

## STM32L412xx device errata

## Applicability

This document applies to the part numbers of STM32L412xx devices and the device variants as stated in this page.

It gives a summary and a description of the device errata, with respect to the device datasheet and reference manual RM394.

Deviation of the real device behavior from the intended device behavior is considered to be a device limitation. Deviation of the description in the reference manual or the datasheet from the intended device behavior is considered to be a documentation erratum. The term “*errata*” applies both to limitations and documentation errata.

**Table 1. Device summary**

Reference	Part numbers
STM32L412xx	STM32L412C8, STM32L412CB, STM32L412K8, STM32L412KB, STM32L412R8, STM32L412RB, STM32L412T8, STM32L412TB

**Table 2. Device variants**

Reference	Silicon revision codes	
	Device marking <sup>(1)</sup>	REV_ID <sup>(2)</sup>
STM32L412xx	A	0x1000

1. Refer to the device datasheet for how to identify this code on different types of package.

2. REV\_ID[15:0] bitfield of DBGMCU\_IDCODE register.

# 1 Summary of device errata

The following table gives a quick reference to the STM32L412xx device limitations and their status:

A = workaround available

N = no workaround available

P = partial workaround available

Applicability of a workaround may depend on specific conditions of target application. Adoption of a workaround may cause restrictions to target application. Workaround for a limitation is deemed partial if it only reduces the rate of occurrence and/or consequences of the limitation, or if it is fully effective for only a subset of instances on the device or in only a subset of operating modes, of the function concerned.

**Table 3. Summary of device limitations**

Function	Section	Limitation	Status
			Rev. A
Core	2.1.1	Interrupted loads to SP can cause erroneous behavior	A
	2.1.2	VDIV or VSQRT instructions might not complete correctly when very short ISRs are used	A
	2.1.3	Store immediate overlapping exception return operation might vector to incorrect interrupt	A
System	2.2.1	Internal voltage reference corrupted upon Stop mode entry with temperature sensing enabled	A
	2.2.2	Full JTAG configuration without NJTRST pin cannot be used	A
	2.2.3	Current injection from VDD to VDDA through analog switch voltage booster	A
	2.2.4	Unstable LSI when it clocks RTC or CSS on LSE	P
	2.2.5	LSESYSDIS has no effects if set just before Stop entry	A
	2.2.6	PB8 and PB11 on 48-pin devices with SMPS remain software configurable	A
	2.2.7	FLASH_ECCR corrupted upon reset or power-down occurring during Flash memory program or erase operation	A
	2.2.9	HSE oscillator long startup at low voltage	P
FW	2.3.1	Code segment unprotected if non-volatile data segment length is zero	A
QUADSPI	2.4.1	First nibble of data not written after dummy phase	A
	2.4.2	Wrong data from memory-mapped read after an indirect mode operation	A
ADC	2.5.1	Writing ADCx_JSQR when JADCSTART and JQDIS are set might lead to incorrect behavior	N
	2.5.2	Wrong ADC result if conversion done late after calibration or previous conversion	A
	2.5.3	Spurious temperature measurement due to spike noise	A
	2.5.4	Selected external ADC inputs unduly clamped to VDD when all analog peripherals are disabled	A
TSC	2.6.2	TSC signal-to-noise concern under specific conditions	A
LPTIM	2.7.1	Device may remain stuck in LPTIM interrupt when entering Stop mode	A
	2.7.2	Device may remain stuck in LPTIM interrupt when clearing event flag	P
RTC and TAMP	2.8.1	Internal tamper flags not output on RTC_OUT1 and RTC_OUT2	N
	2.8.2	Notification of illegal access to secured registers is not reliable	N
	2.8.3	RTC_MISR and TAMP_MISR can be read by non-privileged accesses when privilege-protected	N
	2.8.4	RTC configuration changes ignored at specific conditions	A

Function	Section	Limitation	Status
			Rev. A
RTC and TAMP	2.8.5	Calibration formula changes when LPCAL is set	A
	2.8.6	Calendar initialization may fail in case of consecutive INIT mode entry	A
	2.8.7	Spurious RTC alarm EXTI line event following RTC WUT interrupt	A
	2.8.8	RTC_REFIN and RTC_OUT on PB2 not operating in Stop 2 mode	P
I2C	2.9.1	10-bit master mode: new transfer cannot be launched if first part of the address is not acknowledged by the slave	A
	2.9.3	Wrong data sampling when data setup time (t <sub>SU</sub> ;DAT) is shorter than one I2C kernel clock period	P
	2.9.4	Spurious bus error detection in master mode	A
	2.9.5	Last-received byte loss in reload mode	P
	2.9.6	Spurious master transfer upon own slave address match	P
USART	2.10.1	RTS is active while RE = 0 or UE = 0	A
SPI	2.11.1	BSY bit may stay high at the end of data transfer in slave mode	A
	2.11.2	Wrong CRC in full-duplex mode handled by DMA with imbalanced setting of data counters	A
	2.11.4	SPI master communication failure at high fPCLK within the specified range	N

The following table gives a quick reference to the documentation errata.

**Table 4. Summary of device documentation errata**

Function	Section	Documentation erratum
TSC	2.6.1	Inhibited acquisition in short transfer phase configuration
RTC and TAMP	2.8.9	Setting GPIO properties of PC13 used as RTC_ALARM open-drain output
I2C	2.9.2	Wrong behavior in Stop mode when wakeup from Stop mode is disabled in I2C
SPI	2.11.3	CRC error in SPI slave mode if internal NSS changes before CRC transfer

## 2 Description of device errata

The following sections describe limitations of the applicable devices with Arm® core and provide workarounds if available. They are grouped by device functions.

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### 2.1 Core

Reference manual and errata notice for the Arm® Cortex®-M4F core revision r0p1 is available from <http://infocenter.arm.com>.

#### 2.1.1 Interrupted loads to SP can cause erroneous behavior

This limitation is registered under Arm ID number 752770 and classified into “Category B”. Its impact to the device is minor.

