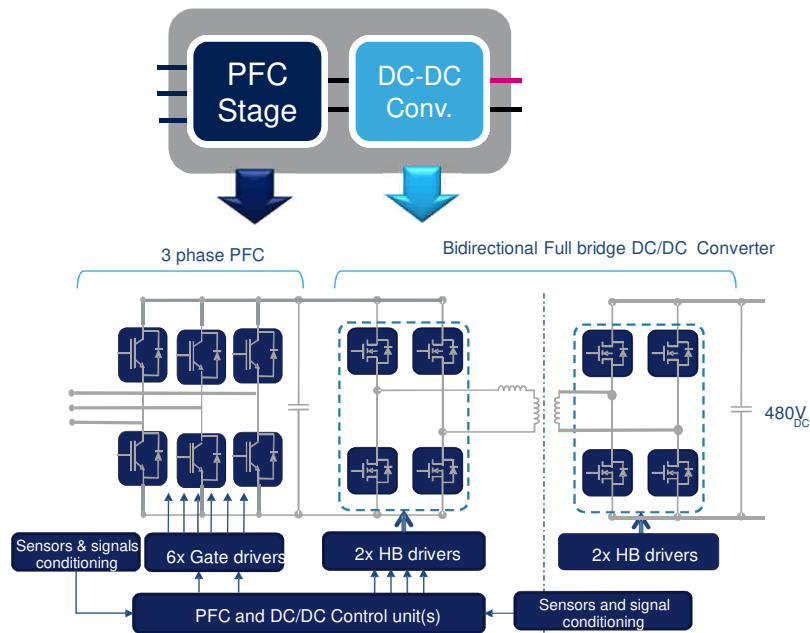
- Usually 3-phase permanent magnet motors are used for traction
- Operating voltage from 48V to 800V
- Bi-directional
  - Feed the electric motor when driving the wheels
  - Stream energy back to HV bus when breaking vehicle
- Nominal power ranging from 10kW (ICE assistance) to 250kW (pure EV)



For bus up to 400V → SiC MOS 650V  
For bus in the range [400V-800V] → SiC MOS 1200V

