Introduction

As the demand for rechargeable batteries increases, so does the demand for battery chargers. There are different kinds of design solutions available for implementing battery chargers. Some of them are dedicated hardware based solutions and some are microcontroller based solutions.

In a microcontroller based solution, you have the flexibility of using the same hardware for charging different batteries and making only slight changes in the software. But there are still some challenges and one of the major challenges is to have a suitable input power supply available. Generally the Buck converter topology is used as a DC-DC converter to provide the controlled output power supply to the batteries. But in this case a problem may arise, for example, if you want to charge a 4.2V Li-ion batteries from a 5V supply due to the presence of the protection diode and other small drops across other components. This drop is generally about 1V which makes it very difficult to provide 4.2V to the Li-ion batteries using the buck converter topology.

This application note describes a simple technique for implementing a non-inverting buck-boost converter which requires only one inductor. This converter is basically the result of cascading a Buck converter with a Boost converter. This converter can be controlled by two PWM signals from the microcontroller and can be used as a Buck converter or Boost converter whenever required. So this solution combined with the flexibility of the ST7 microcontroller can be used to charge a wide range of the batteries using the same hardware.

The example used in this application note is specific to battery chargers but this DC-DC converter can be very useful for portable applications in general or any application which uses rechargeable batteries.
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1 Circuit diagram

The diagram in Figure 1 shows the structure of the modified buck-boost converter.

**Figure 1. Modified buck-boost converter**
2 Theory of operation

You can use this converter as buck-boost converter, as a buck converter or as a boost converter by selecting different combinations of switches SW1 and SW2 driven by the PWM1 and PWM2 signals output by the ST7 microcontroller.

2.1 Buck-boost implementation

This converter can be used as a non inverting buck-boost converter by selecting the operating mode from Table 1 which briefly describes the converter modes.

<table>
<thead>
<tr>
<th>Phase</th>
<th>SW1 (PWM1)</th>
<th>SW2 (PWM2)</th>
<th>Operating modes</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>OFF</td>
<td>OFF</td>
<td>BUCK</td>
</tr>
<tr>
<td>2</td>
<td>OFF</td>
<td>ON</td>
<td>N/A</td>
</tr>
<tr>
<td>3</td>
<td>ON</td>
<td>OFF</td>
<td>BUCK-BOOST</td>
</tr>
<tr>
<td>4</td>
<td>ON</td>
<td>ON</td>
<td>BOOST</td>
</tr>
</tbody>
</table>

Table 1. Operating modes based on switch combinations

If we look at phase 2 in Table 1, here the switch SW1(PWM1) is OFF and switch SW2 (PWM2) is ON. This condition never occurs either in a buck converter or in boost converter. So you should always take care in your software that this condition must not happen. To avoid this, if we assume that initially both switches are in OFF condition then you should use the following guidelines to manage the PWM signals driving the two switches.

1. Keep the frequency of both PWM signals the same, to better control when synchronizing the two PWM signals using the next three guidelines.
2. The duty cycle D1 of control signal PWM1, must be greater than the duty cycle D2 of control signal PWM2.
3. PWM1 should be enabled before the PWM2 signal.
4. PWM1 should be disabled after the PWM2 signal.
Figure 2 shows a timing example for the two PWM signals based on the above guidelines. Here phase 2 does not occur.

- If the duty cycles of the two PWM signals driving SW1 (PWM1) and SW2 (PWM2) are D1 and D2 respectively and
- if we exclude the saturation voltage of the switches from our calculation and
- if the drop across the diodes is $V_{d1}$ and $V_{d2}$ respectively,
- then the output voltage $V_{out}$ is given by the following formula:

$$V_{out} = \left[ V_{in} \cdot D1 - V_{d1} \cdot (1 - D1) \right] / (1 - D2) - V_{d2}$$

As mentioned in [2], theoretically this converter works linearly over a gain range of 0 - 200% of the input voltage.
2.2 Buck converter implementation

If you keep SW2 always in OFF condition and drive SW1 (PWM1) with a PWM signal from the microcontroller then the circuit works like a buck converter, except that you have an additional diode drop $V_{d2}$ due to diode d2.

Figure 3. Buck converter implementation

If the duty cycle of the PWM signal driving SW1 (PWM1) is $D_1$, the output voltage will be given by:

$$V_{out} = V_{in} \cdot D_1 - V_{d1} (1 - D_1) - V_{d2}$$
2.3 Boost converter implementation

Again if you keep switch SW1 (PWM1) always in ON condition and drive the Switch SW2 (PWM2) with a PWM signal generated by the microcontroller, then the combination works as a boost converter except that you have an additional free wheeling diode D1 which you can ignore.

