STM32G431xx/441xx device errata

Applicability

This document applies to the part numbers of STM32G431xx/441xx devices and the device variants as stated in this page.

It gives a summary and a description of the device errata, with respect to the device datasheet and reference manual RM0440.

Deviation of the real device behavior from the intended device behavior is considered to be a device limitation. Deviation of the description in the reference manual or the datasheet from the intended device behavior is considered to be a documentation erratum. The term *"errata"* applies both to limitations and documentation errata.

Table 1. Device summary

Reference	Part numbers
STM32G431xx	STM32G431C6, STM32G431C8, STM32G431CB, STM32G431K6, STM32G431K8, STM32G431KB, STM32G431M6, STM32G431M8, STM32G431MB, STM32G431R6, STM32G431R8, STM32G431RB, STM32G431V6, STM32G431V8, STM32G431VB
STM32G441xx	STM32G441CB, STM32G441KB, STM32G441MB, STM32G441RB, STM32G441VB

Table 2. Device variants

Reference	Silicon revision codes	
	Device marking(1)	REV_ID(2)
STM32G431xx/441xx	Z	0x2001
STM32G431xx/441xx	Y	0x2002
STM32G431xxx/441xx	X	0x2003

1. *Refer to the device datasheet for how to identify this code on different types of package.*
2. *REV_ID[15:0] bitfield of DBGMCU_IDCODE register.*

1 Summary of device errata

The following table gives a quick reference to the STM32G431xx/441xx device limitations and their status:

A = workaround available

N = no workaround available

P = partial workaround available

Applicability of a workaround may depend on specific conditions of target application. Adoption of a workaround may cause restrictions to target application. Workaround for a limitation is deemed partial if it only reduces the rate of occurrence and/or consequences of the limitation, or if it is fully effective for only a subset of instances on the device or in only a subset of operating modes, of the function concerned.

Table 3. Summary of device limitations

Function	Section	Limitation	Status		
			Rev. Z	Rev. Y	Rev. X
Core	2.1.1	Interrupted loads to SP can cause erroneous behavior	A	A	A
	2.1.2	VDIV or VSQRT instructions might not complete correctly when very short ISRs are used	A	A	A
	2.1.3	Store immediate overlapping exception return operation might vector to incorrect interrupt	A	A	A
System	2.2.1	Full JTAG configuration without NJTRST pin cannot be used	A	A	A
	2.2.2	FLASH_ECCR corrupted upon reset or power-down occurring during Flash memory program or erase operation	A	A	A
	2.2.3	Unstable LSI when it clocks RTC or CSS on LSE	P	P	P
	2.2.4	PCROP_RDP option bit production value not inline with the specification	N	-	-
DMA	2.3.1	DMA disable failure and error flag omission upon simultaneous transfer error and global flag clear	A	A	A
DMAMUX	2.4.1	SOFx not asserted when writing into DMAMUX_CFR register	N	N	N
	2.4.2	OFx not asserted for trigger event coinciding with last DMAMUX request	N	N	N
	2.4.3	OFx not asserted when writing into DMAMUX_RGCFR register	N	N	N
	2.4.4	Wrong input DMA request routed upon specific DMAMUX_CxCR register write coinciding with synchronization event	A	A	A
ADC	2.5.1	New context conversion initiated without waiting for trigger when writing new context in ADC_JSQR with JQDIS = 0 and JQM = 0	A	A	A
	2.5.2	Two consecutive context conversions fail when writing new context in ADC_JSQR just after previous context completion with JQDIS = 0 and JQM = 0	A	A	A
	2.5.3	Unexpected regular conversion when two consecutive injected conversions are performed in Dual interleaved mode	A	A	A
	2.5.4	ADC_AWDy_OUT reset by non-guarded channels	A	A	A
	2.5.5	End of 10/8/6-bit ADC conversion disturbing other ADCs	A	-	-
	2.5.6	ADC input channel switching disturbs ongoing conversions	P	-	-
	2.5.7	Wrong ADC result if conversion done late after calibration or previous conversion	A	A	A
	2.5.8	ADC channel 0 converted instead of the required ADC channel	-	A	-

Function	Section	Limitation	Status		
			Rev. Z	Rev. Y	Rev. X
COMP	2.6.1	Comparator previously not fully tested in production	N	N	-
TIM	2.7.1	One-pulse mode trigger not detected in master-slave reset + trigger configuration	P	P	P
	2.7.2	Consecutive compare event missed in specific conditions	N	N	N
	2.7.3	Compare event missed in center-aligned mode 1 and 2 with dithering enabled	N	N	N
	2.7.4	Output compare clear not working with external counter reset	P	P	P
LPTIM	2.8.1	Device may remain stuck in LPTIM interrupt when entering Stop mode	A	A	A
	2.8.2	Device may remain stuck in LPTIM interrupt when clearing event flag	P	P	P
	2.8.3	LPTIM events and PWM output are delayed by 1 kernel clock cycle	P	P	P
RTC and TAMP	2.9.1	Calendar initialization may fail in case of consecutive INIT mode entry	A	A	A
	2.9.2	Alarm flag may be repeatedly set when the core is stopped in debug	N	N	N
	2.9.3	A tamper event fails to trigger timestamp or timestamp overflow events during a few cycles after clearing TSF	N	N	N
	2.9.4	REFCKON write protection associated to INIT KEY instead of CAL KEY	A	A	A
	2.9.5	Tamper flag not set on LSE failure detection	N	N	N
I2C	2.10.1	Wrong data sampling when data setup time (tSU;DAT) is shorter than one I2C kernel clock period	P	P	P
	2.10.2	Spurious bus error detection in master mode	A	A	A
	2.10.3	Spurious master transfer upon own slave address match	P	P	P
	2.10.4	OVR flag not set in underrun condition	N	N	N
	2.10.5	Transmission stalled after first byte transfer	A	A	A
USART	2.11.1	Anticipated end-of-transmission signaling in SPI slave mode	A	A	A
	2.11.2	Data corruption due to noisy receive line	N	N	N
SPI	2.12.1	BSY bit may stay high when SPI is disabled	A	A	A
	2.12.2	BSY bit may stay high at the end of data transfer in slave mode	A	A	A
FDCAN	2.13.1	Desynchronization under specific condition with edge filtering enabled	A	A	A
	2.13.2	Tx FIFO messages inverted under specific buffer usage and priority setting	A	A	A
USB	2.14.1	ESOF interrupt timing desynchronized after resume signaling	A	A	A
	2.14.2	Incorrect CRC16 in the memory buffer	N	N	N

2 Description of device errata

The following sections describe limitations of the applicable devices with Arm® core and provide workarounds if available. They are grouped by device functions.