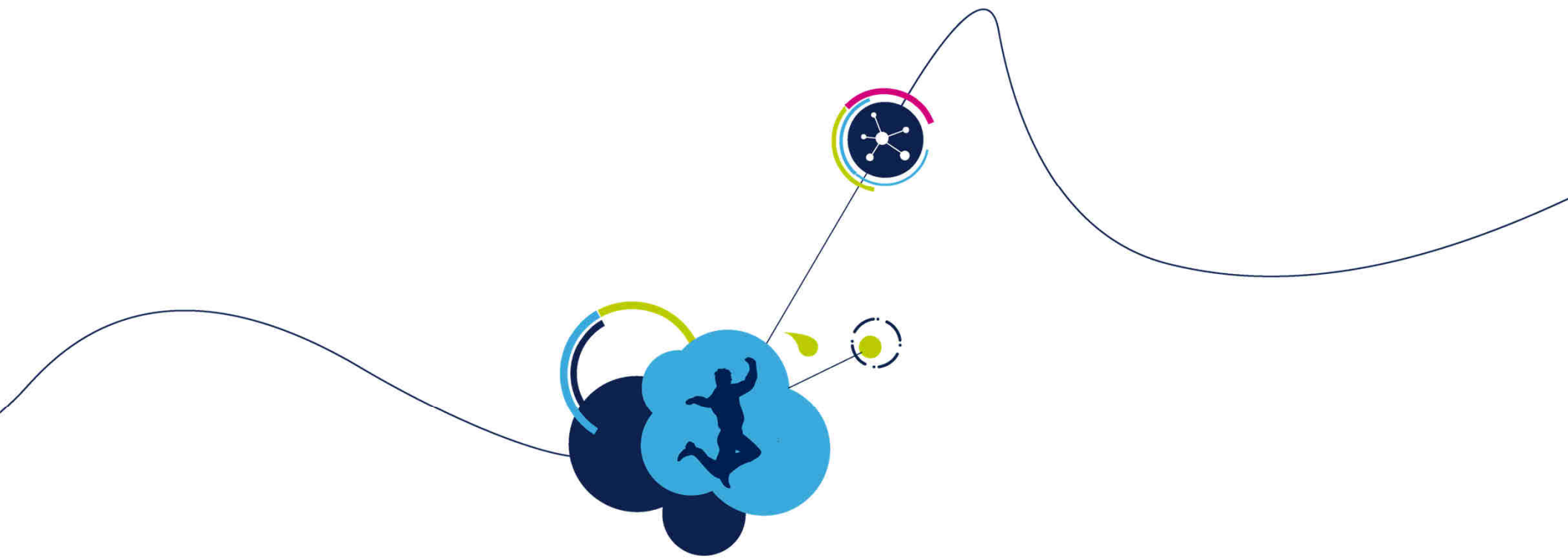
# Battery charger for HEV/EV

30



Single-phase architecture → SiC MOS 650V

Three-phase architecture → mainly SiC MOS 1200V

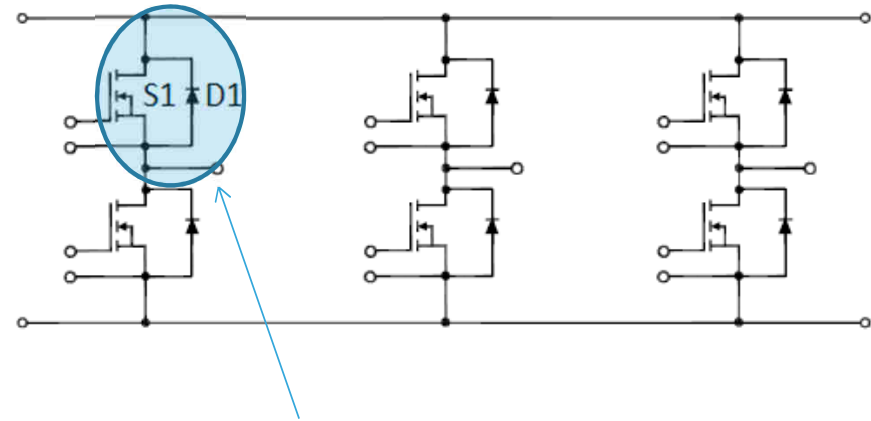


# 80kW EV Traction Inverter Power Loss Estimation:

1200V Gen 3 SiC MOSFETs vs 1200V Si IGBT+Diode

# Operating conditions

- Topology: Three phase inverter
- Bipolar PWM Strategy
- Synchronous rectification (SiC version)
- DC-link voltage:  $800V_{dc}$
- Current 250Arms (peak) 120Arms (nom)
- Switching frequency: 16kHz
- $V_{gs}=+18V/-5V$  for SiC,  $V_{ge}=\pm 15V$  for IGBT
- $\cos(\phi)$ : 0.8
- Modulation index (MI): 1
- Cooling fluid temperature:  $65^{\circ}C$
- $R_{thJ-C}(IGBT-die)=0.19^{\circ}C/W$ ;  
 $R_{thJ-C}(SiC-die)=0.30^{\circ}C/W$
- $T_j \leq 80\% \cdot T_{jmax}^{\circ}C$  at any condition



Si IGBT requires  
antiparallel diode, SiC  
MOSFET does not

Switch (S1+D1) implementation

4 x 1200V, 75A IGBTs + 4 x 1200V,75A Si diodes

vs.

3 x 1200V, 100A SiC MOSFETs SCT110G3D2AG



# Power loss at peak condition

(250A<sub>rms</sub>, 10sec)