##### Description

If an interrupt occurs during the data-phase of a single word load to the stack-pointer (SP/R13), erroneous behavior can occur. In all cases, returning from the interrupt will result in the load instruction being executed an additional time. For all instructions performing an update to the base register, the base register will be erroneously updated on each execution, resulting in the stack-pointer being loaded from an incorrect memory location.

The affected instructions that can result in the load transaction being repeated are:

- LDR SP, [Rn],#imm
- LDR SP, [Rn,#imm]!
- LDR SP, [Rn,#imm]
- LDR SP, [Rn]
- LDR SP, [Rn,Rm]

The affected instructions that can result in the stack-pointer being loaded from an incorrect memory address are:

- LDR SP,[Rn],#imm
- LDR SP,[Rn,#imm]!

As compilers do not generate these particular instructions, the limitation is only likely to occur with hand-written assembly code.

##### Workaround

Both issues may be worked around by replacing the direct load to the stack-pointer, with an intermediate load to a general-purpose register followed by a move to the stack-pointer.

#### 2.1.2 VDIV or VSQRT instructions might not complete correctly when very short ISRs are used

This limitation is registered under Arm ID number 776924 and classified into “Category B”. Its impact to the device is limited.

##### Description

The VDIV and VSQRT instructions take 14 cycles to execute. When an interrupt is taken a VDIV or VSQRT instruction is not terminated, and completes its execution while the interrupt stacking occurs. If lazy context save of floating point state is enabled then the automatic stacking of the floating point context does not occur until a floating point instruction is executed inside the interrupt service routine.

Lazy context save is enabled by default. When it is enabled, the minimum time for the first instruction in the interrupt service routine to start executing is 12 cycles. In certain timing conditions, and if there is only one or two instructions inside the interrupt service routine, then the VDIV or VSQRT instruction might not write its result to the register bank or to the FPSCR.

The failure occurs when the following condition is met:

1. The floating point unit is enabled
2. Lazy context saving is not disabled
3. A VDIV or VSQRT is executed
4. The destination register for the VDIV or VSQRT is one of s0 - s15
5. An interrupt occurs and is taken
6. The interrupt service routine being executed does not contain a floating point instruction
7. Within 14 cycles after the VDIV or VSQRT is executed, an interrupt return is executed

A minimum of 12 of these 14 cycles are utilized for the context state stacking, which leaves 2 cycles for instructions inside the interrupt service routine, or 2 wait states applied to the entire stacking sequence (which means that it is not a constant wait state for every access).

In general, this means that if the memory system inserts wait states for stack transactions (that is, external memory is used for stack data), then this erratum cannot be observed.

The effect of this erratum is that the VDIV or VSQRT instruction does not complete correctly and the register bank and FPSCR are not updated, which means that these registers hold incorrect, out of date, data.

### Workaround

A workaround is only required if the floating point unit is enabled. A workaround is not required if the stack is in external memory.

There are two possible workarounds:

- Disable lazy context save of floating point state by clearing LSPEN to 0 (bit 30 of the FPCCR at address 0xE000EF34).
- Ensure that every interrupt service routine contains more than 2 instructions in addition to the exception return instruction.

## 2.1.3 Store immediate overlapping exception return operation might vector to incorrect interrupt

This limitation is registered under Arm ID number 838869 and classified into “Category B (rare)”. Its impact to the device is minor.

### Description

The core includes a write buffer that permits execution to continue while a store is waiting on the bus. Under specific timing conditions, during an exception return while this buffer is still in use by a store instruction, a late change in selection of the next interrupt to be taken might result in there being a mismatch between the interrupt acknowledged by the interrupt controller and the vector fetched by the processor.

The failure occurs when the following condition is met:

1. The handler for interrupt A is being executed.
2. Interrupt B, of the same or lower priority than interrupt A, is pending.
3. A store with immediate offset instruction is executed to a bufferable location.
  - STR/STRH/STRB <Rt>, [<Rn>,#imm]
  - STR/STRH/STRB <Rt>, [<Rn>,#imm]!
  - STR/STRH/STRB <Rt>, [<Rn>],#imm
4. Any number of additional data-processing instructions can be executed.
5. A BX instruction is executed that causes an exception return.
6. The store data has wait states applied to it such that the data is accepted at least two cycles after the BX is executed.
  - Minimally, this is two cycles if the store and the BX instruction have no additional instructions between them.
  - The number of wait states required to observe this erratum needs to be increased by the number of cycles between the store and the interrupt service routine exit instruction.
7. Before the bus accepts the buffered store data, another interrupt C is asserted which has the same or lower priority as A, but a greater priority than B.

Example:

The processor should execute interrupt handler C, and on completion of handler C should execute the handler for B. If the conditions above are met, then this erratum results in the processor erroneously clearing the pending state of interrupt C, and then executing the handler for B twice. The first time the handler for B is executed it will be at interrupt C's priority level. If interrupt C is pending by a level-based interrupt which is cleared by C's handler then interrupt C will be pending again once the handler for B has completed and the handler for C will be executed.

As the STM32 interrupt C is level based, it eventually becomes pending again and is subsequently handled.

### Workaround

For software not using the memory protection unit, this erratum can be worked around by setting DISDEFWBUF in the Auxiliary Control Register.

In all other cases, the erratum can be avoided by ensuring a DSB occurs between the store and the BX instruction. For exception handlers written in C, this can be achieved by inserting the appropriate set of intrinsics or inline assembly just before the end of the interrupt function, for example:

ARMCC:

```
...
__schedule_barrier();
__asm{DSB};
__schedule_barrier();
}
```

GCC:

```
...
__asm volatile ("dsb 0xf":::"memory");
}
```

## 2.2 System

### 2.2.1 Internal voltage reference corrupted upon Stop mode entry with temperature sensing enabled

#### Description

When entering Stop mode with the temperature sensor channel and the associated ADC(s) enabled, the internal voltage reference may be corrupted.

The occurrence of the corruption depends on the supply voltage and the temperature.