Note: Arm is a registered trademark of Arm Limited (or its subsidiaries) in the US and/or elsewhere.

2.1 Core

Reference manual and errata notice for the Arm® Cortex®-M4 core revision r0p1 is available from http://infocenter.arm.com.

2.1.1 Interrupted loads to SP can cause erroneous behavior

This limitation is registered under Arm ID number 752770 and classified into “Category B”. Its impact to the device is minor.

Description

If an interrupt occurs during the data-phase of a single word load to the stack-pointer (SP/R13), erroneous behavior can occur. In all cases, returning from the interrupt will result in the load instruction being executed an additional time. For all instructions performing an update to the base register, the base register will be erroneously updated on each execution, resulting in the stack-pointer being loaded from an incorrect memory location.

The affected instructions that can result in the load transaction being repeated are:

- LDR SP, [Rn],#imm
- LDR SP, [Rn,#imm]!
- LDR SP, [Rn,#imm]
- LDR SP, [Rn]
- LDR SP, [Rn,Rm]

The affected instructions that can result in the stack-pointer being loaded from an incorrect memory address are:

- LDR SP,[Rn],#imm
- LDR SP,[Rn,#imm]!

As compilers do not generate these particular instructions, the limitation is only likely to occur with hand-written assembly code.

Workaround

Both issues may be worked around by replacing the direct load to the stack-pointer, with an intermediate load to a general-purpose register followed by a move to the stack-pointer.

2.1.2 VDIV or VSQRT instructions might not complete correctly when very short ISRs are used

This limitation is registered under Arm ID number 776924 and classified into “Category B”. Its impact to the device is limited.

Description

The VDIV and VSQRT instructions take 14 cycles to execute. When an interrupt is taken a VDIV or VSQRT instruction is not terminated, and completes its execution while the interrupt stacking occurs. If lazy context save of floating point state is enabled then the automatic stacking of the floating point context does not occur until a floating point instruction is executed inside the interrupt service routine.

Lazy context save is enabled by default. When it is enabled, the minimum time for the first instruction in the interrupt service routine to start executing is 12 cycles. In certain timing conditions, and if there is only one or two instructions inside the interrupt service routine, then the VDIV or VSQRT instruction might not write its result to the register bank or to the FPSCR.

The failure occurs when the following condition is met:

1. The floating point unit is enabled
2. Lazy context saving is not disabled
3. A VDIV or VSQRT is executed
4. The destination register for the VDIV or VSQRT is one of s0 - s15
5. An interrupt occurs and is taken
6. The interrupt service routine being executed does not contain a floating point instruction
7. Within 14 cycles after the VDIV or VSQRT is executed, an interrupt return is executed

A minimum of 12 of these 14 cycles are utilized for the context state stacking, which leaves 2 cycles for instructions inside the interrupt service routine, or 2 wait states applied to the entire stacking sequence (which means that it is not a constant wait state for every access).

In general, this means that if the memory system inserts wait states for stack transactions (that is, external memory is used for stack data), then this erratum cannot be observed.

The effect of this erratum is that the VDIV or VQSRT instruction does not complete correctly and the register bank and FPSCR are not updated, which means that these registers hold incorrect, out of date, data.

Workaround

A workaround is only required if the floating point unit is enabled. A workaround is not required if the stack is in external memory.

There are two possible workarounds:

- Disable lazy context save of floating point state by clearing LSPEN to 0 (bit 30 of the FPCCR at address 0xE000EF34).
- Ensure that every interrupt service routine contains more than 2 instructions in addition to the exception return instruction.

2.1.3 Store immediate overlapping exception return operation might vector to incorrect interrupt

This limitation is registered under Arm ID number 838869 and classified into “Category B (rare)”. Its impact to the device is minor.

Description

The core includes a write buffer that permits execution to continue while a store is waiting on the bus. Under specific timing conditions, during an exception return while this buffer is still in use by a store instruction, a late change in selection of the next interrupt to be taken might result in there being a mismatch between the interrupt acknowledged by the interrupt controller and the vector fetched by the processor.

The failure occurs when the following condition is met:

1. The handler for interrupt A is being executed.
2. Interrupt B, of the same or lower priority than interrupt A, is pending.
3. A store with immediate offset instruction is executed to a bufferable location.
   – STR/STRH/STRB <Rt>, [<Rn>,#imm]
   – STR/STRH/STRB <Rt>, [<Rn>,#imm]!
   – STR/STRH/STRB <Rt>, [<Rn>],#imm
4. Any number of additional data-processing instructions can be executed.
5. A BX instruction is executed that causes an exception return.
6. The store data has wait states applied to it such that the data is accepted at least two cycles after the BX is executed.
   – Minimally, this is two cycles if the store and the BX instruction have no additional instructions between them.
   – The number of wait states required to observe this erratum needs to be increased by the number of cycles between the store and the interrupt service routine exit instruction.
7. Before the bus accepts the buffered store data, another interrupt C is asserted which has the same or lower priority as A, but a greater priority than B.

Example:

The processor should execute interrupt handler C, and on completion of handler C should execute the handler for B. If the conditions above are met, then this erratum results in the processor erroneously clearing the pending state of interrupt C, and then executing the handler for B twice. The first time the handler for B is executed it will be at interrupt C's priority level. If interrupt C is pended by a level-based interrupt which is cleared by C's handler then interrupt C will be pended again once the handler for B has completed and the handler for C will be executed.

As the STM32 interrupt C is level based, it eventually becomes pending again and is subsequently handled.

Workaround

For software not using the memory protection unit, this erratum can be worked around by setting DISDEFWBUF in the Auxiliary Control Register.

