High bandwidth (22 MHz) low offset (200 µV) 5 V op amp

Features

• Gain bandwidth product 22 MHz, unity gain stable
• High accuracy input offset voltage: 50 µV typ., 200 µV max.
• Low input bias current: 2 pA typ.
• Low input voltage noise density: 7 nV/√Hz
• Wide supply voltage range: 1.8 V to 5.5 V
• Output rail-to-rail
• Input common-mode range includes low rail
• Automotive grade and shutdown versions available

Benefits:

– High frequency signal conditioning
– Optimized accuracy for low-side current sensing

Applications

• Low-side current measurement
• Photodiode amplifiers
• Automotive current measurement and sensor signal conditioning
• Strain gauges signal conditioning

Description

The TSV7721, TSV7722 and TSV7723 are single and dual 22 MHz-bandwidth unity-gain-stable amplifiers. The input offset voltage of 200 µV max. (50 µV typ.) at room temperature, optimized for common-mode close to ground makes the TSV772x ideal for low-side current measurements.

The TSV772x can operate from 1.8 V to 5.5 V single supply and it is fully specified on a load of 47 pF, therefore allowing easy usage as A/D converters input buffer.

The TSV772x series offers rail-to-rail output, excellent speed/power consumption ratio, and 22 MHz gain bandwidth product, while consuming just 1.7 mA at 5 V.

The devices also feature an ultra-low input bias current that enables connection to photodiodes and other sensors where current is the key value to be measured.

These features make the TSV772x series ideal for high-accuracy, high-bandwidth sensor interfaces.

Related products

<table>
<thead>
<tr>
<th>TSV792</th>
<th>Rail-to-rail amplifier with higher GBW 50 MHz</th>
</tr>
</thead>
<tbody>
<tr>
<td>TSB7192</td>
<td>22 MHz amplifier with 36 V supply voltage</td>
</tr>
</tbody>
</table>
1 Pin description

1.1 TSV7721 single operational amplifier

Figure 1. Pin connections (top view)

![Pin connection diagram](image)

Table 1. Pin description

<table>
<thead>
<tr>
<th>Pin n°</th>
<th>Pin name</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>OUT</td>
<td>Output channel</td>
</tr>
<tr>
<td>2</td>
<td>VCC-</td>
<td>Negative supply voltage</td>
</tr>
<tr>
<td>3</td>
<td>IN+</td>
<td>Non-inverting input channel</td>
</tr>
<tr>
<td>4</td>
<td>IN-</td>
<td>Inverting input channel</td>
</tr>
<tr>
<td>5</td>
<td>VCC+</td>
<td>Positive supply voltage</td>
</tr>
</tbody>
</table>
1.2 TSV7722 dual operational amplifier

Figure 2. Pin connections (top view)

1. The exposed pad of the DFN8 2x2 can be connected to VCC- or left floating.

Table 2. Pin description

<table>
<thead>
<tr>
<th>Pin n°</th>
<th>Pin name</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>OUT1</td>
<td>Output channel 1</td>
</tr>
<tr>
<td>2</td>
<td>IN1-</td>
<td>Inverting input channel 1</td>
</tr>
<tr>
<td>3</td>
<td>IN1+</td>
<td>Non-inverting input channel 1</td>
</tr>
<tr>
<td>4</td>
<td>VCC-</td>
<td>Negative supply voltage</td>
</tr>
<tr>
<td>5</td>
<td>IN2+</td>
<td>Non-inverting input channel 2</td>
</tr>
<tr>
<td>6</td>
<td>IN2-</td>
<td>Inverting input channel 2</td>
</tr>
<tr>
<td>7</td>
<td>OUT2</td>
<td>Output channel 2</td>
</tr>
<tr>
<td>8</td>
<td>VCC+</td>
<td>Positive supply voltage</td>
</tr>
</tbody>
</table>
1.3 **TSV7723 dual operational amplifier with shutdown option**

Figure 3. Pin connections (top view)

<table>
<thead>
<tr>
<th>Pin n°</th>
<th>Pin name</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>OUT1</td>
<td>Output channel 1</td>
</tr>
<tr>
<td>2</td>
<td>IN1-</td>
<td>Inverting input channel 1</td>
</tr>
<tr>
<td>3</td>
<td>IN1+</td>
<td>Non-inverting input channel 1</td>
</tr>
<tr>
<td>4</td>
<td>VCC-</td>
<td>Negative supply voltage</td>
</tr>
</tbody>
</table>
| 5      | EN1      | Enable input channel 1  
            *(amplifier in shutdown mode when EN pin connected to VCC-)* |
| 6      | EN2      | Enable input channel 2  
            *(amplifier in shutdown mode when EN pin connected to VCC-)* |
| 7      | IN2+     | Non-inverting input channel 2 |
| 8      | IN2-     | Inverting input channel 2 |
| 9      | OUT2     | Output channel 2 |
| 10     | VCC+     | Positive supply voltage |
## Absolute maximum ratings and operating conditions

### Table 4. Absolute maximum ratings

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Value</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>V&lt;sub&gt;C&lt;/sub&gt;C</td>
<td>Supply voltage (referred to VCC- pin)&lt;sup&gt;(1)&lt;/sup&gt;</td>
<td>-0.3 to 6.0</td>
<td>V</td>
</tr>
<tr>
<td>V&lt;sub&gt;id&lt;/sub&gt;</td>
<td>Differential input voltage&lt;sup&gt;(2)&lt;/sup&gt;</td>
<td>± VCC</td>
<td>V</td>
</tr>
<tr>
<td>V&lt;sub&gt;IN&lt;/sub&gt;</td>
<td>Input pins input voltage&lt;sup&gt;(3)&lt;/sup&gt;</td>
<td>V&lt;sub&gt;CC&lt;/sub&gt; - 0.3 V to V&lt;sub&gt;CC&lt;/sub&gt; + 0.3 V</td>
<td>V</td>
</tr>
<tr>
<td>I&lt;sub&gt;IN&lt;/sub&gt;</td>
<td>Input pins input current&lt;sup&gt;(4)&lt;/sup&gt;</td>
<td>± 10</td>
<td>mA</td>
</tr>
<tr>
<td>T&lt;sub&gt;stg&lt;/sub&gt;</td>
<td>Storage temperature</td>
<td>-65 to 150</td>
<td>°C</td>
</tr>
<tr>
<td>R&lt;sub&gt;th-ja&lt;/sub&gt;</td>
<td>Thermal resistance junction-to-ambient&lt;sup&gt;(5)&lt;/sup&gt;</td>
<td>SOT23-5: 250</td>
<td>°C / W</td>
</tr>
<tr>
<td></td>
<td></td>
<td>DFN8 (2 mm x 2 mm): 76</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>MiniSO8: 127</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>MiniSO10: 113</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>SO8: 113</td>
<td></td>
</tr>
<tr>
<td>T&lt;sub&gt;j&lt;/sub&gt;</td>
<td>Maximum junction temperature</td>
<td>150</td>
<td>°C</td>
</tr>
<tr>
<td>ESD</td>
<td>HBM: human body model&lt;sup&gt;(6)&lt;/sup&gt;</td>
<td>4</td>
<td>kV</td>
</tr>
<tr>
<td></td>
<td>CDM: charged device model&lt;sup&gt;(7)&lt;/sup&gt;</td>
<td>1.5</td>
<td>kV</td>
</tr>
</tbody>
</table>