The corruption of the internal voltage reference may cause:

- an overvoltage in  $V_{CORE}$  domain
- an overshoot / undershoot of internal clock (LSI, HSI, MSI) frequencies
- a spurious brown-out reset

The limitation applies to Stop 1 and Stop 2 modes.

#### Workaround

Before entering Stop mode:

- Disable the ADC(s) using the temperature sensor signal as input, and/or
- Disable the temperature sensor channel, by clearing the CH17SEL bit of the ADCx\_CCR register.

Disabling both the ADC(s) and the temperature sensor channel reduces the power consumption during Stop mode.

## 2.2.2 Full JTAG configuration without NJTRST pin cannot be used

### Description

When using the JTAG debug port in Debug mode, the connection with the debugger is lost if the NJTRST pin (PB4) is used as a GPIO or for an alternate function other than NJTRST. Only the 4-wire JTAG port configuration is impacted.

### Workaround

Use the SWD debug port instead of the full 4-wire JTAG port.

## 2.2.3 Current injection from $V_{DD}$ to $V_{DDA}$ through analog switch voltage booster

### Description

With  $V_{DDA}$  below 2.4 V and  $V_{DD}$  above 3 V, a small current injected from VDD line to VDDA line may cause  $V_{DDA}$  to exceed its nominal value.

This current injection only occurs when the I/O analog switch voltage booster is disabled (the BOOSTEN bit of the SYSCFG\_CFGR1 register is cleared) and at least one of the analog peripherals (ADC, COMP, or OPAMP) is enabled.

### Workaround

Enable the I/O analog switch voltage booster, by setting the BOOSTEN bit of the SYSCFG\_CFGR1 register. when  $V_{DDA}$  is below 2.4 V and  $V_{DD}$  is above 3 V.

## 2.2.4 Unstable LSI when it clocks RTC or CSS on LSE

### Description

The LSI clock can become unstable (duty cycle different from 50 %) and its maximum frequency can become significantly higher than 32 kHz, when:

- LSI clocks the RTC, or it clocks the clock security system (CSS) on LSE (which holds when the LSECSSON bit set), and
- the  $V_{DD}$  power domain is reset while the backup domain is not reset, which happens:
  - upon exiting Shutdown mode
  - if  $V_{BAT}$  is separate from  $V_{DD}$  and  $V_{DD}$  goes off then on
  - if  $V_{BAT}$  is tied to  $V_{DD}$  (internally in the package for products not featuring the VBAT pin, or externally) and a short (< 1 ms)  $V_{DD}$  drop under  $V_{DD}(\min)$  occurs

### Workaround

Apply one of the following measures:

- Clock the RTC with LSE or HSE/32, without using the CSS on LSE
- If LSI clocks the RTC or when the LSECSSON bit is set, reset the backup domain upon each  $V_{DD}$  power up (when the BORRSTF flag is set). If  $V_{BAT}$  is separate from  $V_{DD}$ , also restore the RTC configuration, backup registers and anti-tampering configuration.

## 2.2.5 LSESYSDIS has no effects if set just before Stop entry

### Description

The LSE clock disable is only effective two LSE clocks after setting the LSESYSDIS bit of the RCC\_BDCR register. If Stop 2 low-power mode is entered before that instant, the LSE clock is not effectively disabled and its propagation in the device leads to excessive consumption in Stop 2 mode.

**Workaround**

Upon setting the LSESYSDIS bit of the RCC\_BDCR register, wait for at least two LSE clock periods before entering Stop 2 mode.

**2.2.6 PB8 and PB11 on 48-pin devices with SMPS remain software configurable**
**Description**

On 48-pin devices with SMPS support, the non-bonded PB8 and PB11 GPIOs remain software configurable. Their software configuration as floating inputs may lead to increased power consumption.

**Workaround**

Do not configure the PB8 and PB11 GPIOs as floating inputs.

**2.2.7 FLASH\_ECCR corrupted upon reset or power-down occurring during Flash memory program or erase operation**
**Description**

Reset or power-down occurring during a Flash memory location program or erase operation, followed by a read of the same memory location, may lead to a corruption of the FLASH\_ECCR register content.

**Workaround**

Under such condition, erase the page(s) corresponding to the Flash memory location.

**2.2.8 Current injection from V<sub>DD</sub> to V<sub>DDA</sub> through analog switch voltage booster**
**Description**

With V<sub>DDA</sub> below 2.4 V and V<sub>DD</sub> above 3 V, a small current injected from V<sub>DD</sub> line to V<sub>DDA</sub> line may cause V<sub>DDA</sub> to exceed its nominal value.

This current injection only occurs when the I/O analog switch voltage booster is disabled (the BOOSTEN bit of the SYSCFG\_CFGR1 register is cleared) and at least one of the analog peripherals (ADC, COMP, or OPAMP) is enabled.

**Workaround**

Enable the I/O analog switch voltage booster, by setting the BOOSTEN bit of the SYSCFG\_CFGR1 register. when V<sub>DDA</sub> is below 2.4 V and V<sub>DD</sub> is above 3 V.

**2.2.9 HSE oscillator long startup at low voltage**
**Description**

When V<sub>DD</sub> is below 2.7 V, the HSE oscillator may take longer than specified to start up. Several hundred milliseconds might elapse before the HSERDY flag in the RCC\_CR register is set.

**Workaround**

The following sequence is recommended:

1. Configure PH0 and PH1 as standard GPIOs in output mode and low-level state.
2. Enable the HSE oscillator.



## 2.3 FW

### 2.3.1 Code segment unprotected if non-volatile data segment length is zero

#### Description

If during FW configuration the length of firewall-protected non-volatile data segment is set to zero through the LENG[21:8] bitfield of the FW\_NVDSL register, the firewall protection of code segment does not operate.

#### Workaround

Always set the LENG[21:8] bitfield of the FW\_NVDSL register to a non-zero value, even if no firewall protection of data in the non-volatile data segment is required.

## 2.4 QUADSPI

### 2.4.1 First nibble of data not written after dummy phase

#### Description

The first nibble of data to be written to the external Flash memory is lost when the following condition is met:

- QUADSPI is used in indirect write mode.
- At least one dummy cycle is used.