In all other cases, the erratum can be avoided by ensuring a DSB occurs between the store and the BX instruction. For exception handlers written in C, this can be achieved by inserting the appropriate set of intrinsics or inline assembly just before the end of the interrupt function, for example:

ARMCC:

```
...
__schedule_barrier();
__asm{DSB};
__schedule_barrier();
}
```

GCC:

```
...
__asm volatile ("dsb 0xf":::"memory");
}
```

2.2 System

2.2.1 Full JTAG configuration without NJTRST pin cannot be used

Description

When using the JTAG debug port in Debug mode, the connection with the debugger is lost if the NJTRST pin (PB4) is used as a GPIO. Only the 4-wire JTAG port configuration is impacted.

Workaround

Use the SWD debug port instead of the full 4-wire JTAG port.

2.2.2 FLASH_ECCR corrupted upon reset or power-down occurring during Flash memory program or erase operation

Description

Reset or power-down occurring during a Flash memory location program or erase operation, followed by a read of the same memory location, may lead to a corruption of the FLASH_ECCR register content.

Workaround

Under such condition, erase the page(s) corresponding to the Flash memory location.

2.2.3 Unstable LSI when it clocks RTC or CSS on LSE

Description

The LSI clock can become unstable (duty cycle different from 50 %) and its maximum frequency can become significantly higher than 32 kHz, when:

- LSI clocks the RTC, or it clocks the clock security system (CSS) on LSE (which holds when the LSECSSON bit set), and
- the V_{DD} power domain is reset while the backup domain is not reset, which happens:
  - upon exiting Shutdown mode
  - if V_{BAT} is separate from V_{DD} and V_{DD} goes off then on
  - if V_{BAT} is tied to V_{DD} (internally in the package for products not featuring the VBAT pin, or externally) and a short (< 1 ms) V_{DD} drop under V_{DD}(min) occurs

Workaround

Apply one of the following measures:

- Clock the RTC with LSE or HSE/32, without using the CSS on LSE
- If LSI clocks the RTC or when the LSECSSON bit is set, reset the backup domain upon each V_{DD} power up (when the BORRSTF flag is set). If V_{BAT} is separate from V_{DD}, also restore the RTC configuration, backup registers and anti-tampering configuration.

2.2.4 PCROP_RDP option bit production value not inline with the specification

Description

According to the specification, the PCROP_RDP option bit production value is “0”. In STM32G431 Rev Z devices, with date code 947 and date codes less than 944, the PCROP_RDP option bit production value is “1”

Workaround

None

2.3 DMA

2.3.1 DMA disable failure and error flag omission upon simultaneous transfer error and global flag clear

Description

Upon a data transfer error in a DMA channel x, both the specific TEIFx and the global GIFx flags are raised and the channel x is normally automatically disabled. However, if in the same clock cycle the software clears the GIFx flag (by setting the CGIFx bit of the DMA_IFCR register), the automatic channel disable fails and the TEIFx flag is not raised.

This issue does not occur with ST's HAL software that does not use and clear the GIFx flag when the channel is active.

Workaround

Do not clear GIFx flags when the channel is active. Instead, use HTIFx, TCIFx, and TEIFx specific event flags and their corresponding clear bits.

2.4 DMAMUX

2.4.1 SOFx not asserted when writing into DMAMUX_CFR register

Description

The SOFx flag of the DMAMUX_CSR status register is not asserted if overrun from another DMAMUX channel occurs when the software writes into the DMAMUX_CFR register.

This can happen when multiple DMA channels operate in synchronization mode, and when overrun can occur from more than one channel. As the SOFx flag clear requires a write into the DMAMUX_CFR register (to set the corresponding CSOFx bit), overrun occurring from another DMAMUX channel operating during that write operation fails to raise its corresponding SOFx flag.

Workaround

None. Avoid the use of synchronization mode for concurrent DMAMUX channels, if at least two of them potentially generate synchronization overrun.

2.4.2 OFx not asserted for trigger event coinciding with last DMAMUX request

Description

In the DMAMUX request generator, a trigger event detected in a critical instant of the last-generated DMAMUX request being served by the DMA controller does not assert the corresponding trigger overrun flag OFx. The critical instant is the clock cycle at the very end of the trigger overrun condition.

Additionally, upon the following trigger event, one single DMA request is issued by the DMAMUX request generator, regardless of the programmed number of DMA requests to generate.

The failure only occurs if the number of requests to generate is set to more than two (GNBREQ[4:0] > 00001).

Workaround

Make the trigger period longer than the duration required for serving the programmed number of DMA requests, so as to avoid the trigger overrun condition from occurring on the very last DMA data transfer.

2.4.3 OFx not asserted when writing into DMAMUX_RGCFR register

Description

The OFx flag of the DMAMUX_RGSR status register is not asserted if an overrun from another DMAMUX request generator channel occurs when the software writes into the DMAMUX_RGCFR register. This can happen when multiple DMA channels operate with the DMAMUX request generator, and when an overrun can occur from more than one request generator channel. As the OFx flag clear requires a write into the DMAMUX_RGCFR register (to set the corresponding COFx bit), an overrun occurring in another DMAMUX channel operating with another request generator channel during that write operation fails to raise the corresponding OFx flag.

Workaround

None. Avoid the use of request generator mode for concurrent DMAMUX channels, if at least two channels are potentially generating a request generator overrun.

2.4.4 Wrong input DMA request routed upon specific DMAMUX_CxCR register write coinciding with synchronization event

Description

If a write access into the DMAMUX_CxCR register having the SE bit at zero and SPOL[1:0] bitfield at a value other than 00:

- sets the SE bit (enables synchronization),
- modifies the values of the DMAREQ_ID[5:0] and SYNC_ID[4:0] bitfields, and
- does not modify the SPOL[1:0] bitfield,

and if a synchronization event occurs on the previously selected synchronization input exactly two AHB clock cycles before this DMAMUX_CxCR write, then the input DMA request selected by the DMAREQ_ID[5:0] value before that write is routed.

Workaround

Ensure that the SPOL[1:0] bitfield is at 00 whenever the SE bit is 0. When enabling synchronization by setting the SE bit, always set the SPOL[1:0] bitfield to a value other than 00 with the same write operation into the DMAMUX_CxCR register.

2.5 ADC

2.5.1 New context conversion initiated without waiting for trigger when writing new context in ADC_JSQR with JQDIS = 0 and JQM = 0

Description

Once an injected conversion sequence is complete, the queue is consumed and the context changes according to the new ADC_JSQR parameters stored in the queue. This new context is applied for the next injected sequence of conversions.