1. All voltage values, except differential voltage, are with respect to VCC- pin.
2. The differential voltage is the non-inverting input terminal with respect to the inverting input terminal.
3. V<sub>CC</sub> - Vin must not exceed 6 V, Vin must not exceed 6 V.
4. Input current must be limited by a resistor in series with the inputs.
5. R<sub>th</sub> are typical values.
6. Human body model: the test HBM is done in accordance with the standards ESDA-JS-001-2017 and Q100-002
7. Charged device model: the test CDM is done in accordance with the standards ESDA-JS-002-2018 and Q100-011

### Table 5. Operating conditions

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Min.</th>
<th>Max.</th>
<th>Value</th>
</tr>
</thead>
<tbody>
<tr>
<td>V&lt;sub&gt;C&lt;/sub&gt;C</td>
<td>Supply voltage</td>
<td>1.8</td>
<td>5.5</td>
<td>V</td>
</tr>
<tr>
<td>V&lt;sub&gt;cm&lt;/sub&gt;</td>
<td>Common-mode input voltage range</td>
<td>V&lt;sub&gt;CC&lt;/sub&gt; - 0.1</td>
<td>V&lt;sub&gt;CC&lt;/sub&gt; + 1.1</td>
<td>V</td>
</tr>
<tr>
<td>T&lt;sub&gt;oper&lt;/sub&gt;</td>
<td>Operating free air temperature range</td>
<td>-40</td>
<td>125</td>
<td>°C</td>
</tr>
</tbody>
</table>
### Electrical characteristics

Table 6. Electrical characteristics at \( V_{CC^+} = 5.0 \) V, with \( V_{CC} = 0 \) V, \( V_{icm} = V_{CC} / 2 \), \( T = 25^\circ\text{C} \), and OUT pin connected to \( V_{CC} / 2 \) through \( R_L = 10 \) kΩ (unless otherwise specified)

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Conditions</th>
<th>Min.</th>
<th>Typ.</th>
<th>Max.</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>( V_{io} )</td>
<td>Input offset voltage (( V_{icm} = 0 ) V)</td>
<td>( T = 25^\circ\text{C} )</td>
<td>±50</td>
<td>±250</td>
<td>µV</td>
<td></td>
</tr>
<tr>
<td>( \Delta V_{io}/\Delta T )</td>
<td>Input offset voltage drift (( V_{icm} = 0 ) V)</td>
<td>-40°C &lt; ( T &lt; 125^\circ\text{C} )</td>
<td>±4</td>
<td>µV/°C</td>
<td></td>
<td></td>
</tr>
<tr>
<td>( I_{ib} )</td>
<td>Input bias current (( V_{OUT} = V_{CC}/2 )) (^{(1)})</td>
<td>( T = 25^\circ\text{C} )</td>
<td>1</td>
<td>2</td>
<td>pA</td>
<td></td>
</tr>
<tr>
<td>( I_{io} )</td>
<td>Input offset current (( V_{OUT} = V_{CC}/2 )) (^{(1)})</td>
<td>( T = 25^\circ\text{C} )</td>
<td>1</td>
<td>2</td>
<td>pA</td>
<td></td>
</tr>
<tr>
<td>CMR1</td>
<td>Common-mode rejection ratio ( 20\log(\Delta V_{icm}/\Delta V_{io}) ), ( V_{icm} = 0 ) V to ( V_{CC} ), ( 1.1 ) V, ( R_L &gt; 1 ) MΩ</td>
<td>( T = 25^\circ\text{C} )</td>
<td>76</td>
<td>99</td>
<td>dB</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>-40°C &lt; ( T &lt; 125^\circ\text{C} )</td>
<td>74</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>CMR2</td>
<td>Common-mode rejection ratio ( 20\log(\Delta V_{icm}/\Delta V_{io}) ), ( V_{icm} = -0.1 ) V to ( V_{CC} ), ( 1.1 ) V, ( R_L &gt; 1 ) MΩ</td>
<td>( T = 25^\circ\text{C} )</td>
<td>75</td>
<td></td>
<td>dB</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>-40°C &lt; ( T &lt; 125^\circ\text{C} )</td>
<td>60</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>SVR</td>
<td>Supply voltage rejection ratio ( 20\log(\Delta V_{CC}/\Delta V_{io}) ), ( V_{CC} ) = 1.8 V to 5.5 V, ( V_{icm} = 0 ) V, ( R_L &gt; 1 ) MΩ</td>
<td>( T = 25^\circ\text{C} )</td>
<td>85</td>
<td>108</td>
<td>dB</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>-40°C &lt; ( T &lt; 125^\circ\text{C} )</td>
<td>80</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>( A_{VD} )</td>
<td>Large signal voltage gain ( V_{OUT} = 0.3 ) V to ( V_{CC} ), ( 0.3 ) V</td>
<td>( T = 25^\circ\text{C} )</td>
<td>111</td>
<td>130</td>
<td>dB</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>-40°C &lt; ( T &lt; 125^\circ\text{C} )</td>
<td>106</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>( V_{OH} )</td>
<td>High level output voltage ( (V_{OH} = V_{CC} - V_{OUT}) )</td>
<td>( T = 25^\circ\text{C} )</td>
<td>15</td>
<td></td>
<td>mV</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>-40°C &lt; ( T &lt; 125^\circ\text{C} )</td>
<td>25</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>( V_{OL} )</td>
<td>Low level output voltage</td>
<td>( T = 25^\circ\text{C} )</td>
<td>15</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>-40°C &lt; ( T &lt; 125^\circ\text{C} )</td>
<td>25</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>( I_{OUT} )</td>
<td>( I_{sink} (V_{OUT} = V_{CC}) )</td>
<td>( T = 25^\circ\text{C} )</td>
<td>50</td>
<td>70</td>
<td>mA</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>-40°C &lt; ( T &lt; 125^\circ\text{C} )</td>
<td>45</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>( I_{source} (V_{OUT} = 0 ) V)</td>
<td>( T = 25^\circ\text{C} )</td>
<td>45</td>
<td>65</td>
<td>mA</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>-40°C &lt; ( T &lt; 125^\circ\text{C} )</td>
<td>40</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>( I_{CC} )</td>
<td>Supply current (per channel, ( V_{OUT} = V_{CC}/2 ), ( R_L &gt; 1 ) MΩ)</td>
<td>( T = 25^\circ\text{C} )</td>
<td>1.7</td>
<td>2.2</td>
<td>mA</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>-40°C &lt; ( T &lt; 125^\circ\text{C} )</td>
<td>2.5</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