#### Workaround

Use alternate bytes instead of dummy phase to add latency between the address phase and the data phase. This works only if the number of dummy cycles to substitute corresponds to a multiple of eight bits of data.

Example:

- To substitute one dummy cycle, send one alternate byte (only possible in DDR mode with four data lines).
- To substitute two dummy cycles, send one alternate byte in SDR mode with four data lines.
- To substitute four dummy cycles, send two alternate bytes in SDR mode with four data lines, or one alternate byte in SDR mode with two data lines.
- To substitute eight dummy cycles, send one alternate byte in SDR mode with one data line.

### 2.4.2 Wrong data from memory-mapped read after an indirect mode operation

#### Description

The first memory-mapped read in indirect mode can yield wrong data if the QUADSPI peripheral enters memory-mapped mode with bits ADDRESS[1:0] of the QUADSPI\_AR register both set.

#### Workaround

Before entering memory-mapped mode, apply the following measure, depending on access mode:

- Indirect read mode: clear the QUADSPI\_AR register then issue an abort request to stop reading and to clear the BUSY bit.
- Indirect write mode: clear the QUADSPI\_AR register.

**Caution:** The QUADSPI\_DR register must not be written after clearing the QUADSPI\_AR register.

## 2.5 ADC

### 2.5.1 Writing ADCx\_JSQR when JADCSTART and JQDIS are set might lead to incorrect behavior

#### Description

Writing the ADCx\_JSQR register when there is an on-going injected conversion (JADCSTART = 1) might lead to unpredictable ADC behavior if the queues of context are not enabled (JQDIS = 1).

**Workaround**

None.

**2.5.2 Wrong ADC result if conversion done late after calibration or previous conversion**
**Description**

The result of an ADC conversion done more than 1 ms later than the previous ADC conversion or ADC calibration might be incorrect.

**Workaround**

Perform two consecutive ADC conversions in single, scan or continuous mode. Reject the result of the first conversion and only keep the result of the second.

**2.5.3 Spurious temperature measurement due to spike noise**
**Description**

Depending on the MCU activity, internal interference may cause temperature-dependent spike noise on the temperature sensor output to the ADC, resulting in occasional spurious (outlying) temperature measurement.

**Workaround**

Perform a series of measurements and process the acquired data samples such as to obtain a mean value not affected by the outlying samples.

For this, it is recommended to use interquartile mean (IQM) algorithm with at least 64 samples. IQM is based on rejecting the quarters (quartiles) of sample population with the lowest and highest values and on computing the mean value only using the remaining (interquartile) samples.

The acquired sample values are first sorted from lowest to highest, then the sample sequence is truncated by removing the lowest and highest sample quartiles.

Example:

**Table 5. Measurement result after IQM post-processing**

Data	Sample												Mean
	1	2	3	4	5	6	7	8	9	10	11	12	
Acquired	17.2	10.92	9.56	2.12	9.82	10.72	10.6	3.5	9.46	9.78	9.5	1.1	8.69
Sorted	1.1	2.12	3.5	9.46	9.5	9.56	9.78	9.82	10.6	10.72	10.92	17.2	8.69
Truncated	-	-	-	9.46	9.5	9.56	9.78	9.82	10.6	-	-	-	9.79

The measurement result after the IQM post-processing in the example is 9.79. For consistent results, use a minimum of 64 samples. It is recommended to optimize the code performing the sort task such as to minimize its processing power requirements.

**2.5.4 Selected external ADC inputs unduly clamped to V<sub>DD</sub> when all analog peripherals are disabled**
**Description**

When all analog peripherals are disabled, the GPIO(s) selected as ADC input(s) are unduly clamped (through a parasitic diode) to V<sub>DD</sub> instead to V<sub>DDA</sub>. As a consequence, the input voltage is limited to V<sub>DD</sub> + 0.3 V even if V<sub>DDA</sub> is higher than V<sub>DD</sub> + 0.3 V.

*Note:* The selection of GPIOs as ADC inputs is done with the SQy and JSQy bitfields of the ADC\_SQRx and ADC\_JSQR registers, respectively.

### Workaround

Apply one of the following measures:

- Use  $V_{DDA}$  lower than  $V_{DD} + 0.3$  V.
- Keep at least one analog peripheral enabled if GPIOs are selected as ADC inputs.
- Deselect GPIOs as ADC inputs (by clearing `ADC_SQRx` or/and `ADC_JSQR` registers) when no analog peripheral is enabled.

## 2.6 TSC

### 2.6.1 Inhibited acquisition in short transfer phase configuration

#### Description

Some revisions of the reference manual may omit the information that the following configurations of the `TSC_CR` register are forbidden:

- The `PGPSC[2:0]` bitfield set to 000 and the `CTPL[3:0]` bitfield to 0000 or 0001
- The `PGPSC[2:0]` bitfield set to 001 and the `CTPL[3:0]` bitfield to 0000

Failure to respect this restriction leads to an inhibition of the acquisition.

This is a documentation inaccuracy issue rather than a product limitation.

#### Workaround

No application workaround is required.

### 2.6.2 TSC signal-to-noise concern under specific conditions

#### Description

$V_{DD}$  equal to or greater than  $V_{DDA}$  may lead (depending on part) to some degradation of the signal-to-noise ratio on the TSC analog I/O group 2.

The lower are the sampling capacitor ( $C_S$ ) and the sensing electrode ( $C_X$ ) capacitances, the worse is the signal-to-noise ratio degradation.

#### Workaround

Apply one of the following measures:

- Maximize  $C_S$  capacitance.
- Use the analog I/O group 2 as active shield.

## 2.7 LPTIM

### 2.7.1 Device may remain stuck in LPTIM interrupt when entering Stop mode

#### Description

This limitation occurs when disabling the low-power timer (LPTIM).