However, the programming of the new context in ADC_JSQR (change of injected trigger selection and/or trigger polarity) may launch the execution of this context without waiting for the trigger if:

- the queue of context is enabled (JQDIS cleared to 0 in ADC_CFGR), and
- the queue is never empty (JQM cleared to 0 in ADC_CFGR), and
- the injected conversion sequence is complete and no conversion from previous context is ongoing

Workaround

Apply one of the following measures:

- Ignore the first conversion.
- Use a queue of context with JQM = 1.
- Use a queue of context with JQM = 0, only change the conversion sequence but never the trigger selection and the polarity.

2.5.2 Two consecutive context conversions fail when writing new context in ADC_JSQR just after previous context completion with JQDIS = 0 and JQM = 0

Description

When an injected conversion sequence is complete and the queue is consumed, writing a new context in ADC_JSQR just after the completion of the previous context and with a length longer that the previous context, may cause both contexts to fail. The two contexts are considered as one single context. As an example, if the first context contains element 1 and the second context elements 2 and 3, the first context is consumed followed by elements 2 and 3 and element 1 is not executed.

This issue may happen if:

- the queue of context is enabled (JQDIS cleared to 0 in ADC_CFGR), and
- the queue is never empty (JQM cleared to 0 in ADC_CFGR), and
- the length of the new context is longer than the previous one

Workaround

If possible, synchronize the writing of the new context with the reception of the new trigger.

2.5.3 Unexpected regular conversion when two consecutive injected conversions are performed in Dual interleaved mode

Description

In Dual ADC mode, an unexpected regular conversion may start at the end of the second injected conversion without a regular trigger being received, if the second injected conversion starts exactly at the same time than the end of the first injected conversion. This issue may happen in the following conditions:

- two consecutive injected conversions performed in Interleaved simultaneous mode (DUAL[4:0] of ADC_CCR = 0b00011), or
- two consecutive injected conversions from master or slave ADC performed in Interleaved mode (DUAL[4:0]of ADC_CCR = 0b00111)

Workaround

- In Interleaved simultaneous injected mode: make sure the time between two injected conversion triggers is longer than the injected conversion time.
- In Interleaved only mode: perform injected conversions from one single ADC (master or slave), making sure the time between two injected triggers is longer than the injected conversion time.

2.5.4 ADC_AWDy_OUT reset by non-guarded channels

Description

ADC_AWDy_OUT is set when a guarded conversion of a regular or injected channel is outside the programmed thresholds. It is reset after the end of the next guarded conversion that is inside the programmed thresholds.

However, the ADC_AWDy_OUT signal is also reset at the end of conversion of non-guarded channels, both regular and injected.

Workaround

When ADC_AWDy_OUT is enabled, it is recommended to use only the ADC channels that are guarded by a watchdog.

If ADC_AWDy_OUT is used with ADC channels that are not guarded by a watchdog, take only ADC_AWDy_OUT rising edge into account.

2.5.5 End of 10/8/6-bit ADC conversion disturbing other ADCs

Description

The end-of-conversion event of an ADC instance set to 10-, 8-, or 6-bit resolution disturbs the reference voltage, causing a conversion error to any other ADC instances with conversion in progress.

Workaround

Either set the ADCs to 12-bit resolution, or, with concurrent operation of multiple ADCs, avoid the end of conversion of one ADC to occur during the conversion phase of another ADC. For example, set them all to the same resolution and the same sampling duration, and start their sampling phase at the same time.

2.5.6 ADC input channel switching disturbs ongoing conversions

Description

A switch of the ADC input multiplexer to a different input channel disturbs all ADC instances with conversion in progress, thus negatively impacting their DNL. As the device operates the multiplexer during the ADC conversion, either to select the following input of a scan sequence or to open all multiplexer inputs (idle mode), this disturbance source may affect all operating modes except continuous and except bulb sampling.

The DNL impact depends on the input channel and it increases with increasing number of concurrently converting ADCs, the worst case (a few LSBs) being when all the ADCs on the device operate synchronously.

Workaround

When using a single ADC input channel, activate bulb sampling. This prevents the opening of all multiplexer inputs during the conversion cycle.

In scan conversion mode, set the input scan sequence such that two successive conversions are performed on each input. Keep the result of the first and discard the second, as only the second is affected by the input channel switch disturbance.

2.5.7 Wrong ADC result if conversion done late after calibration or previous conversion

Description

The result of an ADC conversion done more than 1 ms later than the previous ADC conversion or ADC calibration might be incorrect.

Workaround

Perform two consecutive ADC conversions in single, scan or continuous mode. Reject the result of the first conversion and only keep the result of the second.

2.5.8 ADC channel 0 converted instead of the required ADC channel

Description

The following ADC conversions unduly convert the ADC internal channel 0 instead of the required channel, and as a consequence, they return zero as conversion result:

1. an injected conversion triggered while regular conversion is on-going
2. the master ADC first injected conversion and the slave ADC first regular conversion (after resume) when dual simultaneous or interleaved dual ADC is used
3. the first regular/injected conversion after a regular or injected software STOP (setting ADSTP or JADSTP bit respectively)
4. the first regular or injected conversion following the DMA end of transfer, after the completion of a regular sequence read by the DMA in one-shot mode

Workaround

Apply one of the following measures:

- Depending on the case, insert a dummy conversion:
  - at the beginning of the injected sequence (not applicable in injected discontinuous mode, JDISCEN = 1) in the case 1
  - after the ADC slave resumes regular conversions in dual simultaneous or interleaved dual ADC mode (not applicable in injected discontinuous mode, JDISCEN = 1) in the case 2
  - after stopping the ADC by software in the case 3
  - after DMA end-of-transfer in the case 4
- Avoid collisions between injected and regular conversions by using triggered regular conversions or by launching regular conversion at the end of injected sequence in the cases 1 and 2.
- After a regular or injected software STOP, disable the ADC with ADDIS = 1 and enable it again with ADEN = 1. This introduces a hardware dummy conversion (the cases 3 and 4).

2.6 COMP

2.6.1 Comparator previously not fully tested in production

Description

The comparator is not fully tested. As a consequence, the V_{offset} parameter may be out of specification. The following specification substitutes the one in the datasheet for these parts:

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{offset}	Comparator offset error	Full V_{DDA} voltage range, full temperature range	-20	-15/+2	3	mV

Workaround

None.