### AC Performance

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Conditions</th>
<th>Min.</th>
<th>Typ.</th>
<th>Max.</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>GBW</td>
<td>Gain bandwidth product</td>
<td>( C_L = 47 ) pF</td>
<td>15</td>
<td>22</td>
<td>MHz</td>
<td></td>
</tr>
<tr>
<td>( F_u )</td>
<td>Unity gain frequency</td>
<td></td>
<td>19.5</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>( \Phi_m )</td>
<td>Phase margin</td>
<td></td>
<td>44</td>
<td>degrees</td>
<td></td>
<td></td>
</tr>
<tr>
<td>( G_m )</td>
<td>Gain margin</td>
<td></td>
<td>8</td>
<td></td>
<td>dB</td>
<td></td>
</tr>
<tr>
<td>SR</td>
<td>Slew rate (^{(2)})</td>
<td></td>
<td>8</td>
<td>11</td>
<td>V/µs</td>
<td></td>
</tr>
</tbody>
</table>

---

\(^{(1)}\) Unless otherwise specified.

\(^{(2)}\) See Section 2 for details.
### Table 7. Electrical characteristics in shutdown mode at $V_{CC+} = 5.0$ V, with $V_{CC-} = 0$ V, $V_{icm} = V_{CC} / 2$, $T = 25^\circ C$, and OUT pin connected to $V_{CC} / 2$ through $R_L = 10 \text{k}\Omega$ (unless otherwise specified)

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Conditions</th>
<th>Min.</th>
<th>Typ.</th>
<th>Max.</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>$I_{CC}$</td>
<td>Shutdown mode $V_{OUT} = V_{CC}/2$, $R_L &gt; 1 \text{ M}\Omega$ (all channels)</td>
<td>$T = 25^\circ C$</td>
<td>2.5</td>
<td>60</td>
<td>nA</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40^\circ C &lt; T &lt; 85^\circ C$</td>
<td></td>
<td>450</td>
<td></td>
<td>µA</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40^\circ C &lt; T &lt; 125^\circ C$</td>
<td></td>
<td>4</td>
<td></td>
<td>µA</td>
</tr>
<tr>
<td>$t_{on}$</td>
<td>Amplifier turn-on time (other channel already on)</td>
<td>$V_{OUT} = V_{CC-} \text{ to } V_{CC-} + 0.2$ V</td>
<td>2</td>
<td></td>
<td></td>
<td>µs</td>
</tr>
<tr>
<td>$t_{init}$</td>
<td>Initialization time (both channels off)</td>
<td>$V_{OUT}$ to 200 mV of final value</td>
<td>7</td>
<td></td>
<td></td>
<td>µs</td>
</tr>
<tr>
<td>$V_{IH}$</td>
<td>EN logic high</td>
<td></td>
<td>2</td>
<td></td>
<td></td>
<td>V</td>
</tr>
<tr>
<td>$V_{IL}$</td>
<td>EN logic low</td>
<td></td>
<td>0.8</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>$I_{IH}$</td>
<td>EN current high</td>
<td>$EN = V_{CC+}$</td>
<td>1</td>
<td></td>
<td></td>
<td>pA</td>
</tr>
<tr>
<td>$I_{IL}$</td>
<td>EN current low</td>
<td>$EN = V_{CC-}$</td>
<td>1</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>$I_{Oleak}$</td>
<td>Output leakage in shutdown mode, $EN = V_{CC-}$</td>
<td>$T = 25^\circ C$</td>
<td>50</td>
<td></td>
<td></td>
<td>pA</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40^\circ C &lt; T &lt; 125^\circ C$</td>
<td>15</td>
<td></td>
<td></td>
<td>nA</td>
</tr>
</tbody>
</table>

1. Guaranteed by design and characterization on a sample of parts, not tested in production.
2. Slew rate value is calculated as the average between positive and negative slew rates.
Table 8. Electrical characteristics at $V_{CC^+} = 3.3$ V, with $V_{CC^-} = 0$ V, $V_{icm} = V_{CC} / 2$, $T = 25^\circ$C, and OUT pin connected to $V_{CC} / 2$ through $R_L = 10$ kΩ (unless otherwise specified)

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Conditions</th>
<th>Min.</th>
<th>Typ.</th>
<th>Max.</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>$V_{io}$</td>
<td>Input offset voltage ($V_{icm} = 0$ V)</td>
<td>$T = 25^\circ$C, $-40^\circ$C &lt; $T &lt; 125^\circ$C</td>
<td>±50</td>
<td>±200</td>
<td>µV</td>
<td></td>
</tr>
<tr>
<td>$\Delta V_{io}/\Delta T$</td>
<td>Input offset voltage drift ($V_{icm} = 0$ V)</td>
<td>$-40^\circ$C &lt; $T &lt; 125^\circ$C</td>
<td>±4</td>
<td></td>
<td>µV/^\circ$C</td>
<td></td>
</tr>
<tr>
<td>$I_{ib}$</td>
<td>Input bias current ($V_{OUT} = V_{CC}/2$)</td>
<td>$T = 25^\circ$C, $-40^\circ$C &lt; $T &lt; 85^\circ$C</td>
<td>1</td>
<td>2</td>
<td>pA</td>
<td></td>
</tr>
<tr>
<td>$I_{io}$</td>
<td>Input offset current ($V_{OUT} = V_{CC}/2$)</td>
<td>$T = 25^\circ$C, $-40^\circ$C &lt; $T &lt; 85^\circ$C</td>
<td>1</td>
<td>2</td>
<td>pA</td>
<td></td>
</tr>
<tr>
<td>CMR1</td>
<td>Common-mode rejection ratio $20 \log (\Delta V_{icm}/\Delta V_{io})$, $V_{icm} = 0$ V to $V_{CC}$, $R_L &gt; 1$ MΩ</td>
<td>$T = 25^\circ$C, $-40^\circ$C &lt; $T &lt; 125^\circ$C</td>
<td>75</td>
<td>96</td>
<td>dB</td>
<td></td>
</tr>
<tr>
<td>CMR2</td>
<td>Common-mode rejection ratio $20 \log (\Delta V_{icm}/\Delta V_{io})$, $V_{icm} = -0.1$ V to $V_{CC}$, $R_L &gt; 1$ MΩ</td>
<td>$T = 25^\circ$C, $-40^\circ$C &lt; $T &lt; 125^\circ$C</td>
<td>71</td>
<td></td>
<td>dB</td>
<td></td>
</tr>
<tr>
<td>$A_{VD}$</td>
<td>Large signal voltage gain $V_{OUT} = 0.3$ V to $V_{CC}$, $0.3$ V)</td>
<td>$T = 25^\circ$C, $-40^\circ$C &lt; $T &lt; 125^\circ$C</td>
<td>107</td>
<td>128</td>
<td>dB</td>
<td></td>
</tr>
<tr>
<td>$V_{OH}$</td>
<td>High level output voltage ($V_{OH} = V_{CC} - V_{OUT}$)</td>
<td>$T = 25^\circ$C, $-40^\circ$C &lt; $T &lt; 125^\circ$C</td>
<td>15</td>
<td></td>
<td>mV</td>
<td></td>
</tr>
<tr>
<td>$V_{OL}$</td>
<td>Low level output voltage</td>
<td>$T = 25^\circ$C, $-40^\circ$C &lt; $T &lt; 125^\circ$C</td>
<td>15</td>
<td></td>
<td>mV</td>
<td></td>
</tr>
<tr>
<td>$I_{OUT}$</td>
<td>$I_{sink}$ ($V_{OUT} = V_{CC}$)</td>
<td>$T = 25^\circ$C, $-40^\circ$C &lt; $T &lt; 125^\circ$C</td>
<td>50</td>
<td>70</td>
<td>mA</td>
<td></td>
</tr>
<tr>
<td>$I_{source}$ ($V_{OUT} = 0$ V)</td>
<td>$T = 25^\circ$C, $-40^\circ$C &lt; $T &lt; 125^\circ$C</td>
<td>45</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>$I_{CC}$</td>
<td>Supply current (per channel, $V_{OUT} = V_{CC}/2$, $R_L &gt; 1$ MΩ)</td>
<td>$T = 25^\circ$C, $-40^\circ$C &lt; $T &lt; 125^\circ$C</td>
<td>1.7</td>
<td>2.2</td>
<td>mA</td>
<td></td>
</tr>
</tbody>
</table>