Figure 4. Boost converter implementation

If the input voltage is $V_{in}$, the duty cycle of the PWM signal driving PWM2 is $D_2$ and the drop across the diode $d_2$ is $V_{d_2}$, then the output voltage $V_{out}$ is given by:

$$V_{out} = V_{in} / (1 - D_2) - V_{d_2}$$
3 Selection of components

The inductor and capacitor can be selected using the formulae given below.

3.1 Inductor selection

The minimum value of the inductor can be selected by choosing the maximum of the values given by the following two formulae:

\[ L_{\text{min}} = \frac{T \times [ (V_{\text{in}} - V_{\text{sat1}}) \times D1 - V_{\text{sat2}} \times D2 - V_{\text{out}} \times (D1 - D2)]}{2 \times I_{\text{out}}} \]  \hspace{1cm} (1)

\[ L_{\text{min}} = \frac{T \times [ V_{\text{d1}} + V_{\text{out}}] \times (1 - D1)}{2 \times I_{\text{out}}} \]  \hspace{1cm} (2)

Here \( V_{\text{sat1}} \) and \( V_{\text{sat2}} \) are the saturation voltages of the two switches SW1 and SW2.

\( I_{\text{out}} \) and \( V_{\text{out}} \) are the maximum output current and voltage respectively.

\( V_{\text{d1}} \) and \( V_{\text{d2}} \) is the drop across diodes d1 and d2.

The duty cycles of the PWM signals driving SW1 and SW2 are D1 and D2 respectively.

3.2 Capacitor selection

The minimum of the capacitor value can be selected by using the following formula, assuming a variation in \( V_{\text{out}} \) of 1% or less

\[ C_{\text{min}} = \frac{100 \times I_{\text{out}} \times (1 - D1) \times T}{V_{\text{out}}} \]  \hspace{1cm} (3)

In practice, we take inductor and capacitor values that are larger than the values calculated using the above formulae.
4 Application in battery charger

We can use the modified non-inverting buck-boost converter in a combination of different modes as required by the application.

4.1 Theory of operation

The DC-DC converter uses a combination of buck-boost converter and boost converter mode to charge the Li-ion battery. In case of Li-ion, the constant current constant voltage (CC CV) charging algorithm is used to charge the battery. Here we have chosen the input voltage just enough to show the functionality of the converter in buck-boost mode and boost mode.

Initially the converter works in Buck-Boost converter mode to charge the battery in constant current mode by keeping the duty cycle of PWM2 constant and varying the duty cycle of PWM1. As soon as there is an overflow condition for the duty cycle of PWM1. The converter switches from buck-boost converter mode to boost converter mode. And then duty cycle of PWM2 is varied to follow the algorithm while SW1 remains in ON condition. Following section shows the software flow chart for this combination.
4.2 Software flowchart

There are many ways in which you can control the operation of this circuit. An example algorithm which can be used for in Li-ion battery charger is given below-

Figure 5. Li-ion battery charger flowchart

1. **INITIALIZE THE PWM SIGNALS WITH DUTY CYCLE D1 AND D2.**

2. **WORKS IN BUCK-BOOST MODE. KEEP D2 FIXED AND VARY D1 TO KEEP THE CURRENT CONSTANT.**

3. **IS THERE A OVERFLOW IN D1?**
   - **No**
   - **YES**

4. **WORKS IN BOOST MODE. KEEP SW1 ALWAYS ON AND VARY THE D2 TO CONTINUE THE ALGORITHM BY KEEPING THE CURRENT OR VOLTAGE CONSTANT AS REQUIRED.**

5. **END**
We used this buck-boost converter in the universal battery charger evaluation board described in AN2390. Figure 6 below shows the general battery charger circuit using the non-inverting buck-boost converter circuit mentioned in this application note. For simplicity, this figure does not show all the connections.

Figure 6. General circuit-based battery charger

Here some results are shown for charging the Li-ion battery using this converter in buck-boost mode. The above charger is is intended for charging a single Li-ion or two NiMH in series using a 5V supply input. Details on the implementation are given in AN2390 where you can find results for NiMH batteries as well as for a charger used simply in buck converter mode.

Some parameter values are as follows:

$V_{in} = 5V$, $I_{in \text{ (MAX)}} = 2A$, $V_{sat1} = V_{sat2} = 0.3V$, $V_{d1} = V_{d2} = 0.5V$, and PWM frequency for both PWM1 and PWM2 = 16KHz.