2.7 TIM

2.7.1 One-pulse mode trigger not detected in master-slave reset + trigger configuration

Description

The failure occurs when several timers configured in one-pulse mode are cascaded, and the master timer is configured in combined reset + trigger mode with the MSM bit set:

OPM = 1 in TIMx_CR1, SMS[3:0] = 1000 and MSM = 1 in TIMx_SMCR.

The MSM delays the reaction of the master timer to the trigger event, so as to have the slave timers cycle-accurately synchronized.

If the trigger arrives when the counter value is equal to the period value set in the TIMx_ARR register, the one-pulse mode of the master timer does not work and no pulse is generated on the output.

Workaround

None. However, unless a cycle-level synchronization is mandatory, it is advised to keep the MSM bit reset, in which case the problem is not present. The MSM = 0 configuration also allows decreasing the timer latency to external trigger events.

2.7.2 Consecutive compare event missed in specific conditions

Description

Every match of the counter (CNT) value with the compare register (CCR) value is expected to trigger a compare event. However, if such matches occur in two consecutive counter clock cycles (as consequence of the CCR value change between the two cycles), the second compare event is missed for the following CCR value changes:

- in edge-aligned mode, from ARR to 0:
  - first compare event: CNT = CCR = ARR
  - second (missed) compare event: CNT = CCR = 0
- in center-aligned mode while up-counting, from ARR-1 to ARR (possibly a new ARR value if the period is also changed) at the crest (that is, when TIMx_RCR = 0):
  - first compare event: CNT = CCR = (ARR-1)
  - second (missed) compare event: CNT = CCR = ARR
- in center-aligned mode while down-counting, from 1 to 0 at the valley (that is, when TIMx_RCR = 0):
  - first compare event: CNT = CCR = 1
  - second (missed) compare event: CNT = CCR = 0

This typically corresponds to an abrupt change of compare value aiming at creating a timer clock single-cycle-wide pulse in toggle mode.

As a consequence:

- In toggle mode, the output only toggles once per counter period (squared waveform), whereas it is expected to toggle twice within two consecutive counter cycles (and so exhibit a short pulse per counter period).
- In center mode, the compare interrupt flag does note rise and the interrupt is not generated.

Note: The timer output operates as expected in modes other than the toggle mode.

Workaround

None.

2.7.3 Compare event missed in center-aligned mode 1 and 2 with dithering enabled

Description

With dithering enabled, the compare event is not generated and the output in toggle mode is incorrect in:

- center-aligned mode 1, with the CCR value between ARR-16 and ARR

- center-aligned mode 2, with the CCR value lower than 16

Note: These CCR values correspond to a waveform with extreme duty cycles: close to 0%, with the pulse width being one timer clock cycle, or close to 100%, with the pulse width being the PWM period minus one timer clock cycle.

The timer output operates as expected if configured in other than toggle mode.

The center-aligned mode 3 is exempt of this failure.

Workaround

None.

2.7.4 Output compare clear not working with external counter reset

Description

The output compare clear event (ocref_clr) is not correctly generated when the timer is configured in the following slave modes: Reset mode, Combined reset + trigger mode, and Combined gated + reset mode.

The PWM output remains inactive during one extra PWM cycle if the following sequence occurs:

1. The output is cleared by the ocref_clr event.
2. The timer reset occurs before the programmed compare event.

Workaround

Apply one of the following measures:

- Use BKIN (or BKIN2 if available) input for clearing the output, selecting the Automatic output enable mode (AOE = 1).
- Mask the timer reset during the PWM ON time to prevent it from occurring before the compare event (for example with a spare timer compare channel open-drain output connected with the reset signal, pulling the timer reset line down).

2.8 LPTIM

2.8.1 Device may remain stuck in LPTIM interrupt when entering Stop mode

Description

This limitation occurs when disabling the low-power timer (LPTIM).

When the user application clears the ENABLE bit in the LPTIM_CR register within a small time window around one LPTIM interrupt occurrence, then the LPTIM interrupt signal used to wake up the device from Stop mode may be frozen in active state. Consequently, when trying to enter Stop mode, this limitation prevents the device from entering low-power mode and the firmware remains stuck in the LPTIM interrupt routine.

This limitation applies to all Stop modes and to all instances of the LPTIM. Note that the occurrence of this issue is very low.

Workaround

In order to disable a low power timer (LPTIMx) peripheral, do not clear its ENABLE bit in its respective LPTIM_CR register. Instead, reset the whole LPTIMx peripheral via the RCC controller by setting and resetting its respective LPTIMxRST bit in RCC_APByRSTRz register.

2.8.2 Device may remain stuck in LPTIM interrupt when clearing event flag

Description

This limitation occurs when the LPTIM is configured in interrupt mode (at least one interrupt is enabled) and the software clears any flag in LPTIM_ISR register by writing its corresponding bit in LPTIM_ICR register. If the interrupt status flag corresponding to a disabled interrupt is cleared simultaneously with a new event detection, the set and clear commands might reach the APB domain at the same time, leading to an asynchronous interrupt signal permanently stuck high.

This issue can occur either during an interrupt subroutine execution (where the flag clearing is usually done), or outside an interrupt subroutine.

Consequently, the firmware remains stuck in the LPTIM interrupt routine, and the device cannot enter Stop mode.

Workaround

To avoid this issue, it is strongly advised to follow the recommendations listed below:

- Clear the flag only when its corresponding interrupt is enabled in the interrupt enable register.
- If for specific reasons, it is required to clear some flags that have corresponding interrupt lines disabled in the interrupt enable register, it is recommended to clear them during the current subroutine prior to those which have corresponding interrupt line enabled in the interrupt enable register.
- Flags must not be cleared outside the interrupt subroutine.

Note: The proper clear sequence is already implemented in the HAL_LPTIM_IRQHandler in the STM32Cube.