**AC Performance**

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Units</th>
</tr>
</thead>
<tbody>
<tr>
<td>GBW</td>
<td>Gain bandwidth product</td>
<td>MHz</td>
</tr>
<tr>
<td>$F_u$</td>
<td>Unity gain frequency</td>
<td>14</td>
</tr>
<tr>
<td>$\Phi_m$</td>
<td>Phase margin</td>
<td>18.5</td>
</tr>
<tr>
<td>$G_m$</td>
<td>Gain margin</td>
<td>42</td>
</tr>
<tr>
<td>SR</td>
<td>Slew rate (2)</td>
<td>7.7</td>
</tr>
<tr>
<td>$t_s$</td>
<td>Settling time</td>
<td>210</td>
</tr>
<tr>
<td>$e_n$</td>
<td>Equivalent input noise voltage</td>
<td>13</td>
</tr>
<tr>
<td>$C_S$</td>
<td>Channel separation (for TSV7722 and TSV7723)</td>
<td>120</td>
</tr>
</tbody>
</table>

1. Guaranteed by design and characterization on a sample of parts, not tested in production.
2. Slew rate value is calculated as the average between positive and negative slew rates.
### Table 9. Electrical characteristics in Shutdown mode at $V_{CC^+} = 3.3$ V, with $V_{CC^-} = 0$ V, $V_{icm} = V_{CC}/2$, $T = 25^\circ$C, and OUT pin connected to $V_{CC}/2$ through $R_L = 10$ kΩ (unless otherwise specified)

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Conditions</th>
<th>Min.</th>
<th>Typ.</th>
<th>Max.</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>$I_{CC}$</td>
<td>Shutdown mode $V_{OUT} = V_{CC}/2$, $R_L &gt; 1$ MΩ (all channels)</td>
<td>$T = 25^\circ$C</td>
<td>2.5</td>
<td>60</td>
<td>nA</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40^\circ$C &lt; $T$ &lt; 85°C</td>
<td></td>
<td>450</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40^\circ$C &lt; $T$ &lt; 125°C</td>
<td></td>
<td>4</td>
<td>µA</td>
<td></td>
</tr>
<tr>
<td>$t_{on}$</td>
<td>Amplifier turn-on time (other channel already on)</td>
<td>$V_{OUT} = V_{CC^+}$ to $V_{CC^-} + 0.2$ V</td>
<td>2</td>
<td></td>
<td>µs</td>
<td></td>
</tr>
<tr>
<td>$t_{init}$</td>
<td>Initialization time (both channels off)</td>
<td>$V_{OUT}$ to 200 mV of final value</td>
<td>11</td>
<td></td>
<td>µs</td>
<td></td>
</tr>
<tr>
<td>$V_{IH}$</td>
<td>EN logic high</td>
<td></td>
<td>2</td>
<td></td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>$V_{IL}$</td>
<td>EN logic low</td>
<td></td>
<td></td>
<td>0.8</td>
<td></td>
<td></td>
</tr>
<tr>
<td>$I_{IH}$</td>
<td>EN current high</td>
<td>$EN = V_{CC^+}$</td>
<td>1</td>
<td></td>
<td>pA</td>
<td></td>
</tr>
<tr>
<td>$I_{IL}$</td>
<td>EN current low</td>
<td>$EN = V_{CC^-}$</td>
<td>1</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>$I_{Oleak}$</td>
<td>Output leakage in shutdown mode, $EN = V_{CC^-}$</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>$T = 25^\circ$C</td>
<td>50</td>
<td></td>
<td>pA</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40^\circ$C &lt; $T$ &lt; 125°C</td>
<td>15</td>
<td></td>
<td>nA</td>
<td></td>
</tr>
</tbody>
</table>
Table 10. Electrical characteristics at $V_{CC+} = 1.8$ V, with $V_{CC-} = 0$ V, $V_{icm} = 0.7$ V, T = 25°C, and OUT pin connected to $V_{CC} / 2$ through $R_L = 10$ kΩ (unless otherwise specified)