\* Typical power loss values

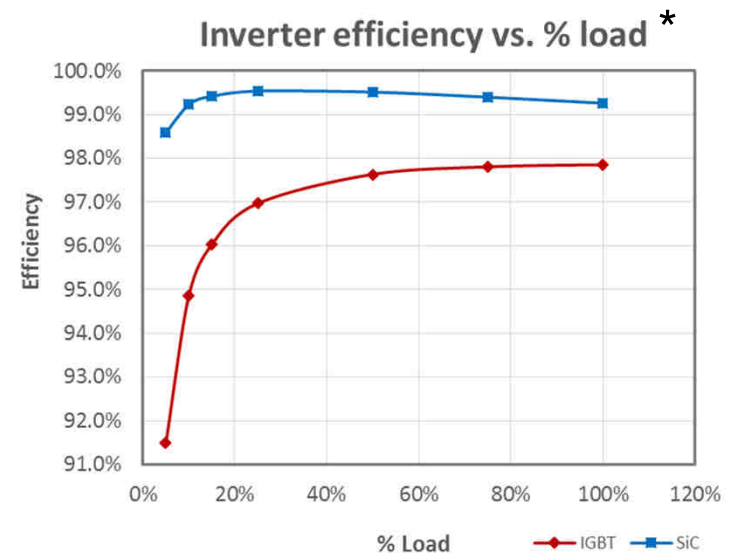
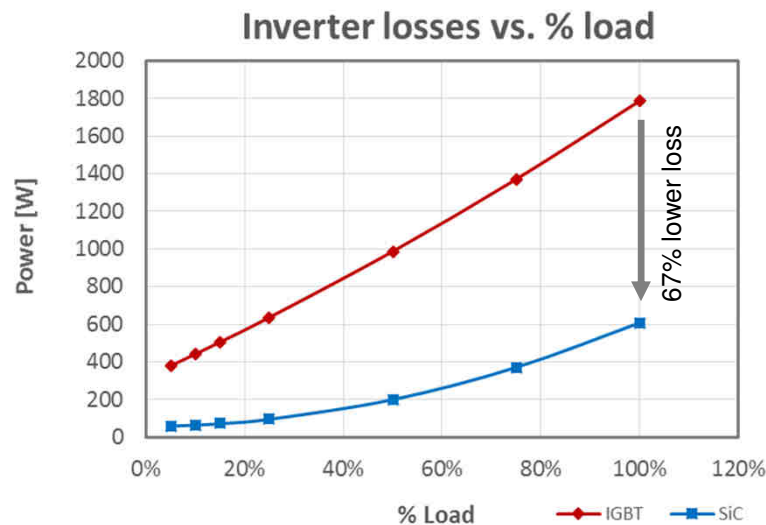
Loss Energy	Si-IGBT + Si-diode Solution	Full-SiC Solution	SiC vs Si per switch (S1+D1)
Total chip-area	180 mm <sup>2</sup> (IGBT) + 90 mm <sup>2</sup> (diode)	78 mm <sup>2</sup>	← 3.5x smaller area
Conduction losses* (W)	196.2	256.1	
Switching losses* (W)	316.6	94.0	← 3.4x lower
Diode's conduction losses* (W)	58.3	49.0	
Diode's Q <sub>rr</sub> losses* (W)	91.1	Negligible	
(S1+D1) Total losses* (W)	662.2	399.2	← 40% lower
Junction Temperature (°C)	134.2	151.5	← T <sub>J</sub> ~ 80% T <sub>jmax</sub>

SiC MOSFET runs at higher junction temperature in spite of lower losses. This is due to the exceptional SiC R<sub>DS(on)</sub> x Area FOM.

# SiC Solution: lower losses, higher efficiency

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$f_{sw}=16\text{kHz}$ , 100% load =  $120\text{A}_{rms}$



**SiC shows much lower loss over the whole load range**

**SiC offers 1.4% higher efficiency or more over the whole load range!**

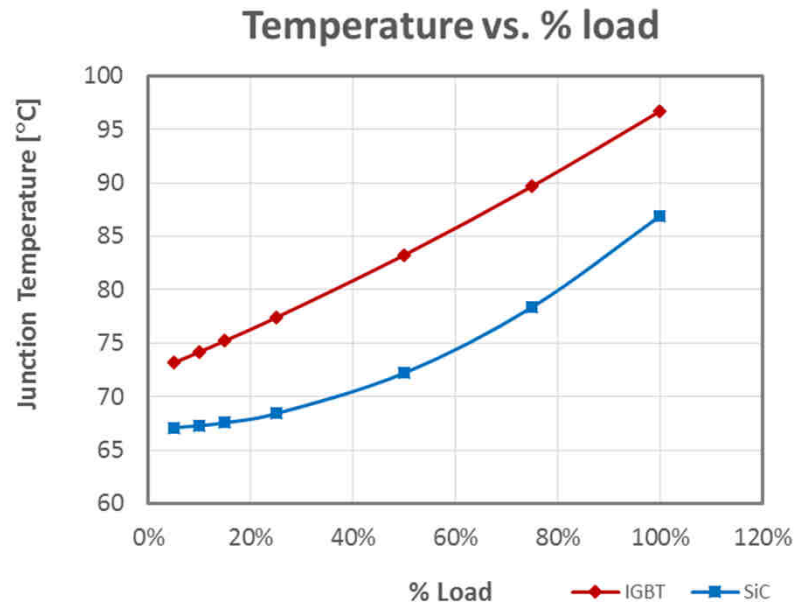
**Lower losses mean smaller cooling system and longer battery autonomy**