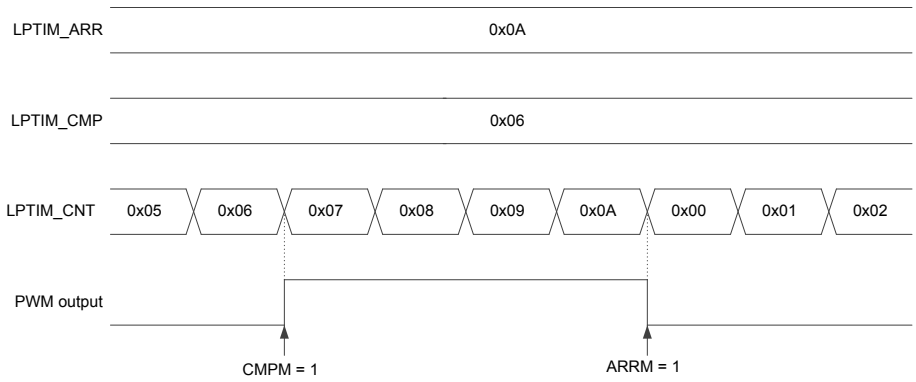
2.8.3 LPTIM events and PWM output are delayed by 1 kernel clock cycle

Description

The compare match event (CMPM), auto reload match event (ARRM), PWM output level and interrupts are updated with a delay of one kernel clock cycle.

Consequently, it is not possible to generate PWM with a duty cycle of 0% or 100%.

The following waveform gives the example of PWM output mode and the effect of the delay:



Workaround

Set the compare value to the desired value minus 1. For instance in order to generate a compare match when LPTM_CNT = 0x08, set the compare value to 0x07.

2.9 RTC and TAMP

2.9.1 Calendar initialization may fail in case of consecutive INIT mode entry

Description

If the INIT bit of the RTC_ICSR register is set between one and two RTCCLK cycles after being cleared, the INITF flag is set immediately instead of waiting for synchronization delay (which should be between one and two RTCCLK cycles), and the initialization of registers may fail. Depending on the INIT bit clearing and setting instants versus the RTCCLK edges, it can happen that, after being immediately set, the INITF flag is cleared during one RTCCLK period then set again. As writes to calendar registers are ignored when INITF is low, a write occurring during this critical period might result in the corruption of one or more calendar registers.

Workaround

After existing the initialization mode, clear the BYPSHAD bit (if set) then wait for RSF to rise, before entering the initialization mode again.

Note: It is recommended to write all registers in a single initialization session to avoid accumulating synchronization delays.

2.9.2 Alarm flag may be repeatedly set when the core is stopped in debug

Description

When the core is stopped in debug mode, the clock is supplied to subsecond RTC alarm downcounter even though the device is configured to stop the RTC in debug.

As a consequence, when the subsecond counter is used for alarm condition (the MASKSS[3:0] bitfield of the RTC_ALRMASSR and/or RTC_ALRMBSSR register set to a non-zero value) and the alarm condition is met just before entering a breakpoint or printf, the ALRAF and/or ALRBF flag of the RTC_SR register is repeatedly set by hardware during the breakpoint or printf, which makes any tentative to clear the flag(s) ineffective.

Workaround

None.

2.9.3 A tamper event fails to trigger timestamp or timestamp overflow events during a few cycles after clearing TSF

Description

With the timestamp on tamper event enabled (TAMPTS bit of the RTC_CR register set), a tamper event is ignored if it occurs:

- within four APB clock cycles after setting the CTSF bit of the RTC_SCR register to clear the TSF flag, while the TSF flag is not yet effectively cleared (it fails to set the TSOVF flag)
- within two ck_apre cycles after setting the CTSF bit of the RTC_SCR register to clear the TSF flag, when the TSF flag is effectively cleared (it fails to set the TSF flag and timestamp the calendar registers)

Workaround

None.

2.9.4 REFCKON write protection associated to INIT KEY instead of CAL KEY

Description

The write protection of the REFCKON bit is unlocked if the key sequence is written in RTC_WPR with the privilege and security rights set by the INITPRIV and INITSEC bits, instead of being set by the CALPRIV and CALSEC bits.

Workaround

Unlock the INIT KEY before writing REFCKON.

2.9.5 Tamper flag not set on LSE failure detection

Description

With the timestamp on tamper event enabled (the TAMPTS bit of the RTC_CR register set), the LSE failure detection (LSE clock stopped) event connected to the internal tamper 3 fails to raise the ITAMP3F and ITAMP3MF flags, although it duly erases or blocks (depending on the internal tamper 3 configuration) the backup registers and other device secrets, and the RTC and TAMP peripherals resume normally upon the LSE restart.

Note: As expected in this particular case, the TSF and TSMF flags remain low as long as LSE is stopped as they require running RTCCLK clock to operate.

Workaround

None.

2.10 I2C

2.10.1 Wrong data sampling when data setup time ($t_{SU;DAT}$) is shorter than one I2C kernel clock period

Description

The I^2C-bus specification and user manual specify a minimum data setup time ($t_{SU;DAT}$) as:

- 250 ns in Standard mode
- 100 ns in Fast mode
- 50 ns in Fast mode Plus

The device does not correctly sample the I^2C-bus SDA line when $t_{SU;DAT}$ is smaller than one I2C kernel clock (I^2C-bus peripheral clock) period: the previous SDA value is sampled instead of the current one. This can result in a wrong receipt of slave address, data byte, or acknowledge bit.

Workaround

Increase the I2C kernel clock frequency to get I2C kernel clock period within the transmitter minimum data setup time. Alternatively, increase transmitter's minimum data setup time. If the transmitter setup time minimum value corresponds to the minimum value provided in the I^2C-bus standard, the minimum I2CCLK frequencies are as follows:

- In Standard mode, if the transmitter minimum setup time is 250 ns, the I2CCLK frequency must be at least 4 MHz.
- In Fast mode, if the transmitter minimum setup time is 100 ns, the I2CCLK frequency must be at least 10 MHz.
- In Fast-mode Plus, if the transmitter minimum setup time is 50 ns, the I2CCLK frequency must be at least 20 MHz.

2.10.2 Spurious bus error detection in master mode

Description

In master mode, a bus error can be detected spuriously, with the consequence of setting the BERR flag of the I2C_SR register and generating bus error interrupt if such interrupt is enabled. Detection of bus error has no effect on the I^2C-bus transfer in master mode and any such transfer continues normally.

Workaround

If a bus error interrupt is generated in master mode, the BERR flag must be cleared by software. No other action is required and the ongoing transfer can be handled normally.