When the user application clears the `ENABLE` bit in the `LPTIM_CR` register within a small time window around one LPTIM interrupt occurrence, then the LPTIM interrupt signal used to wake up the device from Stop mode may be frozen in active state. Consequently, when trying to enter Stop mode, this limitation prevents the device from entering low-power mode and the firmware remains stuck in the LPTIM interrupt routine.

This limitation applies to all Stop modes and to all instances of the LPTIM. Note that the occurrence of this issue is very low.

### Workaround

In order to disable a low power timer (LPTIMx) peripheral, do not clear its ENABLE bit in its respective LPTIM\_CR register. Instead, reset the whole LPTIMx peripheral via the RCC controller by setting and resetting its respective LPTIMxRST bit in RCC\_APBByRSTRz register.

## 2.7.2 Device may remain stuck in LPTIM interrupt when clearing event flag

### Description

This limitation occurs when the LPTIM is configured in interrupt mode (at least one interrupt is enabled) and the software clears any flag in LPTIM\_ISR register by writing its corresponding bit in LPTIM\_ICR register. If the interrupt status flag corresponding to a disabled interrupt is cleared simultaneously with a new event detection, the set and clear commands might reach the APB domain at the same time, leading to an asynchronous interrupt signal permanently stuck high.

This issue can occur either during an interrupt subroutine execution (where the flag clearing is usually done), or outside an interrupt subroutine.

Consequently, the firmware remains stuck in the LPTIM interrupt routine, and the device cannot enter Stop mode.

### Workaround

To avoid this issue, it is strongly advised to follow the recommendations listed below:

- Clear the flag only when its corresponding interrupt is enabled in the interrupt enable register.
- If for specific reasons, it is required to clear some flags that have corresponding interrupt lines disabled in the interrupt enable register, it is recommended to clear them during the current subroutine prior to those which have corresponding interrupt line enabled in the interrupt enable register.
- Flags must not be cleared outside the interrupt subroutine.

*Note:* The proper clear sequence is already implemented in the HAL\_LPTIM\_IRQHandler in the STM32Cube.

## 2.8 RTC and TAMP

### 2.8.1 Internal tamper flags not output on RTC\_OUT1 and RTC\_OUT2

#### Description

RTC\_OUT1 and RTC\_OUT2 can output the TAMPALRM signal. The TAMPALRM signal should be a logical-OR product of all external and internal tamper flags. Instead, when the TAMPOE control bit of the RTC\_CR register is set, the TAMPALRM signal is a logical-OR product of external tamper flags only, ignoring the internal tamper flags.

#### Workaround

None.

### 2.8.2 Notification of illegal access to secured registers is not reliable

#### Description

When an RTC or a TAMP register is globally protected against non-secure accesses, the RTC and TAMP illegal access flag should be raised in the TrustZone illegal access controller upon non-secure accesses. However, the operation of this flag is not reliable. Consequently, it must not be used by the application.

*Note:* The register protection operates correctly: a write-secure-protected register ignores non-secure writes and a read-secure-protected register always returns zero upon non-secure reads.

#### Workaround

None.

### 2.8.3 RTC\_MISR and TAMP\_MISR can be read by non-privileged accesses when privilege-protected

#### Description

The RTC\_MISR register bits can be read by non-privileged accesses even if their corresponding feature is configured with privilege protection.

The TAMP\_MISR register bits can be read by non-privileged accesses even if the TAMPPRIV bit of the TAMP\_PRIVCR register is set.

#### Workaround

None.

### 2.8.4 RTC configuration changes ignored at specific conditions

#### Description

Writes to some register bits may be ignored if done within a short period after exiting Stop or Standby mode and entering Stop or Standby mode again.

The register is correctly written, but the bit value is not propagated in the RTC kernel if the duration in Run or Sleep mode is too short. This concerns the WUTE (wake-up timer enable) bit, the ALRAE and ALRBE (Alarm A and Alarm B enable) bits, the TAMPxE (Tamper x enable) bits, all bits of RTC\_CALR (the RTC calibration register), and the CWUTF (clear wake-up timer flag) bit.

The following paragraphs describe the failure mechanism for each function.

##### Enabling (or disabling) the wake-up timer:

1. The device is in Stop or Standby mode with the wake-up timer disabled (or enabled).
2. The device wakes up from low-power mode and enables (or disables) the wake-up timer.
3. The device enters Stop or Standby mode.

If the duration of the step 2 (device in Run or Sleep mode) is less than one RTCCLK period, the WUTE bit value change may not be taken into account.

##### Enabling (or disabling) alarm A or alarm B:

1. The device is in Stop or Standby mode with the alarm disabled (or enabled).
2. The device wakes up from low-power mode and enables (or disables) the alarm.
3. The device enters Stop or Standby mode.

If the duration of the step 2 (device in Run or Sleep mode) is less than two RTCCLK periods, the ALRAE or ALRBE bit value change may not be taken into account.

##### Enabling a tamper:

1. The device is in Stop or Standby mode with all tampers disabled.
2. The device wakes up from low-power mode and enables at least one tamper.
3. The device enters Stop or Standby mode.

If the duration of the step 2 (device in Run or Sleep mode) is less than two RTCCLK periods, the tamper may remain disabled.

##### Calibration register value change:

1. The device is in Stop or Standby mode with the RECALPF bit cleared.
2. The device wakes up from low-power mode and changes the RTC\_CALR value.
3. The device enters Stop or Standby mode.

If the duration of the step 2 (device in Run or Sleep mode) is less than two RTCCLK periods, the RTC\_CALR new value may not be taken into account.

##### Clearing wake-up timer flag:

1. The device is in Stop or Standby mode and WUTF is set.
2. The device wakes up from low-power mode and clears WUTF by setting the CWUTF bit of the RTC\_SCR register.
3. The device enters Stop or Standby mode.

If the duration of the step 2 (device in Run or Sleep mode) is less than two RTCCLK periods, the WUTF bit may be stuck low and cannot be set when the wake-up timer reaches zero again.