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Conditions</th>
<th>Min.</th>
<th>Typ.</th>
<th>Max.</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>$V_{io}$</td>
<td>Input offset voltage ($V_{icm} = 0$ V)</td>
<td>$T = 25°C$</td>
<td>±50</td>
<td>±250</td>
<td></td>
<td>µV</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40°C &lt; T &lt; 125°C$</td>
<td></td>
<td>±650</td>
<td></td>
<td></td>
</tr>
<tr>
<td>$\Delta V_{io}/\Delta T$</td>
<td>Input offset voltage drift ($V_{icm} = 0$ V)</td>
<td>$-40°C &lt; T &lt; 125°C$</td>
<td>1</td>
<td>2</td>
<td></td>
<td>µV/°C</td>
</tr>
<tr>
<td>$I_b$</td>
<td>Input bias current ($V_{OUT} = V_{CC}/2$)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>pA</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$T = 25°C$</td>
<td>1</td>
<td>2</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40°C &lt; T &lt; 85°C$</td>
<td>10</td>
<td>30</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40°C &lt; T &lt; 125°C$</td>
<td>75</td>
<td>200</td>
<td></td>
<td></td>
</tr>
<tr>
<td>$I_o$</td>
<td>Input offset current ($V_{OUT} = V_{CC}/2$)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>pA</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$T = 25°C$</td>
<td>1</td>
<td>2</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40°C &lt; T &lt; 85°C$</td>
<td>5</td>
<td>20</td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40°C &lt; T &lt; 125°C$</td>
<td>20</td>
<td>100</td>
<td></td>
<td></td>
</tr>
<tr>
<td>CMR1</td>
<td>Common-mode rejection ratio $20 \log(\Delta V_{icm}/\Delta V_{io})$, $V_{icm} = 0$ V to $V_{CC-} = 1.1$ V, $R_L &gt; 1$ MΩ</td>
<td>$T = 25°C$</td>
<td>72</td>
<td>93</td>
<td></td>
<td>dB</td>
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<tr>
<td></td>
<td></td>
<td>$-40°C &lt; T &lt; 125°C$</td>
<td>68</td>
<td></td>
<td></td>
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<tr>
<td>CMR2</td>
<td>Common-mode rejection ratio $20 \log(\Delta V_{icm}/\Delta V_{io})$, $V_{icm} = -0.1$ V to $V_{CC-} = 1.1$ V, $R_L &gt; 1$ MΩ</td>
<td>$T = 25°C$</td>
<td>70</td>
<td></td>
<td></td>
<td>dB</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40°C &lt; T &lt; 125°C$</td>
<td>52</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>$A_{VD}$</td>
<td>Large signal voltage gain $V_{OUT} = 0.3$ V to ($V_{CC-} = 0.3$ V)</td>
<td>$T = 25°C$</td>
<td>101</td>
<td>122</td>
<td></td>
<td>dB</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40°C &lt; T &lt; 125°C$</td>
<td>97</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>$V_{OH}$</td>
<td>High level output voltage ($V_{OH} = V_{CC} - V_{OUT}$)</td>
<td>$T = 25°C$</td>
<td>15</td>
<td></td>
<td></td>
<td>mV</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40°C &lt; T &lt; 125°C$</td>
<td>25</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>$V_{OL}$</td>
<td>Low level output voltage</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>mV</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$T = 25°C$</td>
<td>15</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40°C &lt; T &lt; 125°C$</td>
<td>25</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>$I_{OUT}$</td>
<td>$I_{sink}$ ($V_{OUT} = V_{CC}$)</td>
<td>$T = 25°C$</td>
<td>35</td>
<td>42</td>
<td></td>
<td>mA</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40°C &lt; T &lt; 125°C$</td>
<td>20</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>$I_{source}$ ($V_{OUT} = 0$ V)</td>
<td>$T = 25°C$</td>
<td>20</td>
<td>32</td>
<td></td>
<td>mA</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40°C &lt; T &lt; 125°C$</td>
<td>10</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>Supply current (per channel, $V_{OUT} = V_{CC}/2$, $R_L &gt; 1$ MΩ)</td>
<td>$T = 25°C$</td>
<td>1.7</td>
<td>2.2</td>
<td></td>
<td>mA</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$-40°C &lt; T &lt; 125°C$</td>
<td>2.5</td>
<td></td>
<td></td>
<td></td>
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<tr>
<td>AC Performance</td>
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<td></td>
<td></td>
</tr>
<tr>
<td>GBW</td>
<td>Gain bandwidth product</td>
<td></td>
<td></td>
<td>14</td>
<td>21</td>
<td>MHz</td>
</tr>
<tr>
<td>$F_u$</td>
<td>Unity gain frequency</td>
<td></td>
<td></td>
<td>18</td>
<td></td>
<td></td>
</tr>
<tr>
<td>$\Phi_m$</td>
<td>Phase margin</td>
<td>$C_L = 47$ pF</td>
<td></td>
<td>41</td>
<td></td>
<td>degrees</td>
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<tr>
<td>$G_m$</td>
<td>Gain margin</td>
<td></td>
<td></td>
<td>8</td>
<td></td>
<td>dB</td>
</tr>
<tr>
<td>SR</td>
<td>Slew rate (2)</td>
<td></td>
<td></td>
<td>7.6</td>
<td>11</td>
<td>V/µs</td>
</tr>
<tr>
<td>$e_n$</td>
<td>Equivalent input noise voltage</td>
<td>$f = 1$ kHz</td>
<td></td>
<td>13</td>
<td></td>
<td>nV/√Hz</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$f = 10$ kHz</td>
<td></td>
<td>7</td>
<td></td>
<td></td>
</tr>
<tr>
<td>$C_S$</td>
<td>Channel separation (for TSV7722 and TSV7723)</td>
<td>$f = 1$ kHz</td>
<td></td>
<td>120</td>
<td></td>
<td>dB</td>
</tr>
</tbody>
</table>

1. Guaranteed by design and characterization on a sample of parts, not tested in production.
2. Slew rate value is calculated as the average between positive and negative slew rates.
Table 11. Electrical characteristics in Shutdown mode at $V_{CC+} = 1.8$ V, with $V_{CC-} = 0$ V, $V_{icm} = 0.7$ V, $T = 25^\circ$C, and OUT pin connected to $V_{CC} / 2$ through $R_L = 10$ kΩ (unless otherwise specified)

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Conditions</th>
<th>Min.</th>
<th>Typ.</th>
<th>Max.</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>$I_{CC}$</td>
<td>Shutdown mode $V_{OUT} = V_{CC}/2$, $R_L &gt; 1$ MΩ (all channels)</td>
<td>$T = 25^\circ$C</td>
<td>2.5</td>
<td>60</td>
<td></td>
<td>nA</td>
</tr>
<tr>
<td></td>
<td></td>
<td>-40°C &lt; $T &lt; 85^\circ$C</td>
<td></td>
<td>450</td>
<td></td>
<td>nA</td>
</tr>
<tr>
<td></td>
<td></td>
<td>-40°C &lt; $T &lt; 125^\circ$C</td>
<td></td>
<td>4</td>
<td></td>
<td>µA</td>
</tr>
<tr>
<td>$t_{on}$</td>
<td>Amplifier turn-on time (other channel already on)</td>
<td>$V_{OUT} = V_{CC} - to V_{CC} - + 0.2$ V</td>
<td>1.5</td>
<td></td>
<td></td>
<td>µs</td>
</tr>
<tr>
<td>$t_{init}$</td>
<td>Initialization time (both channels off)</td>
<td>$V_{OUT}$ to 200 mV of final value</td>
<td>38</td>
<td></td>
<td></td>
<td>µs</td>
</tr>
<tr>
<td>$V_{IH}$</td>
<td>EN logic high</td>
<td>1.2</td>
<td></td>
<td></td>
<td></td>
<td>V</td>
</tr>
<tr>
<td>$V_{IL}$</td>
<td>EN logic low</td>
<td>0.6</td>
<td></td>
<td></td>
<td></td>
<td>V</td>
</tr>
<tr>
<td>$I_{IH}$</td>
<td>EN current high</td>
<td>$EN = V_{CC+}$</td>
<td>1</td>
<td></td>
<td></td>
<td>pA</td>
</tr>
<tr>
<td>$I_{IL}$</td>
<td>EN current low</td>
<td>$EN = V_{CC-}$</td>
<td>1</td>
<td></td>
<td></td>
<td>pA</td>
</tr>
<tr>
<td>$I_{Oleak}$</td>
<td>Output leakage in shutdown mode, $EN = V_{CC-}$</td>
<td>$T = 25^\circ$C</td>
<td>50</td>
<td></td>
<td></td>
<td>pA</td>
</tr>
<tr>
<td></td>
<td></td>
<td>-40°C &lt; $T &lt; 125^\circ$C</td>
<td>15</td>
<td></td>
<td></td>
<td>nA</td>
</tr>
</tbody>
</table>
4 Typical performance characteristics