2.10.3 Spurious master transfer upon own slave address match

Description

When the device is configured to operate at the same time as master and slave (in a multi- master I^2C-bus application), a spurious master transfer may occur under the following condition:

- Another master on the bus is in process of sending the slave address of the device (the bus is busy).
- The device initiates a master transfer by bit set before the slave address match event (the ADDR flag set in the I2C_ISR register) occurs.
- After the ADDR flag is set:
  - the device does not write I2C_CR2 before clearing the ADDR flag, or
  - the device writes I2C_CR2 earlier than three I2C kernel clock cycles before clearing the ADDR flag

In these circumstances, even though the START bit is automatically cleared by the circuitry handling the ADDR flag, the device spuriously proceeds to the master transfer as soon as the bus becomes free. The transfer configuration depends on the content of the I2C_CR2 register when the master transfer starts. Moreover, if the I2C_CR2 is written less than three kernel clocks before the ADDR flag is cleared, the I2C peripheral may fall into an unpredictable state.

Workaround

Upon the address match event (ADDR flag set), apply the following sequence.

Normal mode (SBC = 0):

1. Set the ADDRCF bit.
2. Before Stop condition occurs on the bus, write I2C_CR2 with the START bit low.

Slave byte control mode (SBC = 1):

1. Write I2C_CR2 with the slave transfer configuration and the START bit low.
2. Wait for longer than three I2C kernel clock cycles.
3. Set the ADDRCF bit.
4. Before Stop condition occurs on the bus, write I2C_CR2 again with its current value.

The time for the software application to write the I2C_CR2 register before the Stop condition is limited, as the clock stretching (if enabled), is aborted when clearing the ADDR flag.

Polling the BUSY flag before requesting the master transfer is not a reliable workaround as the bus may become busy between the BUSY flag check and the write into the I2C_CR2 register with the START bit set.

2.10.4 OVR flag not set in underrun condition

Description

In slave transmission with clock stretching disabled (NOSTRETCH = 1 in the I2C_CR1 register), an underrun condition occurs if the current byte transmission is completed on the I2C bus, and the next data is not yet written in the TXDATA[7:0] bitfield. In this condition, the device is expected to set the OVR flag of the I2C_ISR register and send 0xFF on the bus.

However, if the I2C_TXDR is written within the interval between two I2C kernel clock cycles before and three APB clock cycles after the start of the next data transmission, the OVR flag is not set, although the transmitted value is 0xFF.

Workaround

None.

2.10.5 Transmission stalled after first byte transfer

Description

When the first byte to transmit is not prepared in the TXDATA register, two bytes are required successively, through TXIS status flag setting or through a DMA request. If the first of the two bytes is written in the I2C_TXDR register in less than two I2C kernel clock cycles after the TXIS/DMA request, and the ratio between APB clock and I2C kernel clock frequencies is between 1.5 and 3, the second byte written in the I2C_TXDR is not internally detected. This causes a state in which the I2C peripheral is stalled in master mode or in slave mode, with clock stretching enabled (NOSTRETCH = 0). This state can only be released by disabling the peripheral (PE = 0) or by resetting it.

Workaround

Apply one of the following measures:

- Write the first data in I2C_TXDR before the transmission starts.
- Set the APB clock frequency so that its ratio with respect to the I2C kernel clock frequency is lower than 1.5 or higher than 3.

2.11 USART

2.11.1 Anticipated end-of-transmission signaling in SPI slave mode

Description

In SPI slave mode, at low USART baud rate with respect to the USART kernel and APB clock frequencies, the *transmission complete* flag TC of the USARTx_ISR register may unduly be set before the last bit is shifted on the transmit line.

This leads to data corruption if, based on this anticipated end-of-transmission signaling, the application disables the peripheral before the last bit is transmitted.

Workaround

Upon the TC flag rise, wait until the clock line remains idle for more than the half of the communication clock cycle. Then only consider the transmission as ended.

2.11.2 Data corruption due to noisy receive line

Description

In UART mode with oversampling by 8 or 16 and with 1 or 2 stop bits, the received data may be corrupted if a glitch to zero shorter than the half-bit occurs on the receive line within the second half of the stop bit.

Workaround

None.

2.12 SPI

2.12.1 BSY bit may stay high when SPI is disabled

Description

The BSY flag may remain high upon disabling the SPI while operating in:

- master transmit mode and the TXE flag is low (data register full).
- master receive-only mode (simplex receive or half-duplex bidirectional receive phase) and an SCK strobing edge has not occurred since the transition of the RXNE flag from low to high.
- slave mode and NSS signal is removed during the communication.

Workaround

When the SPI operates in:

- master transmit mode, disable the SPI when TXE = 1 and BSY = 0.
- master receive-only mode, ignore the BSY flag.
- slave mode, do not remove the NSS signal during the communication.

2.12.2 BSY bit may stay high at the end of data transfer in slave mode

Description

BSY flag may sporadically remain high at the end of a data transfer in slave mode. This occurs upon coincidence of internal CPU clock and external SCK clock provided by master.

In such an event, if the software only relies on BSY flag to detect the end of SPI slave data transaction (for example to enter low-power mode or to change data line direction in half-duplex bidirectional mode), the detection fails.

As a conclusion, the BSY flag is unreliable for detecting the end of data transactions.

Workaround

Depending on SPI operating mode, use the following means for detecting the end of transaction:

- When NSS hardware management is applied and NSS signal is provided by master, use NSS flag.
- In SPI receiving mode, use the corresponding RXNE event flag.
- In SPI transmit-only mode, use the BSY flag in conjunction with a timeout expiry event. Set the timeout such as to exceed the expected duration of the last data frame and start it upon TXE event that occurs with the second bit of the last data frame. The end of the transaction corresponds to either the BSY flag becoming low or the timeout expiry, whichever happens first.

Prefer one of the first two measures to the third as they are simpler and less constraining.

Alternatively, apply the following sequence to ensure reliable operation of the BSY flag in SPI transmit mode:

1. Write last data to data register.
2. Poll the TXE flag until it becomes high, which occurs with the second bit of the data frame transfer.
3. Disable SPI by clearing the SPE bit mandatorily before the end of the frame transfer.
4. Poll the BSY bit until it becomes low, which signals the end of transfer.