**Note:** *The same failures occur if the DBP (disable backup domain write protection) bit of the PWR register is set before changing the RTC configuration, and is cleared soon after.*

#### **Workaround**

Always keep the DBP bit set. When the device wakes up (step 2): clear the RSF flag of the RTC\_ICSR register and wait until it is set again before entering Stop or Standby mode. In case the BYPSHAD bit of the RTC\_CR register is set, clear it before the RSF flag is set. The BYPSHAD bit can then be set again by software.

### **2.8.5 Calibration formula changes when LPCAL is set**

#### **Description**

When the LPCAL bit is set, the frequency calibration formula unduly becomes:

$$f_{CAL} = f_{RTCCLK} \times \left[ \frac{(2^{20} - 1)}{(2^{20} - 1 + CALM - CALP \times 512)} \right]$$

instead of:

$$f_{CAL} = f_{RTCCLK} \times \left[ \frac{2^{20}}{(2^{20} + CALM - CALP \times 512)} \right]$$

As a consequence, the RTC frequency in the application that keeps the LPCAL bit set (to reduce power consumption) is slightly different from the frequency measured with the LPCAL bit cleared.

#### **Workaround**

In an application keeping the LPCAL bit set, apply a compensation reflecting the difference of the frequency formulas.

**Note:** *LPCAL remains set when a new calibration value is applied. Checking the calibration result is only for validation or test purposes.*

### **2.8.6 Calendar initialization may fail in case of consecutive INIT mode entry**

#### **Description**

If the INIT bit of the RTC\_ICSR register is set between one and two RTCCLK cycles after being cleared, the INITF flag is set immediately instead of waiting for synchronization delay (which should be between one and two RTCCLK cycles), and the initialization of registers may fail. Depending on the INIT bit clearing and setting instants versus the RTCCLK edges, it can happen that, after being immediately set, the INITF flag is cleared during one RTCCLK period then set again. As writes to calendar registers are ignored when INITF is low, a write occurring during this critical period might result in the corruption of one or more calendar registers.

#### **Workaround**

After exiting the initialization mode, clear the BYPSHAD bit (if set) then wait for RSF to rise, before entering the initialization mode again.

**Note:** *It is recommended to write all registers in a single initialization session to avoid accumulating synchronization delays.*

### **2.8.7 Spurious RTC alarm EXTI line event following RTC WUT interrupt**

#### **Description**

As expected, WUT interrupt (if enabled in the RTC) generates an event on the EXTI line 20 of the event controller. However, it also causes a spurious event on the EXTI line 18 that should only respond to alarm interrupts and not to WUT interrupts.

#### **Workaround**

Upon EXTI 18 event, identify the source of interrupt by reading WUT and alarm interrupt flags.

## 2.8.8 RTC\_REFIN and RTC\_OUT on PB2 not operating in Stop 2 mode

### Description

In Stop 2 low-power mode, the RTC\_REFIN function does not operate and the RTC\_OUT function does not operate if mapped on the PB2 pin.

### Workaround

Apply one of the following measures:

- Use Stop 1 mode instead of Stop 2. This ensures the operation of both functions.
- Map RTC\_OUT to the PC13 pin. This ensures the operation of the RTC\_OUT function in either low-power mode. However, it has no effect to the RTC\_REFIN function.

## 2.8.9 Setting GPIO properties of PC13 used as RTC\_ALARM open-drain output

### Description

Some reference manual revisions may omit the information that the PC13 GPIO must be set as input when the RTC\_OR register configures PC13 as open-drain output of the RTC\_ALARM signal.

*Note:* Enabling the internal pull-up function through the PC13 GPIO settings allows sparing an external pull-up resistor. This is a documentation issue rather than a product limitation.

### Workaround

No application workaround is required provided that the described GPIO setting is respected.

## 2.9 I2C

### 2.9.1 10-bit master mode: new transfer cannot be launched if first part of the address is not acknowledged by the slave

#### Description

An I<sup>2</sup>C-bus master generates STOP condition upon non-acknowledge of I<sup>2</sup>C address that it sends. This applies to 7-bit address as well as to each byte of 10-bit address.

When the MCU set as I<sup>2</sup>C-bus master transmits a 10-bit address of which the first byte (5-bit header + 2 MSBs of the address + direction bit) is not acknowledged, the MCU duly generates STOP condition but it then cannot start any new I<sup>2</sup>C-bus transfer. In this spurious state, the NACKF flag of the I2C\_ISR register and the START bit of the I2C\_CR2 register are both set, while the START bit should normally be cleared.

#### Workaround

In 10-bit-address master mode, if both NACKF flag and START bit get simultaneously set, proceed as follows:

1. Wait for the STOP condition detection (STOPF = 1 in I2C\_ISR register).
2. Disable the I2C peripheral.
3. Wait for a minimum of three APB cycles.
4. Enable the I2C peripheral again.

### 2.9.2 Wrong behavior in Stop mode when wakeup from Stop mode is disabled in I2C

#### Description

The correct use of the I2C peripheral, if the wakeup from Stop mode by I2C is disabled (WUPEN = 0), is to disable it (PE = 0) before entering Stop mode, and re-enable it when back in Run mode.

Some reference manual revisions may omit this information.

Failure to respect the above while the MCU operating as slave or as master in multi-master topology enters Stop mode during a transfer ongoing on the I<sup>2</sup>C-bus may lead to the following:

1. BUSY flag is wrongly set when the MCU exits Stop mode. This prevents from initiating a transfer in master mode, as the START condition cannot be sent when BUSY is set.
2. If clock stretching is enabled (NOSTRETCH = 0), the SCL line is pulled low by I2C and the transfer stalled as long as the MCU remains in Stop mode.  
The occurrence of such condition depends on the timing configuration, peripheral clock frequency, and I<sup>2</sup>C-bus frequency.

This is a description inaccuracy issue rather than a product limitation.

#### Workaround

No application workaround is required.