\* The simulated efficiency takes into account only the losses due to the switches and diodes forming the bridge inverter

# Remarks about junction temperature

35

$f_{sw}=16\text{kHz}$ , 100% load =  $120\text{A}_{rms}$



- $R_{thJ-C(IGBT-die)} = 0.19^{\circ}\text{C/W}$
- $R_{thJ-C(SiC-die)} = 0.30^{\circ}\text{C/W}$

- SiC solution is better than Silicon in reliability since SiC has lower  $\Delta(T_j - T_{fluid})$  up to 100% load.
- Cooling fluid temperature:  $65^{\circ}\text{C}$  for both SiC MOS and Si IGBT, this means the IGBT cooling system must be more efficient due to IGBT higher losses

# SiC MOSFET enables EV cost savings

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## Battery cost savings

EV traction inverter runs at ~ 15% load on average

At 15% load, SiC based inverter gives 3.4% efficiency improvement

Battery capacity required for SiC based EV is only 96.6% that of IGBT based → **SiC based EV needs only 82.1 kWh battery to give same range as IGBT based version with 85 kWh battery**

Typical battery cost: \$150 per kWh

**Battery cost savings with SiC based inverter (this example) : \$435**

## Heat sink considerations

Heat sink must be sized according to power dissipation at maximum operating condition

Can specify inverter heat sink assuming dissipation at 125% rated load (for added margin):

Switch Type	IGBT	SiC MOSFET
Inverter Power Dissipation	2235W	910W

SiC based inverter will only need to dissipate **41% as much heat** versus IGBT version

**SiC MOSFET allows smaller, lower cost heatsink**

# SiC MOSFET traction inverter

## Key advantages

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- More than 50% module/package size reduction
  - Much smaller semiconductor area giving ultra-compact solution
- >1.4% efficiency improvement and 67% lower loss:
  - Much lower loss at low load allows smaller battery for same range
- 60% cooling system downsize:
  - Lower losses at full load giving smaller cooling system
  - Lower  $\Delta T$  ( $T_j - T_{\text{fluid}}$ ) in the whole load range giving better reliability

# STGAP1AS: advanced galvanically isolated gate driver

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**AEC-Q100 grade 1**

**Wide operating range ( $T_A$  -40°C to +125°C)**

**5 A sink/source current**

**High Voltage Rail up to 1.5 kV**

Wide drive voltage range (+ 36 V / -10V)