$R_L = 10\ k\Omega$ connected to $V_{CC}/2$ and $C_L = 47\ pF$, unless otherwise specified.

**Figure 4. Supply current vs. supply voltage**

**Figure 5. Input offset voltage distribution at $V_{CC} = 5\ V$**

**Figure 6. Input offset voltage distribution at $V_{CC} = 1.8\ V$**

**Figure 7. Input offset voltage vs. temperature at $V_{CC} = 5\ V$**

**Figure 8. Input offset voltage vs. temperature at $V_{CC}=1.8\ V$**

**Figure 9. Input offset voltage thermal coeff. at $V_{CC}=5\ V$**
Figure 10. Input offset voltage thermal coefficient at $V_{CC}=1.8$ V

Figure 11. Input offset voltage vs. supply voltage

Figure 12. Input offset voltage vs. common-mode voltage at $V_{CC} = 5$ V

Figure 13. Input offset voltage vs. common-mode voltage at $V_{CC} = 1.8$ V

Figure 14. Input bias current vs. temp. at $V_{ICM} = V_{CC} / 2$

Figure 15. Input bias current vs. common-mode voltage at $V_{CC} = 5$ V
Figure 16. Output current vs. output voltage at $V_{CC} = 5\,\text{V}$

Figure 17. Output current versus output voltage at $V_{CC}=1.8\,\text{V}$

Figure 18. Output saturation voltage ($V_{OL}$) vs. supply voltage

Figure 19. Output saturation voltage ($V_{OH}$) vs. supply voltage

Figure 20. Positive slew rate at $V_{CC} = 5\,\text{V}$

Figure 21. Negative slew rate at $V_{CC} = 5\,\text{V}$
Figure 22. Slew rate vs. V\textsubscript{CC}

Figure 23. Open loop bode diagram at V\textsubscript{CC} = 5 V

Figure 24. Open loop bode diagram at V\textsubscript{CC} = 1.8 V

Figure 25. Closed loop bode diagram at V\textsubscript{CC} = 5 V

Figure 26. Closed loop bode diagram at V\textsubscript{CC} = 1.8 V

Figure 27. Phase margin vs. common-mode voltage and load current at V\textsubscript{CC} = 5 V
Figure 28. Phase margin vs. capacitive load

Figure 29. Small step response at $V_{CC} = 5\, V$

Figure 30. Small step response at $V_{CC} = 1.8\, V$

Figure 31. Desaturation from low rail at $V_{CC} = 5\, V$

Figure 32. Desaturation from high rail at $V_{CC} = 5\, V$

Figure 33. Settling time output high to low at $V_{CC} = 5\, V$
Figure 34. Settling time output low to high at $V_{CC} = 5$ V

Figure 35. Small step overshoot vs. load capacitance

Figure 36. Linearity vs. load resistance at $V_{CC} = 5$ V

Figure 37. Noise vs. frequency

Figure 38. Noise versus time at $V_{CC} = 5$ V

Figure 39. THD+N vs. frequency
Figure 40. THD+N vs. output voltage

Figure 41. CMRR vs. frequency at $V_{CC} = 5\,V$

Figure 42. PSRR vs. frequency at $V_{CC} = 5\,V$

Figure 43. Supply current vs. supply voltage in shutdown mode

Figure 44. Turn-on time at $V_{CC} = 5\,V$

Figure 45. Turn-on time at $V_{CC} = 1.8\,V$
5 Application information

5.1 Operating voltages

The TSV7722 device can operate from 1.8 to 5.5 V. The parameters are fully specified at 1.8 V, 3.3 V and 5 V power supplies. However, the parameters are very stable over the full $V_{CC}$ range and several characterization curves show the TSV7722 device characteristics over the full operating range. Additionally, the main specifications are guaranteed in extended temperature range from -40 to 125 °C.

The TSV7722 device is low rail input, and rail-to-rail output. The common-mode operating range is from $V_{cc-} - 0.1 V$, to $V_{cc+} - 1.1 V$. The op amp $V_{io}$ is trimmed at $V_{cc} = 3.3 V$, $V_{icm} = 0 V$, and thus the DC precision is optimized for operation with $V_{icm}$ close to $Vcc$.

5.2 Input offset voltage drift over the temperature

The maximum input voltage drift over temperature is defined as the offset variation related to the offset value measured at 25 °C. The operational amplifier is one of the main circuits of the signal conditioning chain, and the amplifier input offset is a major contributor to the chain accuracy. The signal chain accuracy at 25 °C can be compensated during production at application level. The maximum input voltage drift overtemperature enables the system designer to anticipate the effect of temperature variations.

The maximum input voltage drift overtemperature is computed using the following equation:

$$\frac{\Delta V_{io}}{\Delta T} = \max \left[ \frac{V_{io}(T) - V_{io}(25°C)}{T - 25°C} \right]$$

(1)

Where $T = -40 °C$ and 125 °C.

The TSV7721, TSV7722, TSV7723 datasheet maximum value is guaranteed by measurements on a representative sample size ensuring a $C_{pk}$ (process capability index) greater than 1.3.

5.3 Unused channel

When one of the two channels of the TSV7722 is not used, it must be properly connected in order to avoid internal oscillations that can negatively impact the signal integrity on the other channel, as well as the current consumption. Two different configurations can be used:

Gain configuration: the channel can be set in gain, the input can be set to any voltage within the $V_{icm}$ operating range.

Comparator configuration: the channel can be set to a comparator configuration (without negative feedback). In this case, positive and negative inputs can be set to any value provided these values are significantly different (100 mV or more, to avoid oscillation between positive and negative state).