### 2.9.3 Wrong data sampling when data setup time ( $t_{\text{SU;DAT}}$ ) is shorter than one I2C kernel clock period

#### Description

The I<sup>2</sup>C-bus specification and user manual specify a minimum data setup time ( $t_{\text{SU;DAT}}$ ) as:

- 250 ns in Standard mode
- 100 ns in Fast mode
- 50 ns in Fast mode Plus

The device does not correctly sample the I<sup>2</sup>C-bus SDA line when  $t_{\text{SU;DAT}}$  is smaller than one I2C kernel clock (I<sup>2</sup>C-bus peripheral clock) period: the previous SDA value is sampled instead of the current one. This can result in a wrong receipt of slave address, data byte, or acknowledge bit.

#### Workaround

Increase the I2C kernel clock frequency to get I2C kernel clock period within the transmitter minimum data setup time. Alternatively, increase transmitter's minimum data setup time. If the transmitter setup time minimum value corresponds to the minimum value provided in the I<sup>2</sup>C-bus standard, the minimum I2CCLK frequencies are as follows:

- In Standard mode, if the transmitter minimum setup time is 250 ns, the I2CCLK frequency must be at least 4 MHz.
- In Fast mode, if the transmitter minimum setup time is 100 ns, the I2CCLK frequency must be at least 10 MHz.
- In Fast-mode Plus, if the transmitter minimum setup time is 50 ns, the I2CCLK frequency must be at least 20 MHz.

### 2.9.4 Spurious bus error detection in master mode

#### Description

In master mode, a bus error can be detected spuriously, with the consequence of setting the BERR flag of the I2C\_SR register and generating bus error interrupt if such interrupt is enabled. Detection of bus error has no effect on the I<sup>2</sup>C-bus transfer in master mode and any such transfer continues normally.

#### Workaround

If a bus error interrupt is generated in master mode, the BERR flag must be cleared by software. No other action is required and the ongoing transfer can be handled normally.



## 2.9.5 Last-received byte loss in reload mode

### Description

If in master receiver mode or slave receive mode with SBC = 1 the following conditions are all met:

- I<sup>2</sup>C-bus stretching is enabled (NOSTRETCH = 0)
- RELOAD bit of the I2C\_CR2 register is set
- NBYTES bitfield of the I2C\_CR2 register is set to N greater than 1
- byte N is received on the I<sup>2</sup>C-bus, raising the TCR flag
- N - 1 byte is not yet read out from the data register at the instant TCR is raised,

then the SCL line is pulled low (I<sup>2</sup>C-bus clock stretching) and the transfer of the byte N from the shift register to the data register inhibited until the byte N-1 is read and NBYTES bitfield reloaded with a new value, the latter of which also clears the TCR flag. As a consequence, the software cannot get the byte N and use its content before setting the new value into the NBYTES field.

For I2C instances with independent clock, the last-received data is definitively lost (never transferred from the shift register to the data register) if the data N - 1 is read within four APB clock cycles preceding the receipt of the last data bit of byte N and thus the TCR flag raising. Refer to the product reference manual or datasheet for the I2C implementation table.

### Workaround

- In master mode or in slave mode with SBC = 1, use the reload mode with NBYTES = 1.
- In master receiver mode, if the number of bytes to transfer is greater than 255, do not use the reload mode. Instead, split the transfer into sections not exceeding 255 bytes and separate them with repeated START conditions.
- Make sure, for example through the use of DMA, that the byte N - 1 is always read before the TCR flag is raised. Specifically for I2C instances with independent clock, make sure that it is always read earlier than four APB clock cycles before the receipt of the last data bit of byte N and thus the TCR flag raising.

The last workaround in the list must be evaluated carefully for each application as the timing depends on factors such as the bus speed, interrupt management, software processing latencies, and DMA channel priority.

## 2.9.6 Spurious master transfer upon own slave address match

### Description

When the device is configured to operate at the same time as master and slave (in a multi-master I<sup>2</sup>C-bus application), a spurious master transfer may occur under the following condition:

- Another master on the bus is in process of sending the slave address of the device (the bus is busy).
- The device initiates a master transfer by bit set before the slave address match event (the ADDR flag set in the I2C\_ISR register) occurs.
- After the ADDR flag is set:
  - the device does not write I2C\_CR2 before clearing the ADDR flag, or
  - the device writes I2C\_CR2 earlier than three I2C kernel clock cycles before clearing the ADDR flag

In these circumstances, even though the START bit is automatically cleared by the circuitry handling the ADDR flag, the device spuriously proceeds to the master transfer as soon as the bus becomes free. The transfer configuration depends on the content of the I2C\_CR2 register when the master transfer starts. Moreover, if the I2C\_CR2 is written less than three kernel clocks before the ADDR flag is cleared, the I2C peripheral may fall into an unpredictable state.

### Workaround

Upon the address match event (ADDR flag set), apply the following sequence.

Normal mode (SBC = 0):

1. Set the ADDRCONF bit.
2. Before Stop condition occurs on the bus, write I2C\_CR2 with the START bit low.

Slave byte control mode (SBC = 1):

1. Write I2C\_CR2 with the slave transfer configuration and the START bit low.
2. Wait for longer than three I2C kernel clock cycles.
3. Set the ADDRCF bit.
4. Before Stop condition occurs on the bus, write I2C\_CR2 again with its current value.

The time for the software application to write the I2C\_CR2 register before the Stop condition is limited, as the clock stretching (if enabled), is aborted when clearing the ADDR flag.

Polling the BUSY flag before requesting the master transfer is not a reliable workaround as the bus may become busy between the BUSY flag check and the write into the I2C\_CR2 register with the START bit set.

## 2.10 USART

### 2.10.1 RTS is active while RE = 0 or UE = 0

#### Description

The RTS line is driven low as soon as RTSE bit is set, even if the USART is disabled (UE = 0) or the receiver is disabled (RE = 0), that is, not ready to receive data.

#### Workaround

Upon setting the UE and RE bits, configure the I/O used for RTS into alternate function.

## 2.11 SPI

### 2.11.1 BSY bit may stay high at the end of data transfer in slave mode

#### Description

BSY flag may sporadically remain high at the end of a data transfer in slave mode. This occurs upon coincidence of internal CPU clock and external SCK clock provided by master.