Note: The alternative method can only be used with relatively fast CPU speeds versus relatively slow SPI clocks or/and long last data frames. The faster is the software execution, the shorter can be the duration of the last data frame.

2.13 FDCAN

2.13.1 Desynchronization under specific condition with edge filtering enabled

Description

FDCAN may desynchronize and incorrectly receive the first bit of the frame if:

- the edge filtering is enabled (the EFBI bit of the FDCAN_CCCR register is set), and
- the end of the integration phase coincides with a falling edge detected on the FDCAN_Rx input pin

If this occurs, the CRC detects that the first bit of the received frame is incorrect, flags the received frame as faulty and responds with an error frame.

Note: This issue does not affect the reception of standard frames.

Workaround

Disable edge filtering or wait for frame retransmission.

2.13.2 Tx FIFO messages inverted under specific buffer usage and priority setting

Description

Two consecutive messages from the Tx FIFO may be inverted in the transmit sequence if:

- FDCAN uses both a dedicated Tx buffer and a Tx FIFO (the TFQM bit of the FDCAN_TXBC register is cleared), and
- the messages contained in the Tx buffer have a higher internal CAN priority than the messages in the Tx FIFO.

Workaround

Apply one of the following measures:

- Ensure that only one Tx FIFO element is pending for transmission at any time:

  The Tx FIFO elements may be filled at any time with messages to be transmitted, but their transmission requests are handled separately. Each time a Tx FIFO transmission has completed and the Tx FIFO gets empty (TFE bit of FDACN_IR set to 1) the next Tx FIFO element is requested.
- Use only a Tx FIFO:

  Send both messages from a Tx FIFO, including the message with the higher priority. This message has to wait until the preceding messages in the Tx FIFO have been sent.
- Use two dedicated Tx buffers (for example, use Tx buffer 4 and 5 instead of the Tx FIFO). The following pseudo-code replaces the function in charge of filling the Tx FIFO:

```
Write message to Tx Buffer 4
Transmit Loop:
    Request Tx Buffer 4 - write AR4 bit in FDCAN_TXBAR
    Write message to Tx Buffer 5
    Wait until transmission of Tx Buffer 4 complete (IR bit in FDCAN_IR),
    read TO4 bit in FDCAN_TXBTO
    Request Tx Buffer 5 - write AR5 bit of FDCAN_TXBAR
    Write message to Tx Buffer 4
    Wait until transmission of Tx Buffer 5 complete (IR bit in FDCAN_IR),
    read TO5 bit in FDCAN_TXBTO
```

2.14 USB

2.14.1 ESOF interrupt timing desynchronized after resume signaling

Description

Upon signaling resume, the device is expected to allow full 3 ms of time to the host or hub for sending the initial SOF (start of frame) packet, without triggering SUSP interrupt. However, the device only allows two full milliseconds and unduly triggers SUSP interrupt if it receives the initial packet within the third millisecond.

Workaround

When the device initiates resume (remote wakeup), mask the SUSP interrupt by setting the SUSPM bit for 3 ms, then unmask it by clearing SUSPM.

2.14.2 Incorrect CRC16 in the memory buffer

Description

Memory buffer locations are written starting from the address contained in the ADDRn_RX for a number of bytes corresponding to the received data packet length, CRC16 inclusive (that is, data payload length plus two bytes), or up to the last allocated memory location defined by BL_SIZE and NUM_BLOCK, whichever comes first. In the former case, the CRC16 checksum is written wrongly, with its least significant byte going to both memory buffer byte locations expected to receive the least and the most significant bytes of the checksum.

Although the checksum written in the memory buffer is wrong, the underlying CRC checking mechanism in the USB peripheral is fully functional.

Workaround

Ignore the CRC16 data in the memory buffer.

Revision history

Table 4. Document revision history

Date	Version	Changes
19-Apr-2019	1	Initial release.
17-Dec-2019	2	Added in Section 2 Description of device errata and in Table 3. Summary of device limitations: • erratum FLASH_ECCR corrupted upon reset or power-down occurring during Flash memory program or erase operation • erratum OVR flag not set in underrun condition • Transmission stalled after first byte transfer • silicon revision Y Removed: • erratum *First double-word of Flash memory corrupted upon reset or power-down while programming* • erratum *Data cache might be corrupted during Flash memory read-while-write operation*
22-Jun-2020	3	Added in Section 2 Description of device errata and in Table 3. Summary of device limitations: • erratum ADC channel 0 converted instead of the required ADC channel • erratum Consecutive compare event missed in specific conditions • erratum Compare event missed in center-aligned mode 1 and 2 with dithering enabled • erratum Output compare clear not working with external counter reset • erratum A tamper event fails to trigger timestamp or timestamp overflow events during a few cycles after clearing TSF • erratum REFCKON write protection associated to INIT KEY instead of CAL KEY • erratum Tamper flag not set on LSE failure detection • erratum Anticipated end-of-transmission signaling in SPI slave mode • erratum Data corruption due to noisy receive line • erratum ESOF interrupt timing desynchronized after resume signaling • erratum Incorrect CRC16 in the memory buffer • erratum Desynchronization under specific condition with edge filtering enabled • erratum Tx FIFO messages inverted under specific buffer usage and priority setting
28-Jul-2020	4	Added in Section 2 Description of device errata and in Table 3. Summary of device limitations: • erratum Comparator previously not fully tested in production • erratum VREF+ output voltage disturbed by some GPIOs toggling
31-Aug-2020	5	Added in Section 2 Description of device errata and in Table 3. Summary of device limitations: PCROP_RDP option bit production value not inline with the specification Removed : VREF+ output voltage disturbed by some GPIOs toggling
05-Nov-2020	6	Updated: • Document's scope and all tables in the document to add revision X information. • erratum DMA disable failure and error flag omission upon simultaneous transfer error and global flag clear

Date	Version	Changes
		Added: • erratum New context conversion initiated without waiting for trigger when writing new context in ADC_JSQR with JQDIS = 0 and JQM = 0 • erratum Two consecutive context conversions fail when writing new context in ADC_JSQR just after previous context completion with JQDIS = 0 and JQM = 0 • erratum Unexpected regular conversion when two consecutive injected conversions are performed in Dual interleaved mode • erratum ADC_AWDy_OUT reset by non-guarded channels • erratum LPTIM events and PWM output are delayed by 1 kernel clock cycle

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