Also let’s say the maximum value of $V_{out} = 5.5V$ and $I_{out} = 1.2A$ and the maximum duty cycle of D1 is 95% and D2 is equal to 30%.

Then using the formulas given in Section 3:

$L_{min} = 21 \mu H$ (from equation 1) or $10 \mu H$ (from equation 2). So we should choose a value larger than 21 \mu H.

$C_{min} = 70uF$ (from equation 3). So we need to choose a value higher than 70 uF.

In this example, $L = 75 \mu H$, $C = 470 \mu F$ are taken, which are larger than the values calculated using the formulas hence will support the application.

The following table shows some of the readings taken for different values of the duty cycles D1 and D2.

<table>
<thead>
<tr>
<th>PWM1</th>
<th>PWM2</th>
<th>VOUT</th>
<th>Cd1</th>
<th>P-</th>
<th>P+</th>
<th>L d2</th>
<th>VIN</th>
<th>SW2</th>
<th>SW1</th>
<th>PW</th>
<th>VB</th>
<th>IB</th>
<th>RS</th>
<th>Sense resistor</th>
<th>Battery</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>Lmin</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>21 \mu H (from equation 1) or 10 \mu H (from equation 2).</td>
<td>So we should choose a value larger than 21 \mu H.</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>10 \mu H (from equation 2).</td>
<td>So we should choose a value larger than 21 \mu H.</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
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<td></td>
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<tr>
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<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>$C_{min} = 70uF$ (from equation 3).</td>
<td>So we need to choose a value higher than 70 uF.</td>
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<td></td>
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<td></td>
<td></td>
<td></td>
<td>In this example, $L = 75 \mu H$, $C = 470 \mu F$ are taken, which are larger than the values calculated using the formulas hence will support the application.</td>
<td>The following table shows some of the readings taken for different values of the duty cycles D1 and D2.</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>
In above results, we have not added $V_{sat}$ (0.3V), diode drop (0.5V) and the drop across the series resistor (connected to measure battery current) as shown in Figure 6 to get the actual output voltage of the non-inverting buck-boost converter.

For example it is given that $V_{out}$ (battery voltage) = 4.2V,

but the actual output voltage of the converter is equal to:

$$V_{out} = 4.2V \text{ (Battery Voltage } V_{bat}) + 0.3V \text{(Switch Drop } V_{sat1}) + 0.5V \text{(Protection Diode } V_{d2}) + 0.4V \text{ (Drop across Sense resistor)} = 5.4V.$$ 

So we are able to achieve 5.4V from a 5V supply thus validating the concept.

<table>
<thead>
<tr>
<th>SI No</th>
<th>$V_{in}$ (V)</th>
<th>$I_{in}$ (A)</th>
<th>$V_{out}$ (V)</th>
<th>$I_{out}$ (A)</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>5</td>
<td>1.23</td>
<td>3.9</td>
<td>1.05</td>
</tr>
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<td>4</td>
<td>5</td>
<td>1.27</td>
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<td>5</td>
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<td>4.2</td>
<td>0.80</td>
</tr>
<tr>
<td>6</td>
<td>5</td>
<td>0.85</td>
<td>4.2</td>
<td>0.712</td>
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<tr>
<td>7</td>
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<td>0.65</td>
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<td>5</td>
<td>0.33</td>
<td>4.15</td>
<td>0.28</td>
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<tr>
<td>9</td>
<td>5</td>
<td>0.31</td>
<td>4.15</td>
<td>0.26</td>
</tr>
<tr>
<td>10</td>
<td>5</td>
<td>0.16</td>
<td>4.10</td>
<td>0.129</td>
</tr>
</tbody>
</table>
6 Conclusion

This application note describes a simple but very useful technique for implementing a microcontroller controlled low cost non-inverting buck-boost converter. This converter can be used as a buck converter or as a boost converter or as a buck-boost converter. The example shows the application of this converter in a battery charger. However it can also be used in other portable applications or any application getting its power from a rechargeable battery. Or you could use this technique to make a USB charger to charge Li-ion batteries or 3 or more NiMH Cells in series.
7 References


Revision history

Table 3. Document revision history

<table>
<thead>
<tr>
<th>Date</th>
<th>Revision</th>
<th>Changes</th>
</tr>
</thead>
<tbody>
<tr>
<td>21-Aug-2007</td>
<td>1</td>
<td>Initial release.</td>
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