5.4 EMI rejection

The electromagnetic interference (EMI) rejection ratio, or EMIIRR, describes the EMI immunity of operational amplifiers. An adverse effect that is common to many op amps is a change in the offset voltage as a result of RF signal rectification. EMIIRR is defined in Eq. (2):

$$EMIRR = 20 \cdot \log \left( \frac{V_{in\, pp}}{\Delta V_{io}} \right)$$

(2)

The TSV7722 has been specially designed to minimize susceptibility to EMIIRR and shows a low sensitivity. As can be seen in Figure 46. EMIIRR on In+, In- and Out pins, EMI rejection ratio has been measured on both inputs and output, from 400 MHz to 2.4 GHz.
EMIRR performances might be improved by adding small capacitances (in the pF range) on the inputs, power supply and output pins. These capacitances help to minimize the impedence of these nodes at high frequencies.

5.5 Maximum power dissipation

The usable output load current drive is limited by the maximum power dissipation allowed by the device package. The absolute maximum junction temperature for the TSV7722 is 150 °C. The junction temperature can be estimated as follows:

\[ T_J = P_D \times \theta_{JA} + T_A \] \hspace{1cm} (3)

- **T**\(_J\) is the die junction temperature
- **P**\(_D\) is the power dissipated in the package
- **\(\theta_{JA}\)** is the junction to ambient thermal resistance of the package.
- **T**\(_A\) is the ambient temperature.

The power dissipated in the package **P**\(_D\) is the sum of the quiescent power dissipated and the power dissipated by the output stage transistor. It is calculated as follows:

- When the op amp is sourcing the current:
  \[ P_D = (V_{CC} \times I_{CC}) + (V_{CC} + V_{OUT}) \times I_{Load} \]
- When the op amp is sinking the current:
  \[ P_D = (V_{CC} \times I_{CC}) + (V_{OUT} - V_{CC}) \times I_{Load} \]

Do not exceed the 150 °C maximum junction temperature for the device. Exceeding the junction temperature limit can cause degradation in the parametric performance or even destroy the device.

5.6 Capacitive load and stability

Stability analysis must be performed for large capacitive loads over 47 pF; increasing the load capacitance to high values produces gain peaking in the frequency response, with overshoot and ringing in the step response. Generally, unity gain configuration is the worst situation for stability and the ability to drive large capacitive loads. For additional capacitive load drive capability in unity-gain configuration, stability can be improved by inserting a small resistor **R**\(_{ISO}\) (10 Ω to 22 Ω) in series with the output (see Figure 35. Small step overshoot vs. load capacitance). This resistor significantly reduces ringing while maintaining DC performance for purely capacitive loads. However, if there is a resistive load in parallel with the capacitive load, a voltage divider is created introducing a gain error at the output and slightly reducing the output swing. The error introduced is proportional to the ratio **R**\(_{ISO}\) / **R**\(_L\). **R**\(_{ISO}\) modifies the maximum capacitive load acceptable from a stability point of view, as described in Figure 47. Test configuration for **R**\(_{ISO}\):
Please note that $R_{\text{ISO}} = 22 \, \Omega$ is sufficient to make the TSV7722 stable whatever the capacitive load.

### 5.7 Resistor values for high speed op amp design

Due to its high gain bandwidth product (GBP), this op amp is particularly sensitive to parasitic impedances. Board parasitics should be taken into account in any sensitive design. Indeed, excessive parasitic (both capacitive and inductive) in the op amp frequency range can alter performances and stability. These issues can often be mitigated by lowering the resistive impedances. More specifically, the RC network created by the schematic resistors ($R_f$ and $R_g$) and the parasitic capacitances of both the op amp (as documented in Table 6 to Table 10 and illustrated in Figure 48) and the PCB can generate a pole below or in the same order of magnitude than the closed-loop bandwidth of the circuit. In this case, the feedback circuit is not able to fully play its role at high frequency, and the application can be unstable. This issue can happen when the schematic gain is low (typically $< 5$), or the device is used in follower mode with a resistor in the feedback. In these cases, it is advised to use a low value feedback resistor ($R_f$), typically 1 kΩ.

Also, some designs use an input resistor on the positive input, generally of the same value than the input resistance on the negative input. This resistor can be useful to balance the input currents on the positive and negative inputs, and reduce the impact of those input currents on precision. However, this is not useful on the TSV7722 as the input currents are very low. Furthermore, this resistor can also interact with the input capacitances to generate a pole. The frequency of this pole should be kept higher than the closed-loop bandwidth frequency. The macromodel provided takes into account the circuit parasitic capacitors. Thus, a transient SPICE simulation (100 mV step) is an easy way to evaluate the stability of the application. However, this cannot replace a hardware evaluation of the application circuit.
5.8 Settling time

Settling time in an application can be defined as the amount of time between the input changes, and the output reaching its final value. It is usually defined with a given tolerance, so the output stability is reached when the output stays within the given range around the final value. In Figure 33. Settling time output high to low at $V_{CC} = 5\, V$ and Figure 34. Settling time output low to high at $V_{CC} = 5\, V$, the settling time is measured in an inverting configuration, using the so-called “false summing node” circuit.

Figure 49. Settling time measurement configuration

This circuit is used with a step input voltage from a positive or negative value, to 0 V. The measurement point being $(V_{in} + V_{out}) / 2$, and $V_{out}$ being in an ideal circuit equal to $V_{in}$, the measurement point gives half of the error on $V_{out}$, comparatively to $V_{in}$. This error is compared to the tolerance, 0.1% for this circuit, to deduce the settling time. This characteristic is particularly useful when driving an ADC. It is related to the slew rate, GBP and stability of the circuit. It also varies with the circuit gain, the circuit load, and the input voltage step value. However, computing the value of the settling time in a given configuration is not straightforward. The macromodel can give a good estimation, but prototyping can be needed for fine circuit optimization.
5.9 **Shutdown function (TSV7723)**

The operational amplifier is enabled when the EN pin is pulled high. To disable the amplifier, the EN must be pulled down to VCC-. When in shutdown mode, the amplifier output is in a high impedance state. The EN pin must never be left floating, but must be tied to VCC+ or VCC-.

The turn-on time is calculated for an output variation of ± 200 mV (see Figure 47 & Figure 48. Figure 51 shows the test configurations).

![Figure 50. Test configuration](image.png)

5.10 **PCB layout recommendations**

Particular attention must be paid to the layout of the PCB tracks connected to the amplifier, load, and power supply. The power and ground traces are critical as they must provide adequate energy and grounding for all circuits. The best practice is to use short and wide PCB traces to minimize voltage drops and parasitic inductance. In addition, to minimizing parasitic impedance over the entire surface, a multi-via technique that connects the bottom and top layer ground planes together in many locations is often used. The copper traces that connect the output pins to the load and supply pins should be as wide as possible to minimize trace resistance.