In such an event, if the software only relies on BSY flag to detect the end of SPI slave data transaction (for example to enter low-power mode or to change data line direction in half-duplex bidirectional mode), the detection fails.

As a conclusion, the BSY flag is unreliable for detecting the end of data transactions.

#### Workaround

Depending on SPI operating mode, use the following means for detecting the end of transaction:

- When NSS hardware management is applied and NSS signal is provided by master, use NSS flag.
- In SPI receiving mode, use the corresponding RXNE event flag.
- In SPI transmit-only mode, use the BSY flag in conjunction with a timeout expiry event. Set the timeout such as to exceed the expected duration of the last data frame and start it upon TXE event that occurs with the second bit of the last data frame. The end of the transaction corresponds to either the BSY flag becoming low or the timeout expiry, whichever happens first.

Prefer one of the first two measures to the third as they are simpler and less constraining.

Alternatively, apply the following sequence to ensure reliable operation of the BSY flag in SPI transmit mode:

1. Write last data to data register.
2. Poll the TXE flag until it becomes high, which occurs with the second bit of the data frame transfer.
3. Disable SPI by clearing the SPE bit mandatorily before the end of the frame transfer.
4. Poll the BSY bit until it becomes low, which signals the end of transfer.

*Note:* The alternative method can only be used with relatively fast CPU speeds versus relatively slow SPI clocks or/and long last data frames. The faster is the software execution, the shorter can be the duration of the last data frame.

### 2.11.2 Wrong CRC in full-duplex mode handled by DMA with imbalanced setting of data counters

#### Description

When SPI is handled by DMA in full-duplex master or slave mode with CRC enabled, the CRC computation may temporarily freeze for the ongoing frame, which results in corrupted CRC.

This happens when the receive counter reaches zero upon the receipt of the CRC pattern (as the receive counter was set to a value greater, by CRC length, than the transmit counter). An internal signal dedicated to receive-only mode is left unduly pending. Consequently, the signal can cause the CRC computation to freeze during a next transaction in which DMA TXE event service is accidentally delayed (for example, due to DMA servicing a request from another channel).

#### Workaround

Apply one of the following measures prior to each full-duplex SPI transaction:

- Set the DMA transmission and reception data counters to equal values. Upon the transaction completion, read the CRC pattern out from RxFIFO separately by software.
- Reset the SPI peripheral via peripheral reset register.

### 2.11.3 CRC error in SPI slave mode if internal NSS changes before CRC transfer

#### Description

Some reference manual revisions may omit the information that the device operating as SPI slave must be configured in software NSS control if the SPI master pulses the NSS (for example in NSS pulse mode). Otherwise, the transition of the internal NSS signal after the CRCNEXT flag is set might result in wrong CRC value computed by the device and, as a consequence, in a CRC error. As a consequence, the NSS pulse mode cannot be used along with the CRC function.

This is a documentation error rather than a product limitation.

#### Workaround

No application workaround is required as long as the device operating as SPI slave is duly configured in software NSS control.

### 2.11.4 SPI master communication failure at high $f_{PCLK}$ within the specified range

#### Description

The SPI peripheral configured as master, with the SPIx\_CR1 register's CPHA bit set and BR[2:0] bitfield written with 001 ( $f_{PCLK}/4$ ), may spuriously generate an extra clock pulse at the end of the last data frame to transfer (at the end of which the clock is expected to stop) if the PCLK frequency exceeds the values as per the table. This leads to a desynchronization of the SPI data flow.

**Table 6. Maximum  $f_{PCLK}$  for safe SPI operation**

SPI instance	$f_{PCLK(max)}$ [MHz]	
	Range 1	Range 2
SPI1	66.7	>26 (OK)
SPI2	54.0	23.5

#### Workaround

None.

## Revision history

Table 7. Document revision history

Date	Version	Changes
17-Oct-2018	1	Initial release.
10-Jun-2021	2	<p>Added errata:</p> <ul style="list-style-type: none"> <li>• Current injection from VDD to VDDA through analog switch voltage booster</li> <li>• FLASH_ECCR corrupted upon reset or power-down occurring during Flash memory program or erase operation</li> <li>• HSE oscillator long startup at low voltage</li> <li>• Selected external ADC inputs unduly clamped to VDD when all analog peripherals are disabled</li> <li>• TSC signal-to-noise concern under specific conditions</li> <li>• Device may remain stuck in LPTIM interrupt when clearing event flag</li> <li>• Internal tamper flags not output on RTC_OUT1 and RTC_OUT2</li> <li>• Notification of illegal access to secured registers is not reliable</li> <li>• RTC_MISR and TAMP_MISR can be read by non-privileged accesses when privilege-protected</li> <li>• RTC configuration changes ignored at specific conditions</li> <li>• Calibration formula changes when LPCAL is set</li> <li>• Calendar initialization may fail in case of consecutive INIT mode entry</li> <li>• Setting GPIO properties of PC13 used as RTC_ALARM open-drain output</li> </ul> <p>Modified errata:</p> <ul style="list-style-type: none"> <li>• CRC error in SPI slave mode if internal NSS changes before CRC transfer</li> <li>• SPI master communication failure at high fPCLK within the specified range</li> <li>• VDIV or VSQRT instructions might not complete correctly when very short ISRs are used</li> <li>• Internal voltage reference corrupted upon Stop mode entry with temperature sensing enabled</li> <li>• Full JTAG configuration without NJTRST pin cannot be used</li> <li>• Inhibited acquisition in short transfer phase configuration</li> <li>• Device may remain stuck in LPTIM interrupt when entering Stop mode</li> <li>• Wrong behavior in Stop mode when wakeup from Stop mode is disabled in I2C</li> <li>• Wrong data sampling when data setup time (t<sub>SU;DAT</sub>) is shorter than one I2C kernel clock period</li> <li>• Last-received byte loss in reload mode</li> </ul> <p>Removed erratum <i>First double-word of Flash memory corrupted upon reset or power down while programming</i> (superseded with the erratum FLASH_ECCR corrupted upon reset or power-down occurring during Flash memory program or erase operation).</p>

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