**Short propagation delay**

100 ns typ.; 130 ns max over temperature

**Excellent CMTI rating**

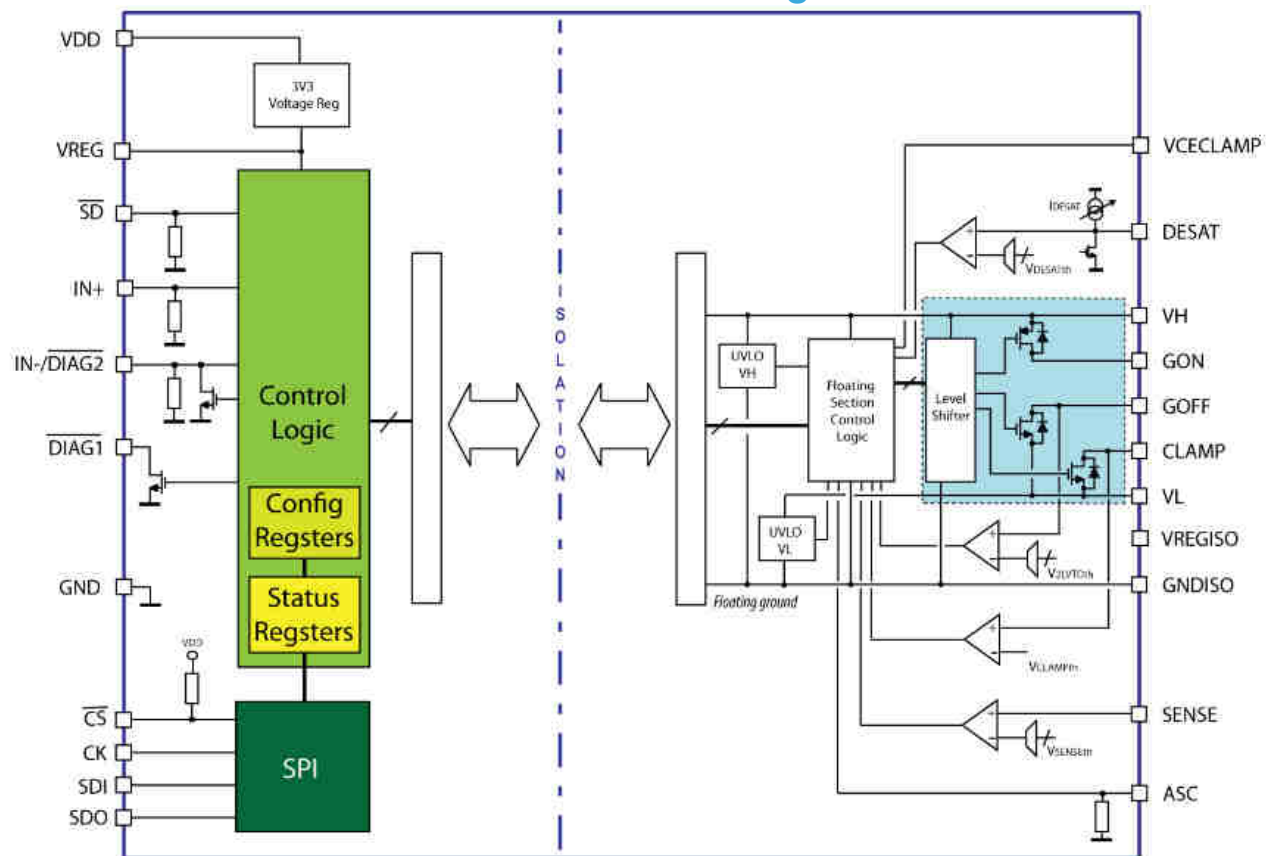
50 V/ns across full temperature range

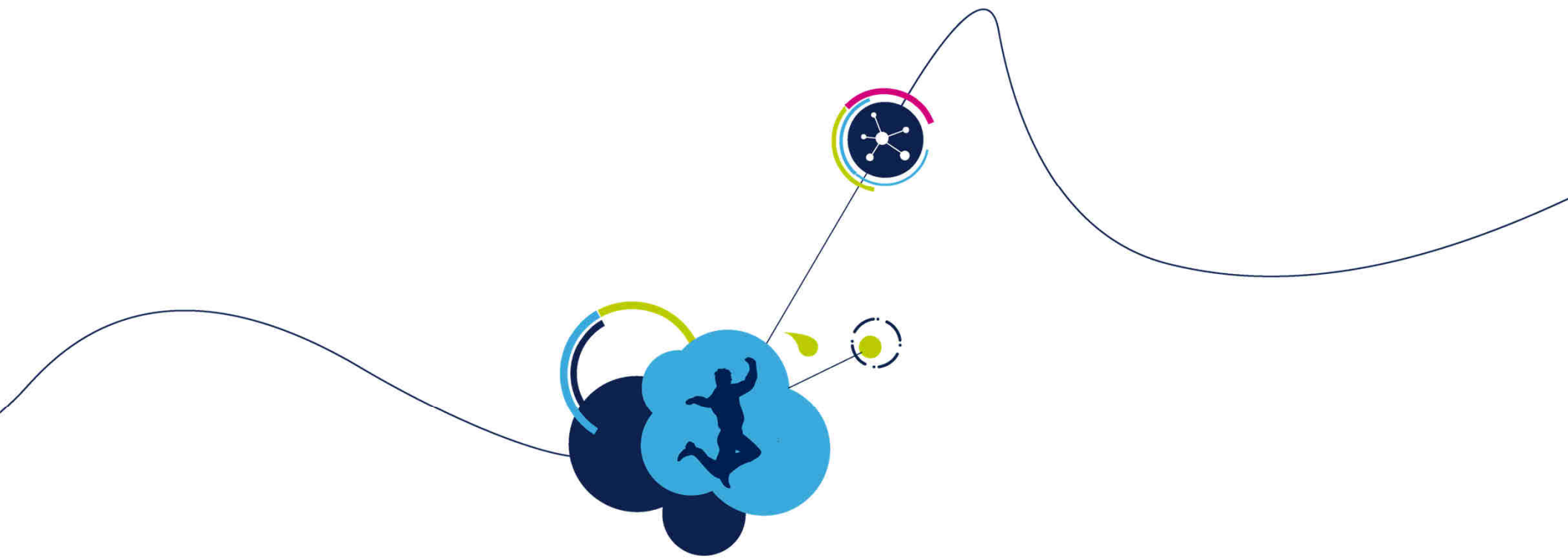
**Advanced features**

5A Active Miller clamp, Desaturation detection, 2-level turn-off, VCEClamp, ASC



**STGAP1AS Block Diagram**





# Conclusions

# Component cost considerations

SiC MOSFET vs IGBT (1200V)