5.11 **Decoupling capacitor**

In order to ensure op amp full functionality, it is mandatory to place a decoupling capacitor of at least 22 nF as close as possible to the op amp supply pins. A good decoupling helps to reduce electromagnetic interference impact.

5.12 **Macro model**

Accurate macro models of the TSV7722 device are available on the STMicroelectronics’ website at: www.st.com. These models are a trade-off between accuracy and complexity (that is, time simulation) of the TSV7722 operational amplifier. They emulate the nominal performance of a typical device at 25°C within the specified operating conditions mentioned in the datasheet. They also help to validate a design approach and to select the right operational amplifier, but they do not replace on-board measurements.
6 Typical applications

6.1 Low-side current sensing

Power management mechanisms are found in most electronic systems. Current sensing is useful for protecting applications. The low-side current sensing method consists of placing a sense resistor between the load and the circuit ground. The resulting voltage drop is amplified using the TSV772x (see Figure 51. Low-side current sensing schematic).

**Figure 51. Low-side current sensing schematic**

\[ V_{out} = R_{shunt} \cdot I \left( 1 - \frac{R_g}{R_{g2} + R_f} \right) \left( 1 + \frac{R_f}{R_g} \right) + I_p \cdot \frac{R_g \cdot R_f}{R_{g2} + R_{f2}} \left( 1 + \frac{R_f}{R_g} \right) - I_n \cdot R_f - V_{io} \left( 1 + \frac{R_f}{R_g} \right) \]  

(4)

Assuming that \( R_{g2} = R_{g1} = R_g \), \( R_{f2} = R_{f1} = R_f \), this equation can be simplified as follows:

\[ V_{out} = R_{shunt} \cdot I \cdot \frac{R_f}{R_g} - V_{io} \left( 1 + \frac{R_f}{R_g} \right) + R_f \cdot I_{io} \]  

(5)

The main advantage of using the TSV7722 for a low-side current sensing relies on its low \( V_{io} \), compared to general purpose operational amplifiers. For the same current and targeted accuracy, the shunt resistor can be chosen with a lower value, resulting in lower power dissipation, lower drop in the ground path, and lower cost. Particular attention must be paid to the matching and precision of \( R_{g1}, R_{g2}, R_f, R_{shunt} \), to maximize the accuracy of the measurement. Furthermore, on the TSV7722, the \( V_{io} \) is trimmed, and thus reaches its minimum value, at \( V_{icm} = 0 \) V. This allows optimized precision for low-side current sensing application without precision degradation due to the CMRR.
6.2 Photodiode transimpedance amplification

The TSV7722, with high bandwidth and slew rate, is well suited for photodiode signal conditioning in a transimpedance amplifier circuit. This application is useful in high performance UV sensors, smoke detectors or particle sensors.

![Photodiode transimpedance amplifier circuit](image)

The transimpedance amplifier circuit converts the small photodiode output current in the nA range, into a voltage signal readable by an ADC following Eq. (6):

\[ V_{Out} = R_f \cdot I_{photodiode} \] (6)

The feedback resistance is usually in the MΩ range, in order to get a large enough voltage output range. However, together with the diode parasitic capacitance, the op amp input capacitances and the PCB stray capacitance, this feedback network creates a pole that makes the circuit oscillate. Using a small (few pF) capacitor in parallel with the feedback resistor is mandatory to stabilize the circuit. The value of this capacitor can be tuned to optimize the application settling time with a SPICE simulation using the op amp macromodel, or by prototyping.

For more details on tuning this circuit, please read the application note AN4451.
In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at: www.st.com. ECOPACK is an ST trademark.
7.1 SOT23-5 package information

Figure 53. SOT23-5 package outline

Table 12. SOT23-5 package mechanical data

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<tr>
<td>A2</td>
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<td>C</td>
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7.2 DFN8 2x2 package information

Figure 54. DFN8 2x2 package outline

Table 13. DFN8 2x2 package mechanical data

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<tr>
<td>ddd</td>
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Figure 55. DFN8 2x2 recommended footprint

Note: The exposed pad of the DFN8 2x2 can be connected to VCC- or left floating.
### 7.3 MiniSO8 package information

**Figure 56. MiniSO8 package outline**

![MiniSO8 package outline](image)

**Table 14. MiniSO8 package mechanical data**

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7.4 SO8 package information

Figure 57. SO8 package outline

Table 15. SO8 mechanical data

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7.5 MiniSO10 package information

Figure 58. SO8 recommended footprint

Figure 59. MiniSO10 package outline
Table 16. MiniSO10 mechanical data

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## 8 Ordering information

### Table 17. Order code

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<td>•</td>
<td>K217</td>
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<td>-40 to +125°C</td>
<td>DFN8 2x2</td>
<td>2</td>
<td>K2A</td>
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1. Qualified and characterized according to AEC Q100 and Q003 or equivalent, advanced screening according to AEC Q001 and Q002 or equivalent. For qualification status detail, check "Maturity Status Link" on the first page of the datasheet, then the "Quality and Reliability" tab on www.st.com
## Table 18. Document revision history

<table>
<thead>
<tr>
<th>Date</th>
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<td>Initial release.</td>
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<tr>
<td>16-Mar-2021</td>
<td>2</td>
<td>Updated the &quot;Related products&quot; table in cover page. Added Section 1 Pin description, Section 1.1 TSV7721 single operational amplifier, Section 1.2 TSV7722 dual operational amplifier and Section 1.3 TSV7723 dual operational amplifier with shutdown option. Changed from 2.5 mA to 2.8 mA for &quot;Maximum supply current -40 °C &lt; T &lt; 125 °C and Vcc=5 V, 3.3 V, 1.8 V&quot;. Minor text changes.</td>
</tr>
<tr>
<td>25-May-2021</td>
<td>3</td>
<td>Changed name and description pin 5, pin 6 in Figure 3 and Table 3. Updated: $V_{IH}$, $V_{IL}$, $I_{IH}$, $I_{IL}$ parameter in Table 6, Table 7 and Table 8, Figure 20 and Figure 21. Added: Figure 43, Figure 44, Figure 45 and Section 5.9.</td>
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<tr>
<td>13-Oct-2021</td>
<td>4</td>
<td>Updated $I_{CC}$ parameter and max. value row $T = 25°C$ in Table 6, Table 8 and Table 10.</td>
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<tr>
<td>15-Nov-2021</td>
<td>5</td>
<td>Updated Updated Figure 41 and Figure 42.</td>
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<tr>
<td>29-Mar-2022</td>
<td>6</td>
<td>Updated title, features and related products on the cover page.</td>
</tr>
<tr>
<td>19-Dec-2022</td>
<td>7</td>
<td>Added $I_{g}$ and $I_{o}$ new values, -40 °C ≤ T ≤ 85 °C conditions in Table 6, Table 8 and Table 10.</td>
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