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- Today
  - Price of SiC MOSFET is 4 – 4.5x relative to IGBT
- Near Term (2 – 3 years)
  - 2.5x vs IGBT, cost reduction from improvements in  $R_{DS(on)}$  x area FOM, and higher volume
- Long Term (5 – 10 years)
  - Further development and larger wafer diameter needed to continue to bring cost down

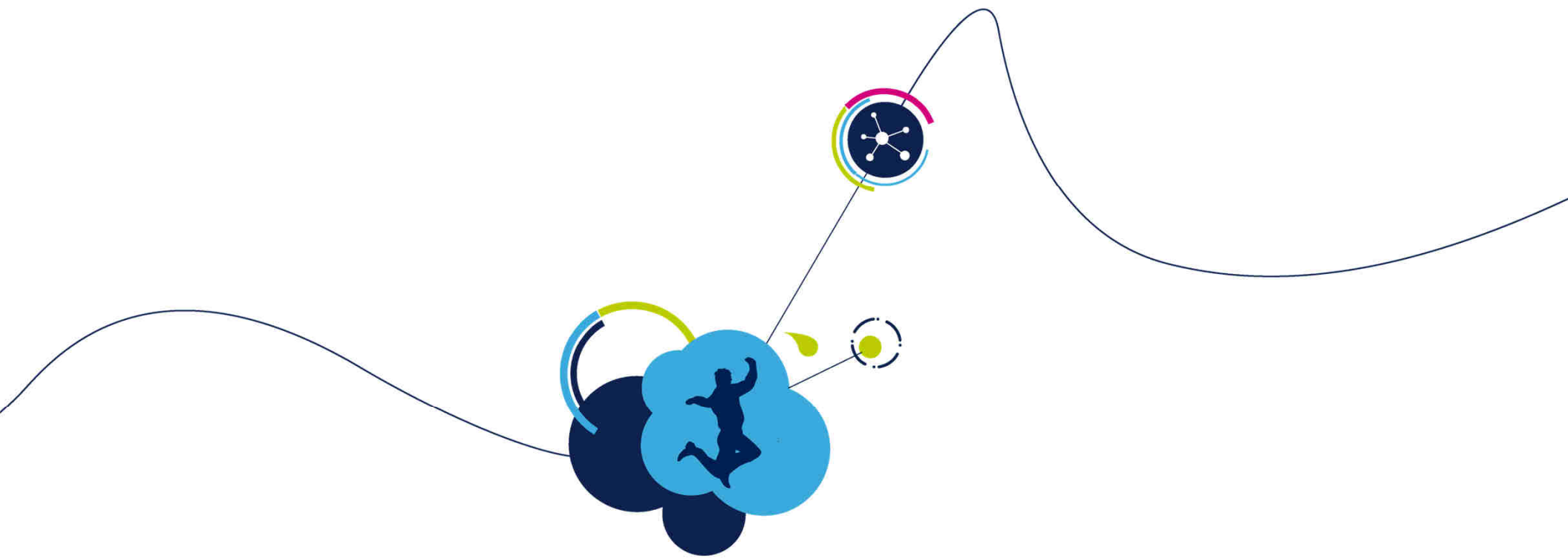


# SiC MOSFET price roadmap

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- Today SiC represents an attractive but still expensive solution for many applications
  - Valid when considering just the component cost
- For cost benefit of SiC to be recognized, must consider cost impact on whole system as SiC enables:
  - Smaller cooling system (saving space and weight)
  - Smaller footprint (more compact electronics)
  - Higher efficiency (less energy used)
  - Ability to use higher switching frequency (smaller passive components)
  - Higher reliability (smaller  $\Delta T$  across load range)

- SiC MOSFET-based power converters now offer system level benefits compared to silicon IGBT-based solutions
  - Traction inverter example shows how SiC can improve reliability and reduce system level cost
- SiC MOSFETs provide reduced footprint today compared to silicon based solutions and further footprint reductions are still possible
- Higher volume use and further innovation of SiC will continue to push down the cost and further displace silicon power transistors in the future



Thank You!



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